

Microwave-Controlled Generation of Shaped Single Photons in Circuit Quantum Electrodynamics

M. Pechal,^{*} L. Huthmacher, C. Eichler, S. Zeytinoğlu, A. A. Abdumalikov, Jr., S. Berger, A. Wallraff, and S. Filipp

Department of Physics, ETH Zurich, CH-8093 Zurich, Switzerland

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Large-scale quantum information processors or quantum communication networks will require reliable exchange of information between spatially separated nodes. The links connecting these nodes can be established using traveling photons that need to be absorbed at the receiving node with high efficiency. This is achievable by shaping the temporal profile of the photons and absorbing them at the receiver by time reversing the emission process. Here, we demonstrate a scheme for creating shaped microwave photons using a superconducting transmon-type three-level system coupled to a transmission line resonator. In a second-order process induced by a modulated microwave drive, we controllably transfer a single excitation from the third level of the transmon to the resonator and shape the emitted photon. We reconstruct the density matrices of the created single-photon states and show that the photons are antibunched. We also create multi-peaked photons with a controlled amplitude and phase. In contrast to similar existing schemes, the one we present here is based solely on microwave drives, enabling operation with fixed frequency transmons.

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One of the most important challenges in the rapidly developing field of quantum information processing and quantum communication is efficient quantum state transfer between spatially separated quantum bits. Even though systems incorporating only a few qubits successfully use coupling schemes based on atomic vibrational modes [1], discrete electromagnetic modes of microwave cavities [2], or on-chip resonators [3,4], itinerant rather than localized photons are preferable as information carriers for distributing entanglement and quantum networking over larger distances [5].

A quantum channel between distant qubits can be established in a variety of ways, for example, in heralded schemes using interference and subsequent detection of photons radiated [6,7] or scattered [8] from the qubits. A deterministic approach relying on reabsorption of a photon emitted from one qubit by another has been the subject of several theoretical proposals [9,10]. This scheme requires efficient generation of single photons on demand [11–13], their entanglement with the emitting qubit [14–19], and a temporal shape of the photon that allows time reversal of the emission process. It necessitates techniques for the generation of controllably shaped single photons that have been realized with optical photons [20–23] and shown to

enable photon reabsorption [24], albeit with a limited efficiency.

Here, we utilize superconducting circuits as one of the promising platforms for quantum information processing, offering good coherence times [25–28] and strong coupling between qubits and microwave photons [29] in an easily controllable and compact solid-state system. In this so-called circuit quantum electrodynamics architecture [30], the field emitted from a superconducting circuit is confined to a one-dimensional transmission line without any additional need for spatial mode matching. It can easily be routed between different elements of a network and it has been recently shown that a classical microwave field can be received and stored with high fidelity [31,32]. For superconducting systems, photon-shaping schemes have been proposed [10] and experimentally realized based on tunable couplers controlling the emission rate of a photon localized in a resonator into a transmission line [33,34]. Systems with a fixed resonator emission rate but a tunable coupling between the resonator and the qubit [35] can also be used for photon shaping. Both of these approaches rely on flux tuning of a SQUID loop to achieve control over the qubit-resonator or resonator-transmission line coupling. However, because of the varying Josephson inductance of the loop, the frequency of the resonator changes along with the coupling. Therefore, to control the phase of the emitted photon as well as its envelope, the frequency shift needs to be compensated by an additional tunable parameter, such as the qubit-resonator detuning [36].

In this paper, we present an alternative, microwave-based approach to photon shaping. Both the amplitude and the

^{*}mpechal@phys.ethz.ch

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phase of the emitted photon are controlled by a single phase- and amplitude-modulated microwave signal that induces an effective tunable qubit-resonator coupling via a second-order process to transfer the qubit excitation to the resonator field. Since qubit frequency and coupling remain fixed in this scheme, it can be realized in circuits without additional tuning elements.

Our device, described in more detail in Appendix A, consists of an on-chip single-sided $\lambda/2$ transmission line resonator with resonance frequency $\omega_r/2\pi = 7.224$ GHz and linewidth $\kappa/2\pi = 24$ MHz coupled with strength $g/2\pi = 35$ MHz to a transmon-type superconducting circuit [37]. The transition frequency between the ground state $|g\rangle$ and the first excited state $|e\rangle$ of the transmon is tuned to $\omega_q/2\pi = 8.640$ GHz. In the presented photon-shaping protocol, we make explicit use of the multilevel structure of the transmon, approximating it as an effective three-level ladder-type system with the transition frequency between the first and the second excited state $|f\rangle$ offset from ω_q by the anharmonicity $\alpha/2\pi = -421$ MHz. The direct transition between states $|g\rangle$ and $|f\rangle$ is forbidden to first order.

The bare eigenstates of the qubit-resonator system are coupled by the time-independent Jaynes-Cummings interaction and by an external microwave drive [Fig. 1(a)] applied to the qubit through its gate line. The coupled system is in the dispersive regime [30]; that is, the qubit and the resonator are detuned from each other by $\Delta = \omega_q - \omega_r \gg g$. Therefore, energy exchange between the two is strongly suppressed.

To obtain a tunable Jaynes-Cummings-type coupling allowing effectively resonant swapping of excitation, we use second-order processes [38–40]. We apply a qubit drive at the frequency $\omega_d = 2\omega_q + \alpha - \omega_r$ corresponding to the

energy difference between the states $|f0\rangle$ and $|g1\rangle$, where the numbers in the kets label Fock states of the resonator. In a reference frame rotating at the frequency ω_d , the Hamiltonian of the system is

$$H(t) = \delta_q b^\dagger b + \frac{1}{2} ab^\dagger b^\dagger b b + \delta_r a^\dagger a + g(ab^\dagger + a^\dagger b) + \frac{1}{2} [\Omega^*(t)b + \Omega(t)b^\dagger],$$

with the effective transition frequencies of the transmon $\delta_q = \omega_q - \omega_d = -\Delta - \alpha$ and the resonator $\delta_r = \omega_r - \omega_d = -2\Delta - \alpha$. $\Omega(t) = \Omega_0(t)e^{i\phi(t)}$ is the complex drive strength corresponding to the real drive signal $\Omega_0(t) \cos[\omega_d t - \phi(t)]$ with slowly varying amplitude $\Omega_0(t)$ and phase $\phi(t)$. The operators a , a^\dagger are the annihilation and creation operators of the resonator and b , b^\dagger their analogs for the transmon $b = |g\rangle\langle e| + \sqrt{2}|e\rangle\langle f| + \sqrt{3}|f\rangle\langle h| + \dots$ [37]. This simplified picture of the transmon as an anharmonic oscillator with a Kerr-type nonlinearity is a good approximation in the limit of small anharmonicity; that is, $|\alpha| \ll \omega_q$. In the sample used for our experiment, we have $|\alpha|/\omega_q \approx 1/20$ and we employ high drive amplitudes up to $\Omega/2\pi \approx 1$ GHz. To obtain good quantitative agreement between the experimental data and numerical simulations, we need to use the full model of the transmon-resonator system [37].

The states $|f0\rangle$ and $|g1\rangle$ are resonant in the rotating frame, while the intermediate states $|e0\rangle$ and $|e1\rangle$ are far off resonant. They can, therefore, be adiabatically eliminated, giving rise to an effective Jaynes-Cummings-type coupling between the resonator and the two-level system consisting of qubit states $|g\rangle$ and $|f\rangle$ [Fig. 1(b)]. This interaction is described by the effective Hamiltonian

$$H_{\text{eff}}(t) = \Delta_{f0g1}(t)|f0\rangle\langle f0| + \tilde{g}(t)|f0\rangle\langle g1| + \text{H.c.}, \quad (1)$$

which we simplify by absorbing the parts of the ac Stark shifts independent of Ω into the renormalized frequencies of the qubit and the resonator. The remaining ac Stark shift $\Delta_{f0g1}(t)$ of the transition frequency between $|f0\rangle$ and $|g1\rangle$ is to leading order quadratic in the drive strength Ω . Using perturbation theory, the effective second-order coupling $\tilde{g}(t)$ [41] can be expressed as

$$\tilde{g}(t) = \frac{1}{\sqrt{2}} \frac{g\alpha}{\Delta(\Delta + \alpha)} \Omega(t). \quad (2)$$

Within the second-order approximation valid for coupling g and drive strength Ω much smaller than $|\Delta|$ and $|\Delta + \alpha|$, the strength of the effective coupling is limited to $|\tilde{g}(t)| \ll |\alpha|$. In typical transmon circuits with anharmonicities on the order of few hundred MHz, this restricts $\tilde{g}/2\pi$ to values below approximately 10 MHz and, thus, the shortest time in which a photon can be generated using this scheme is roughly $\pi/\tilde{g} \gtrsim 50$ ns. To suppress decoherence during the

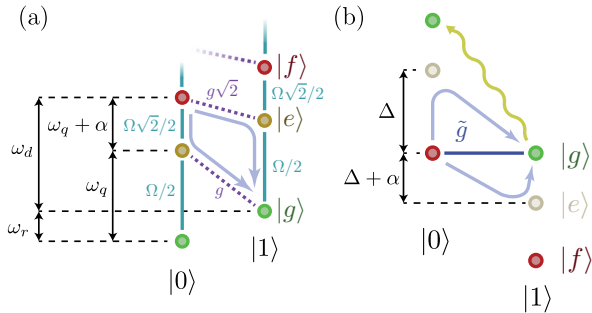


FIG. 1. (a) Energy level diagram of the qubit-resonator system. The matrix elements of the drive Ω and the Jaynes-Cummings coupling g are indicated by solid lines and dotted lines, respectively, connecting the coupled bare states. The two second-order paths connecting the states $|f0\rangle$ and $|g1\rangle$ are indicated by arrows. (b) In the rotating frame of the drive, the states $|f0\rangle$ and $|g1\rangle$ are nearly resonant. The second-order effective coupling \tilde{g} between them after adiabatic elimination of the intermediate states $|e0\rangle$ and $|e1\rangle$ is indicated by the solid blue line. The decay of the state $|g1\rangle$ into $|g0\rangle$ by photon emission is shown by the yellow arrow.

emission process, stronger effective couplings and, therefore, shorter photon pulses may be achievable with more anharmonic qubits, such as the fluxonium [42]. Alternatively, transmon geometries with longer coherence times [26–28] may be employed.

In analogy to the optical domain experiment with atoms in Ref. [21], we use the tunable effective coupling to generate single photons by controllably transferring population of the $|f0\rangle$ state into $|g1\rangle$, which decays into $|g0\rangle$ by photon emission. Since the drive is off resonant from the $|g0\rangle \rightarrow |e0\rangle$ transition, the system remains trapped in the ground state [Fig. 1(b)], ensuring that only a single photon is emitted. By modulating the amplitude and the phase of the drive signal in time using sideband mixing, we control the temporal shape of the output field $a_{\text{out}}(t) = \sqrt{\kappa}a(t)$ [43], resulting in the emission of a single photon state $|1\rangle = \int \psi(t)a_{\text{out}}^\dagger(t)|0\rangle dt$ characterized by its mode function $\psi(t)$.

For the initial characterization of the created single-photon states $|1\rangle$, we prefer to analyze their superposition with vacuum ($|0\rangle + |1\rangle$)/ $\sqrt{2}$. In contrast to a single-photon Fock state $|1\rangle$, this superposition state has a nonzero average voltage proportional to the output field $\langle a_{\text{out}}(t) \rangle$ [44] given by $\psi(t)/2$. In circuit QED, this voltage is readily determined using standard heterodyne measurements, which also capture the phase of the mode function $\psi(t)$. In addition, we characterize single-photon Fock states $|1\rangle$ directly by measuring the emitted power proportional to $\langle a_{\text{out}}^\dagger(t)a_{\text{out}}(t) \rangle = |\psi(t)|^2$ and also higher-order moments of the output field that allow us to extract their full density matrices [13,45].

In order to prepare the photon superposition state $(|0\rangle + |1\rangle)/\sqrt{2}$, we drive the $|g\rangle \rightarrow |e\rangle$ and $|e\rangle \rightarrow |f\rangle$ transitions sequentially [Fig. 2(a)] to initialize the transmon in the state $(|g\rangle + |f\rangle)/\sqrt{2}$. This state is then mapped coherently onto $(|0\rangle + |1\rangle)/\sqrt{2}$ by driving the second-order process discussed above with a pulse [Fig. 2(b)] of the form

$$\Omega(t) = \Omega_0 \sin^2(\pi t/T) \exp[i\phi(t)]. \quad (3)$$

The pulse $\Omega(t)$ is parametrized by its maximum amplitude Ω_0 , its duration T , and its phase $\phi(t)$. We adjust the phase $\phi(t)$ to compensate for the amplitude-dependent Stark shift $\Delta_{f0g1}(t)$, keeping the phase of the photon waveform constant over the duration of the pulse (see Appendix B). We choose this \sin^2 shape of the drive pulse as a particularly simple continuous function with only two free parameters and—unlike other candidates such as a Gaussian function—with no need for an arbitrary cutoff.

To determine the waveform of the photon state, we measure both voltage quadratures $I(t)$ and $Q(t)$ of the output field emitted into the detection line in a heterodyne setup [13]. The averaged value $\langle v(t) \rangle$ of the complex signal

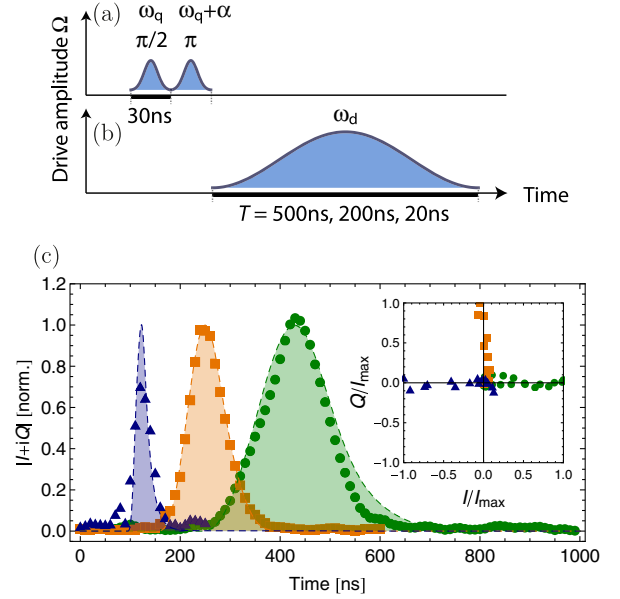


FIG. 2. (a) Amplitude of the drive signal used to initialize the transmon and (b) transfer the excitation into the resonator to generate a symmetric photon shape. (c) The normalized voltage amplitude $|I + iQ|$ of shaped photons obtained for drive pulses with $T = 20$ ns, $\Omega_0/2\pi = 680$ MHz (blue triangles), $T = 200$ ns, $\Omega_0/2\pi = 700$ MHz (orange squares), and $T = 500$ ns, $\Omega_0/2\pi = 600$ MHz (green circles). The dashed lines show the simulated photon shapes scaled to normalize their peak values to unity. The measured traces are fitted to the simulation with only the time shifts and scaling factors as fit parameters. The inset shows both voltage quadratures I and Q of the photon pulses.

$v(t) = I(t) + iQ(t)$ is then proportional to the expectation value $\langle a_{\text{out}}(t) \rangle = \psi(t)/2$. To detect the weak single-photon signal, a broadband low-noise high-electron mobility transistor (HEMT) amplifier is used. For measurements requiring only narrow bandwidth not exceeding approximately 10 MHz, we also employ a phase-preserving Josephson parametric amplifier [46,47] to operate at higher signal-to-noise ratio [45]. A boxcar filter is applied to the digitized signal to reduce noise and to filter out the pump tone of the parametric amplifier as well as to compensate the dc offset of the analog-to-digital converter.

We perform measurements of the photon waveform for different values of T and Ω_0 to study the dependence of the photon shape on these parameters [see Appendix B and Figs. 6(a) and 6(b)] and to maximize its symmetry. We quantify the photon symmetry using a parameter s given by the scalar product of the waveform with its time inverse [see Eq. (B1)]. By choosing T and Ω_0 resulting in high values of the symmetry parameter s , we are able to prepare symmetric photons of controlled length in good approximation, as shown in Fig. 2(c). Performing qubit tomography measurements after the photon emission process shows that, for the two longer photon pulses ($T = 200$ ns, $T = 500$ ns), the initial $|f0\rangle$ state is nearly emptied by

the drive pulse, having a residual population on the order of 1%–2%. On comparable time scales, the reduction of the population due to relaxation is negligible. The peak drive strengths for these pulses are $\Omega_0/2\pi = 700$ MHz and $\Omega_0/2\pi = 600$ MHz, respectively. The corresponding peak amplitudes of the effective coupling $\tilde{g}/2\pi$ given by Eq. (2) are 5.2 and 4.4 MHz, consistent with numerical diagonalization of the Hamiltonian, which yields values of $\tilde{g}/2\pi = 5.5$ and 4.6 MHz. The symmetry parameter s reaches a value of 0.98 for both longer pulses. This high symmetry can be obtained only for pulses much longer than the cavity rise time $1/\kappa \approx 7$ ns. For comparison, we show a short pulse with $T = 20$ ns. This photon pulse is not long enough for a complete population transfer from $|f0\rangle$ to $|g1\rangle$ with the drive pulse amplitudes used, which leads to a reduced emission efficiency. Its symmetry parameter $s = 0.92$ also does not reach the high values obtained for the longer pulses.

To prepare a symmetric photon state that can be easily reabsorbed by a quantum node, it is important that not only its amplitude but also its phase is symmetric in time. We achieve this using the ac Stark shift calibration procedure described in Appendix B. By choosing the time-dependent phase $\phi(t)$ of the drive pulse given by Eq. (3) such that $\dot{\phi}(t) = -\Delta_{f0g1}(t)$, we make the phase of the photon pulse constant in time, as illustrated in the inset of Fig. 2(c), noting that the ratio of the two signal quadratures is in good approximation time independent for each of the three pulses.

The system dynamics including imperfections of the second-order transition and decoherence are modeled by solving the master equation for the full Jaynes-Cummings Hamiltonian including three resonator and six transmon levels and evaluating the output field $a_{\text{out}}(t)$ to obtain the simulation data in Fig. 2(c).

To demonstrate the single-photon nature of the emitted field, we measure its moments using the propagating field tomography method described in detail in Refs. [45,48]. We first determine the mode function $\psi(t)$ of the photon pulse, which is proportional to the observed averaged coherent signal $\langle v(t) \rangle$. Then, in each realization of the experiment, the measured single-shot voltage $v(t)$ is processed by a digital Chebyshev filter with a shape approximately matched to $\psi(t)$ and the result V is recorded in a 2D histogram. The observable V can be described by an operator $A + h^\dagger$, where A is the temporal field mode $A = \int \psi^*(t) a_{\text{out}}(t) dt$ and h is a noise mode, assumed to be in a thermal state that is characterized by the effective noise temperature of the amplification chain [45]. The moments $\langle (V^*)^m V^n \rangle$ extracted from the recorded histogram and the noise moments $\langle (h^\dagger)^k h^l \rangle$ determined in an equivalent measurement with only vacuum at the input of the detection chain are used to calculate the field moments $\langle (A^\dagger)^i A^j \rangle$ [45].

For both the symmetrically shaped photon superposition state $(|0\rangle + |1\rangle)/\sqrt{2}$ and the single-photon Fock state $|1\rangle$

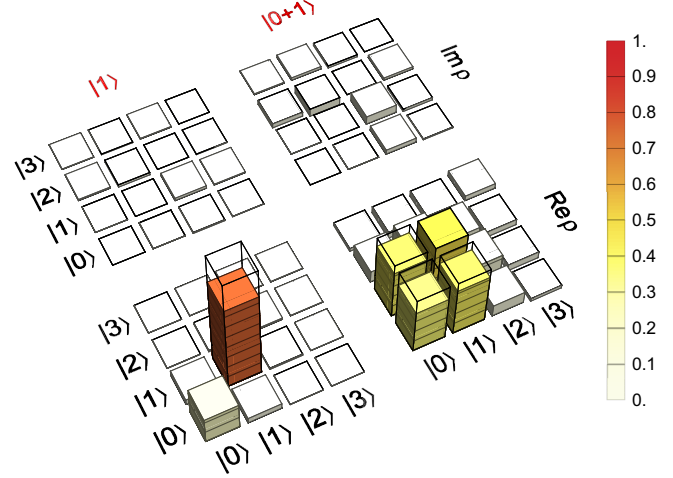


FIG. 3. The real and imaginary parts of the measured density matrices ρ of the symmetric temporal photon mode with fidelities $F = 76\%$ and $F = 86\%$ to the respective ideal states $|1\rangle$ and $(|0\rangle + |1\rangle)/\sqrt{2}$. The wire frames show the single-photon Fock state $|1\rangle$ (all density matrix elements equal to 1 or 0) and the ideal superposition state $(|0\rangle + |1\rangle)/\sqrt{2}$ (all density matrix elements equal to $1/2$ or 0). The ticks in the bars represent increments of 0.1 .

described above, we find the normalized fourth-order moments $g^{(2)}(0) = \langle A^\dagger A^\dagger A A \rangle / \langle A^\dagger A \rangle^2$ of 0.03 ± 0.07 and 0.06 ± 0.02 , respectively, which lie well below the classical limit of $g^{(2)}(0) = 1$ expected for coherent states, showing a high degree of antibunching. The density matrices ρ with fidelities $F = 86\%$ and 76% of the respective photon states shown in Fig. 3 are extracted from the measured moments [45] by employing a maximum likelihood algorithm. From the numerical simulation of the emission process, we find the normalization conditions $\langle A^\dagger A \rangle = 0.39$ and 0.79 . This is lower than the values of $\langle A^\dagger A \rangle = 1/2$ and 1 expected for the ideal states $(|0\rangle + |1\rangle)/\sqrt{2}$ and $|1\rangle$ due to the reduced photon emission efficiency of $(79 \pm 1)\%$ limited by the finite lifetime $T_1^f = (550 \pm 5)$ ns of the $|f\rangle$ state. This accounts for the deviation of the diagonal elements of ρ from the theoretical values while the off-diagonal elements are reduced due to the loss of coherence between the qubit states $|g\rangle$ and $|f\rangle$ on a time scale of $T_2^{gf} = (580 \pm 30)$ ns. This estimate of the emission efficiency assumes perfect initialization of the transmon in the $|f\rangle$ state in order to evaluate the shaping process separately from the preparation procedure. The total efficiency including a realistic initial state preparation is approximately 76% , that is, about 6% lower due to relaxation during the initialization pulses and thermal population of the excited states (see Appendix A).

To demonstrate the rapid amplitude- and phase-modulation capability of our all-microwave photon-shaping scheme, we prepare six-peaked single-photon pulses similar to the double-peaked pulses demonstrated with optical frequency photons [21]. For this purpose the transmon is

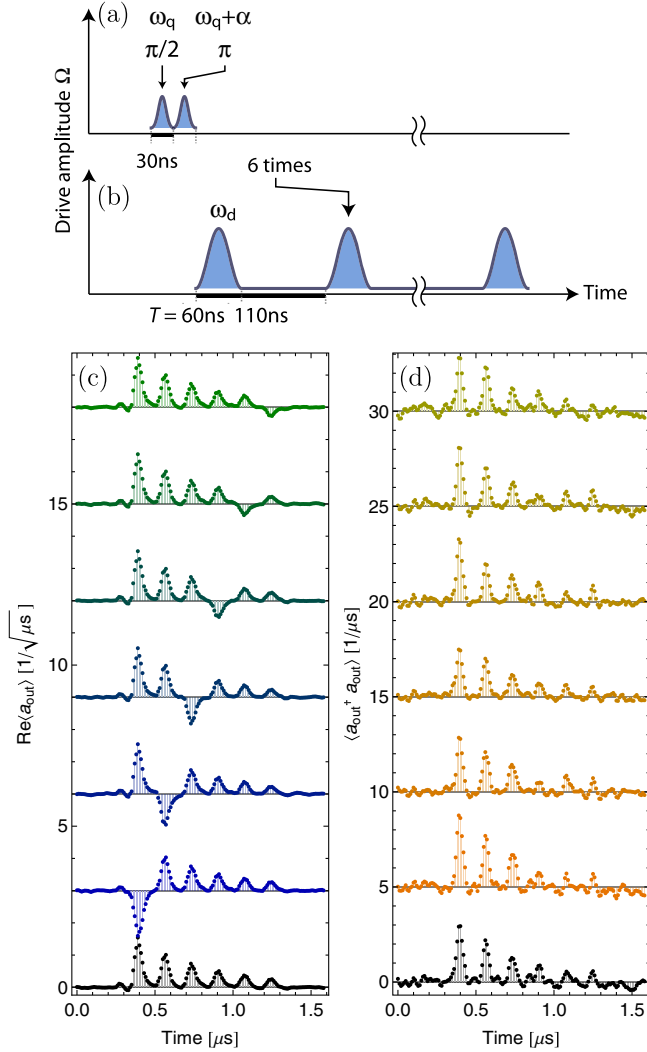


FIG. 4. (a) Amplitude of the drive signal used to initialize the transmon and (b) transfer the excitation into the resonator to generate a multi-peaked photon shape. (c) The real part of the measured output field a_{out} as a function of time for six-peak photons with phases of each of the peaks flipped by π one by one. The bottom-most trace is a reference with all peaks having the same phase. (d) Averaged emitted power as a function of time for the same photon pulses as in (c).

again prepared in the state $(|g\rangle + |f\rangle)/\sqrt{2}$ [Fig. 4(a)]. The subsequent photon-shaping signal consists of a train of six identical \sin^2 pulses of amplitude $\Omega_0/2\pi \approx 350$ MHz and length $T = 60$ ns separated by 170 ns [Fig. 4(b)]. These parameters are adjusted to make the overlap between the photon peaks small while keeping the overall duration of the pulse train short to minimize decoherence.

We show that the phases of the individual peaks in the photon waveform can be controlled independently. As an example, we change the phase of any one of the subsequent photon peaks by π by adjusting the phase of the corresponding drive pulses leading to a change of sign in the detected voltage, Fig. 4(c). This phase control can be

achieved while keeping the emitted power unchanged, as illustrated in Fig. 4(d). The power is measured as a function of time by squaring and subsequent averaging of the digitized and filtered voltages. The noise power is then subtracted in postprocessing.

In conclusion, we present a photon-shaping technique relying fully on a phase- and amplitude-controlled microwave drive acting on three transmon levels. We show that this method can be used to generate single-photon pulses of symmetric shapes with a controllable amplitude and phase. We prepare a multi-peaked photon with individually tunable phases of all the peaks. Such photon states may be used to encode information in the rich multidimensional time-bin Hilbert space of a traveling photon [49]. The shaping method can be refined by using simulations and optimization techniques to find drive pulses needed to generate desired photon shapes. With the time-reversed scheme applied to a second distant qubit, the emitted photon can, in principle, be reabsorbed to map its quantum state onto a qubit [9]. The simple nature of this scheme with respect to the required control elements also makes it a prime candidate for use in 3D circuit QED architectures [25].

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APPENDIX A: SYSTEM PARAMETERS

Here, we provide supplemental information about parameters of the sample, measurement setup, and methods.

We pattern the transmission line resonator using photolithography on a niobium-coated sapphire wafer [Fig. 5(a)]. The transmon circuit is fabricated by e -beam lithography and shadow evaporation of aluminium. We measure the charging energy $E_C/h \approx 406$ MHz and maximum Josephson energy $E_J^{\text{max}}/h \approx 25$ GHz of the device. To tune the transition frequency of the transmon, we apply a magnetic flux bias using a superconducting coil, and to measure its quantum state, we employ dispersive readout [50] by probing the reflection coefficient of the resonator [Fig. 5(b)].

The microwave pulses used to initialize the transmon in the $|f0\rangle$ state and to transfer its excitation into the resonator are generated by single sideband mixing using in-phase and quadrature channels synthesized by an arbitrary waveform generator (AWG).

The output of the transmission line resonator is routed to a Josephson parametric dimer (JPD) amplifier of the type described in Ref. [47] via two circulators—the first serving to separate the incoming drive signal from the output of the resonator, the second to separate the input and output of the parametric amplifier—and a directional coupler used to pump the amplifier [Fig. 5(b)]. The pump tone is split and

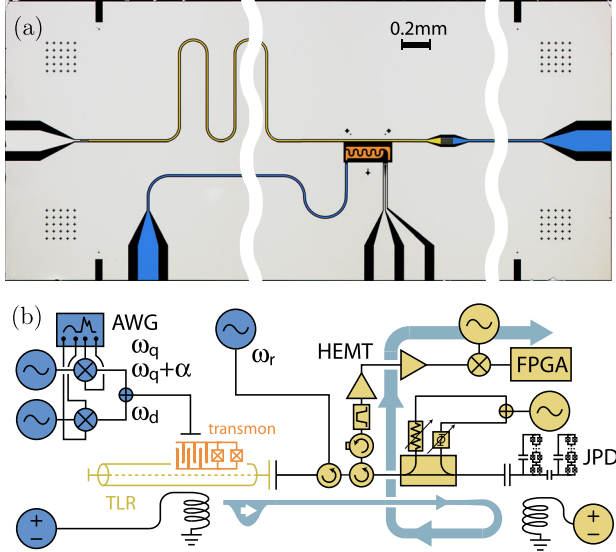


FIG. 5. (a) A false color microscope image of the chip showing the transmission line resonator coupled to the transmon circuit (rectangular structure in the middle) and the microwave control lines: the resonator input/output port, the transmon charge line (blue), and the unused flux line (gray). (b) Schematic of the measurements setup (see Appendix A). The control electronics is shown in blue, the measurement chain in yellow, and the transmon circuit coupled to the transmission line resonator (TLR) in orange.

applied to two inputs of the directional coupler with relative phase and amplitude chosen to ensure destructive interference of the pump tone in the amplified signal. We operate the parametric amplifier in the phase-preserving mode with a pump detuned from the signal by 240 MHz. Its gain is adjusted to approximately 20 dB with a bandwidth of 30 MHz. For measurements that do not require the added gain, the parametric amplifier is not pumped and the signal is reflected with a gain of unity. The signal is then filtered by a 4–8 GHz bandpass filter and amplified by a cryogenic HEMT amplifier with a gain of 35 dB. At room temperature, the signal is further amplified by 60 dB and down-converted with a local oscillator detuned by 25 MHz from the resonator frequency. The resulting intermediate frequency (IF) signal is digitized by a Field-Programmable Gate Array (FPGA) equipped with an analog-to-digital converter at a rate of 100 MS/s. The two quadrature components of the IF signal are then determined by digital down-conversion and low-pass filtering and are further averaged. Alternatively, their values are integrated and the result recorded in a histogram. The efficiency of the detection chain is characterized by comparing the measured noise floor with the measured power of a single-photon pulse, resulting in a noise number of $N_0 \approx 10$, which corresponds to a detection efficiency of approximately 9%. This efficiency is limited predominantly by the losses between the sample and the parametric amplifier, the noise

added by the detection chain, and imperfections in mode matching due to the used Chebyshev digital filter being only an approximation of the true photon shape.

The state preparation pulses driving the qubits from $|g\rangle$ to $|e\rangle$ and from $|e\rangle$ to $|f\rangle$ are Gaussian with a standard deviation $\sigma = 5$ ns truncated to a finite length of 6σ . We calibrate their amplitudes by Rabi oscillation measurements and their frequencies using Ramsey interferometry. The latter measurement is also used to extract the dephasing times of the transition $T_2^{ge} = (1640 \pm 50)$ ns, $T_2^{ef} = (557 \pm 8)$ ns, and $T_2^{gf} = (580 \pm 30)$ ns. The relaxation times $T_1^e = (2000 \pm 200)$ ns and $T_1^f = (550 \pm 5)$ ns are determined by time-resolved measurements of the excited-state population using three-level quantum state tomography [51].

The steady-state thermal population of the first excited state of the transmon is measured to be approximately 13%, significantly higher than the theoretical equilibrium value 0.2% corresponding to the physical base temperature of 50 mK. The source of the excess thermal population is likely due to undesired elevated temperatures of the still and 100 mK stages of the employed cryostat. To prepare the transmon in its ground state, we swap the thermal population of the first excited state $|e\rangle$ into $|f\rangle$ and make use of the $|f0\rangle \rightarrow |g1\rangle$ transition to further transfer it into the resonator. As the resonator relaxation rate κ is much higher than the qubit decay rate, this results in a decrease of the total system energy. By repeating the cooling step several times, we reach $|e\rangle$ state thermal population of approximately 3%. This also indicates that the mean thermal excitation of the resonator is significantly lower than that of the qubit.

APPENDIX B: CALIBRATION OF THE PHOTON-SHAPING DRIVE PULSE

Here, we explain the methods used to calibrate the photon-shaping drive pulses.

To determine the optimal values of Ω_0 and T in the drive pulse [Eq. (3)] for generating a symmetric photon shape, we measure the heterodyne voltage of the shaped superposition state $(|0\rangle + |1\rangle)/\sqrt{2}$ for a range of T between 60 and 500 ns and $\Omega_0/2\pi$ between 0 and 1000 MHz. A set of the measured photon pulse waveforms chosen to illustrate the dependence of the shape on Ω_0 and T is shown in Figs. 6(a) and 6(b).

The observed effect of the drive pulse length on the length of the photon pulse and the efficiency of its emission, which is reflected in the amplitude of the detected voltage, is illustrated in Fig. 6(a). Shorter drive pulses lead to shorter photon waveforms (curves at the bottom of the plot) and vice versa.

The influence of the drive pulse amplitude is shown in Fig. 6(b). Photon shapes generated with stronger drive pulses are shorter and display signatures of Rabi oscillations (the smaller side peaks in the two curves at the top of the plot). On the other hand, weak pulses result in an

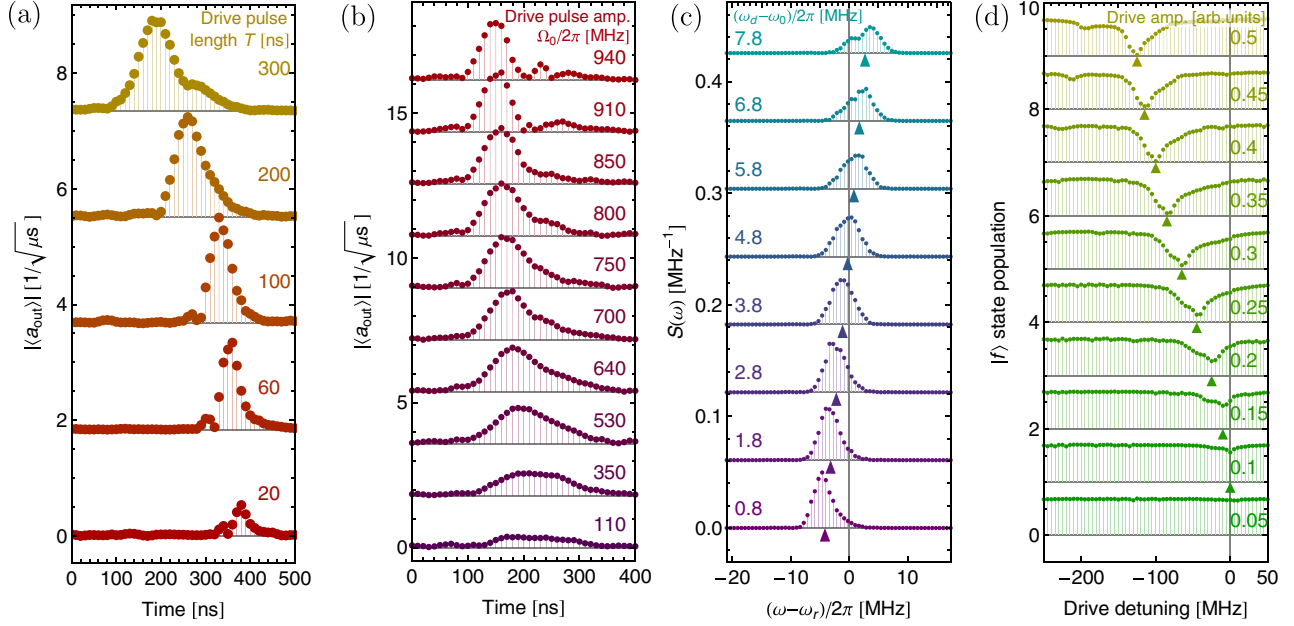


FIG. 6. (a) The measured voltage waveform of the photon superposition state $(|0\rangle + |1\rangle)/\sqrt{2}$ for the indicated length of the drive pulse T at a fixed pulse amplitude $\Omega_0/2\pi = 420$ MHz and (b) for the indicated amplitude of the pulse Ω_0 at a fixed pulse length $T = 300$ ns. (c) Fourier transform of the measured photon pulse waveform obtained for $\Omega_0/2\pi = 600$ MHz and $T = 500$ ns in a reference frame rotating at the resonator frequency ω_r . The different curves correspond to the indicated values of the detuning between the drive frequency ω_d and the expected zero-amplitude limit of the $|f0\rangle \rightarrow |g1\rangle$ transition frequency $\omega_0 = 2\omega_q + \alpha - \omega_r$. The triangle marks the center frequency of the peak. (d) Population of the $|f\rangle$ state after a 100 ns square pulse driving the $|f0\rangle \rightarrow |g1\rangle$ transition plotted versus detuning of the pulse from ω_0 for the indicated pulse amplitudes (shown here in units of the full scale output amplitude of the AWG). The triangles mark the minima of the curves.

incomplete population transfer and, therefore, reduced efficiency of the emission process, as illustrated by the reduced amplitude of the bottom-most waveform.

In order to quantify the symmetry of the measured waveform, we calculate the overlap s of the averaged signal $\langle v(t) \rangle = \langle I(t) + iQ(t) \rangle$ with the time-reversed copy of itself:

$$s = \max_{t_0} \frac{|\int \langle v(2t_0 - t) \rangle^* \langle v(t) \rangle dt|}{\int |\langle v(t) \rangle|^2 dt}. \quad (\text{B1})$$

We then numerically maximize s with respect to the amplitude Ω_0 and length T of the drive pulse. We find values of s close to 1 for various combinations of these parameters, e.g., $s = 0.98$ for $\Omega_0/2\pi = 700$ MHz, $T = 200$ ns, and $s = 0.99$ for $\Omega_0/2\pi = 600$ MHz, $T = 500$ ns, whose corresponding photon waveforms are also shown in Fig. 2(c).

The maximization of the symmetry parameter also helps us fine-tune the frequency of the drive pulse to resonance with the $|f0\rangle \rightarrow |g1\rangle$ transition. Since a constant detuning leads to a linear phase drift in the emitted photon waveform, the symmetry value s is reduced when the drive pulse is off resonant. Alternatively, we can fine-tune the drive frequency using direct measurements of the photon frequency as shown in Fig. 6(c). Here, we plot the Fourier transform

of the photon waveform and observe that its peak shifts as the drive frequency ω_d is tuned across the $|f0\rangle \rightarrow |g1\rangle$ transition frequency $\omega_0 = 2\omega_q + \alpha - \omega_r$ expected in the zero-amplitude limit. We then adjust the drive frequency to eliminate the detuning of the photon. The results of these two methods for calibrating the drive frequency are consistent with each other.

The conversion between the output amplitude of the drive signal set at the AWG and the drive parameter $\Omega(t)$ is obtained using measurements of the ac Stark shift presented in Fig. 6(d). We prepare the transmon in the $|f\rangle$ state, apply pulses with different frequencies close to ω_0 , and extract the drive detuning at which the $|f\rangle$ state is maximally depleted after the pulse. This detuning determines the Stark shift. By repeating the measurement for different amplitudes of the drive pulse and comparing the obtained Stark shifts with the results of numerical diagonalization of the transmon-resonator Hamiltonian, we can match each AWG output amplitude with the corresponding value of Ω_0 .

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