

Inverse design of higher-order photonic topological insulators

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The recently discovered second-order photonic topological insulators (SPTIs) are characterized by gapped edge states and robust corner states, and they provide novel approaches to the traditional ways to manipulate light. In a general case, the overlapped band gap of nontrivial and trivial photonic crystals composing SPTIs is narrow, which barely allows for the production of strongly localized states. Here, we introduce an intelligent numerical approach for the inverse design of large classes of SPTIs with great flexibility for controlling the properties of topological edge and corner states. In the optimized designs, the overlapped band gap of the nontrivial and trivial photonic crystals substantially exceeds that of the existing SPTIs, and it enables the existence of highly localized corner and edge states with nearly flat dispersion. We design several structures supporting both topological edge and corner states at the desired frequencies. Through programming newly created SPTIs, we suggest a strategy for routing topological edge and corner states. Our findings pave the way for the development of integrated photonic devices with topological protection and innovative functionalities.

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I. INTRODUCTION

The discovery of topological phases opened up new horizons in the study of condensed-matter physics [1–6]. Topological insulators with robust gapless edge states were expected to deliver a broad range of applications ranging from spintronics to quantum computation [7–9]. Recently, the concept of topological phases has entered the realm of photonics [10–14], revolutionizing some of the traditional views on light manipulation and light-matter interaction. Many novel photonic devices, such as backscattering-immune sharply bent waveguides [15], spin-polarized switches [16], robust delay lines [17], nonreciprocal devices [18], and topological laser [19,20], have been predicted and demonstrated in experiment. Recently, a novel type of so-called second-order topological insulator featuring gapped edge states and in-gap corner/hinge states has been proposed in the field of electronics [21–25], and it was then extended to photonic systems [26,27]. The second-order corner states of photonic crystals (PCs) [28–30] have been experimentally verified, and the photonic crystal nanocavity with a high Q factor based on a topological corner state has been demonstrated [31].

The occurrence of topological states and corner states depends on the spatial distribution of the constituent material

within the unit cell of trivial and nontrivial photonic crystals. The existing designs of second-order photonic topological insulators (SPTIs) mainly relies on the trial-and-error approach. The resulting properties of the topological insulators may be far from optimum. For instance, the overlapped band gap of nontrivial and trivial PCs is relatively narrow and barely produces strongly localized topological edge and corner states. Topological photonics provides an unprecedented means for developing next-generation integrated photonic devices [10], ranging from optical interconnects [32] to quantum technologies [33]. One of the key elements of integrated photonic devices is the wavelength router, which separates light with various wavelengths into different channels and dramatically increases the data capacity in a fiber or waveguide [34,35]. However, wavelength routers utilizing topological edge and corner states have never been predicted and explored. The design of wavelength routers requires that topological edge and corner states occur at various frequency levels, which creates a challenge for the current design approach.

This paper presents an intelligent topology optimization approach for designing SPTIs and other types of higher-order topological structures. We uncover the procedure to achieve the overlapped band gap of the created nontrivial and trivial PCs for more than twice that of the current SPTI designs, and we demonstrate highly localized corner states and an edge state with nearly flat dispersion. In addition, the proposed approach provides great flexibility for tuning the operating frequencies of the topological edge and corner states. By programming newly created SPTIs, we demonstrate high-performance four-channel photonic routers for topological edge states and three-channel photonic routers for topological corner states.

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II. INVERSE-DESIGN APPROACH

Topology optimization employs a robust numerical algorithm to seek the layout of materials within the design domain, so as to minimize or maximize the defined objective function [36]. Topology optimization has been proven to be an efficient and effective tool for the design of PCs [37–46]. Here, we will extend topology optimization to design the unit cell of nontrivial and trivial PCs for SPTIs. The 2D hexagonal unit cells of PCs are made of silicon (the relative permittivity, $\varepsilon = 12$) with C_{6v} symmetry. The lattice constant is a , and the frequency is normalized as $\Omega = \omega a/2\pi c$, where c denotes the speed of light in vacuum. The transverse magnetic (TM) mode is assumed, where the magnetic field is confined in the xy plane and the electric field is perpendicular to the plane.

Topology optimization first constructs a nontrivial PC (NP), which has two dipolar modes located above two quadrupolar modes in the band diagram. The unit cell is discretized with N finite elements, and the design variables, x_e ($e = 1, 2, \dots, N$), are the artificial densities of elements in the unit cell. $x_e = 0$ denotes that element e is full of air and $x_e = 1$ for the specified dielectric material, e.g., silicon in our examples. The optimization objective is to find the optimal discrete values of x_e , so that the dipolar modes and quadrupolar modes occur at the specified frequencies, ω_1 and ω_2 ($\omega_1 > \omega_2$). The specific resonant mode at the specified frequency can be excited by maximizing the frequency-averaged local density of states (LDOS) [41,47,48], $\tilde{f}(\omega, \mathbf{J}, \mathbf{E})$, emitted from a harmonic current source, \mathbf{J} . The electric field, \mathbf{E} , is the solution of the steady-state Maxwell equation, $\nabla \times (1/\mu)\nabla \times \mathbf{E} - \varepsilon(\mathbf{r})\omega^2\mathbf{E} = i\omega\mathbf{J}$. The current \mathbf{J} is judiciously constructed in the Supplemental Material [59], Sec. 1, to excite the dipolar and quadrupolar modes, respectively. To solve it mathematically, the problem is converted to the topology optimization formulation with objective and constraint functions as follows:

$$\begin{aligned} & \text{Find : } x_e \ (e = 1, 2, \dots, N) \\ & \text{min : } t \\ & \text{s.t. : } t \geq \frac{1}{\tilde{f}_1(\omega_1, \mathbf{J}_1, \mathbf{E}_1)}, \quad t \geq \frac{1}{\tilde{f}_2(\omega_2, \mathbf{J}_2, \mathbf{E}_2)} \\ & \quad x_e \in \{0, 1\}. \end{aligned} \quad (1)$$

\tilde{f}_1 and \tilde{f}_2 are the frequency-averaged LDOS for the dipolar and quadrupolar modes, respectively. The details on topology optimization can be found in the Supplemental Material [59].

By gradually modifying the targeted frequencies of dipolar and quadrupolar modes, we obtain a nontrivial PC (denoted as NP1) as shown in Fig. 1(a), with a wide band gap between $\Omega = 0.433$ and 0.620 in Fig. 1(b). The insets in Fig. 1(b) illustrate that the two dipolar modes, p_x/p_y , are located above the two quadrupolar modes, $d_{x^2-y^2}/d_{xy}$, at Γ . Such a parity-inverse band order indicates that the band gap is nontrivial [16,18]. Next, we create a trivial PC (denoted as TP) by topology optimization [49], which aims to maximize the band gap between the second band and the third band. The optimized unit cell and its band diagram are given in Figs. 1(c) and 1(d), respectively, and its band gap ranges from $\Omega = 0.434$ to 0.583 . The overlapped band-gap width between the nontrivial and trivial PCs achieves 29.3%, which is more than twice that compared with those of the intuitive

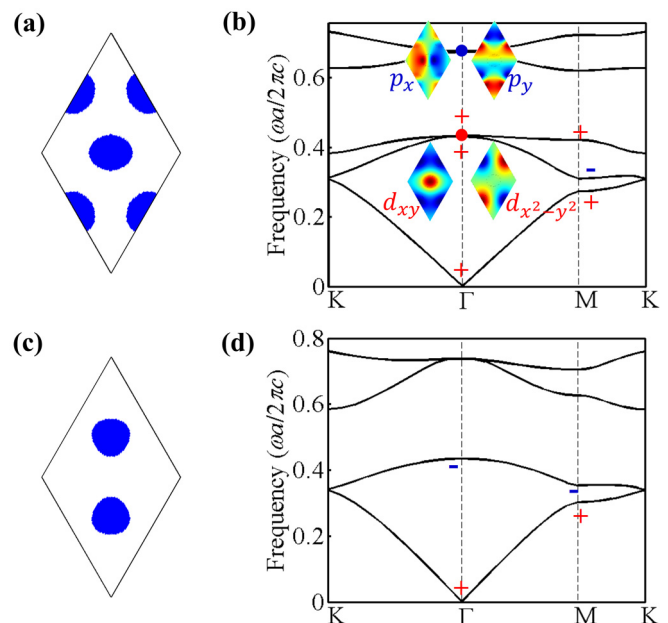


FIG. 1. Optimized primitive unit cell for topological (a) non-trivial PC and (c) trivial PC. Parts (b) and (d) are band diagrams for topological nontrivial PC and trivial PC, respectively. The insets are electric-field profiles of dipolar and quadrupolar modes at the Γ point.

SPTIs in Refs. [28,29,31]. Such a wide overlapped band gap enables highly localized edge and corner states, and is a key to providing the programming room for potential applications.

Here, we use the bulk polarization to confirm the topological properties of the optimized PCs. Because of the C_{6v} symmetry, the bulk polarization $\mathbf{P} = (P_x, P_y)$ can be calculated by the parities at high symmetric points through [25,50–52]

$$P_x = P_y = \frac{1}{2} \left(\sum_n q_n \bmod 2 \right), \quad (-1)^{q_n} = \frac{\eta_n(M)}{\eta_n(\Gamma)}, \quad (2)$$

where $i = x, y$ stands for the direction; η_n denotes the parity at the high-symmetry point for the n th band, which means even- or odd-symmetry behavior under inversion at the C_{6v} point group center. The p modes have an odd parity ($-$), while the d and s modes have an even parity ($+$). The parities at high symmetric points are labeled in the band diagrams. Derived from Eq. (2), the bulk polarization for NP1 equals $(1/2, 1/2)$, which ascertains that NP1 is in a topological nontrivial phase. The bulk polarization for TP is $(0, 0)$, which confirms that TP is in a topological trivial phase.

It is noticed that the design of NP1 is similar to the perturbed kagome lattice for the quantum spin Hall effect-based photonic topological insulator [16]. The design of TP is the same as that of the honeycomb lattice with the Dirac cone at K , which is used as a precursor to create the quantum valley Hall effect-based photonic topological insulator [53]. However, the topological mechanism exploited here is based on topological polarization and thus is fundamentally different from the Chern-related quantum valley/spin Hall effect [54].

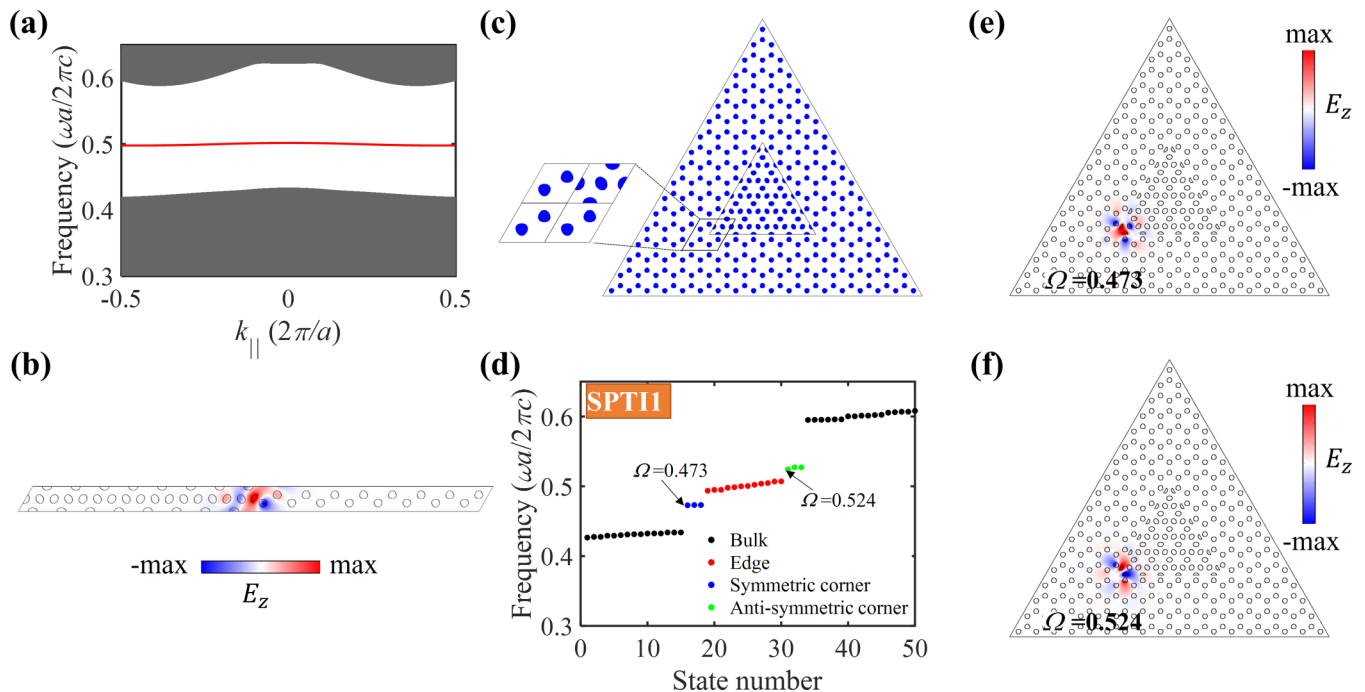


FIG. 2. (a) Band diagram for a supercell with a domain wall between 8 NP1s and 8 TP1s as shown in (b); (b) distribution of the electric field for a 1D edge state at Γ ; (c) metastructure constructed by NP1s and TP1s for the calculation of corner states; (d) eigenfrequencies of the metastructure in (c); (e) distribution of the electric field for a symmetric corner state; and (f) distribution of the electric field for an antisymmetric corner state.

III. TOPOLOGICAL EDGE AND CORNER STATES

According to the bulk-boundary correspondence principle [55,56], topological edge states emerge at the interface between trivial and nontrivial PCs. To demonstrate the topological edge states, a supercell composed of eight unit cells of NP1 and eight unit cells of TP is constructed with a domain wall as indicated in Fig. 2(b). The numerical simulation is performed in COMSOL MULTIPHYSICS. Figure 2(a) presents the projected band diagram. It can be seen that one edge state with almost flat dispersion occurs inside the overlapped band gap of NP1 and TP. The mode profile at Γ [Fig. 2(b)] reveals that the electric field is highly localized at the interface and is decayed exponentially into the bulk. The group index for this edge state is larger than 60, indicating the slow light propagation, which has many potential applications in nonlinear optics [57]. However, such a gapped 1D edge state is fundamentally different from the gapless 1D edge state in topological photonic crystals based on quantum spin Hall effects (QSHEs) [58]. The gapped edge state is also a prerequisite for the realization of a 0D corner state [28,29]. To confirm the existence of a 0D corner state, we construct a triangle metastructure, as shown in Fig. 2(c), with NP1 at the center and TP at the surrounding. Figure 2(d) displays the eigenfrequencies of the metastructure. Different from the existing SPTIs [26–29] with symmetric corner states only, we observe three near-degenerate symmetric corner states (blue dots) and three antisymmetric corner states (green dots). The diagonal mirror plane is the angular bisector of the corner. These in-gap symmetric and antisymmetric corner states emerge at the frequencies between the 1D edge states (red dots) and the bulk bands (black dots). Figures 2(e) and 2(f)

show the electric-field distribution for the symmetric corner state and the antisymmetric corner state, respectively. They clearly show that the electric field is highly localized symmetrically or antisymmetrically at the corner. The symmetric corner mode supports a high- Q factor with 1.3×10^6 , which significantly exceeds that in Ref. [26]. Note that the finite Q -factor corresponds to the in-plane leakiness of the mode, calculated here by employing scattering boundary conditions for the two-dimensional simulations.

IV. TUNING FREQUENCIES OF THE EDGE AND CORNER STATES

By specifying different frequencies of quadrupolar modes in topology optimization, various optimized nontrivial PCs can be obtained. Figure 3(a) shows four different nontrivial PCs by specifying the normalized frequency of quadrupolar modes at 0.440, 0.450, 0.470, and 0.485, respectively. These optimized nontrivial PCs are denoted as NP2, NP3, NP4, and NP5, respectively. Thus, five SPTIs (SPTI1, SPTI2, SPTI3, SPTI4, and SPTI5) can be created by assembling these optimized nontrivial PCs with a TP. Figure 3(b) shows superimposed projected band diagrams for the supercells of SPTI1, SPTI3, SPTI4, and SPTI5, demonstrating that the edge states of these SPTIs occur at different frequencies. These edge states occur at the overlapped band gap between $\Omega = 0.485$ and 0.588. Figures 3(c) and 3(d) show the eigenfrequencies of the metastructures, which are similar to the one shown in Fig. 2(c) except for replacing NP1 with NP2 and NP3, respectively. It can be seen from Figs. 2(d), 3(c), and 3(d)

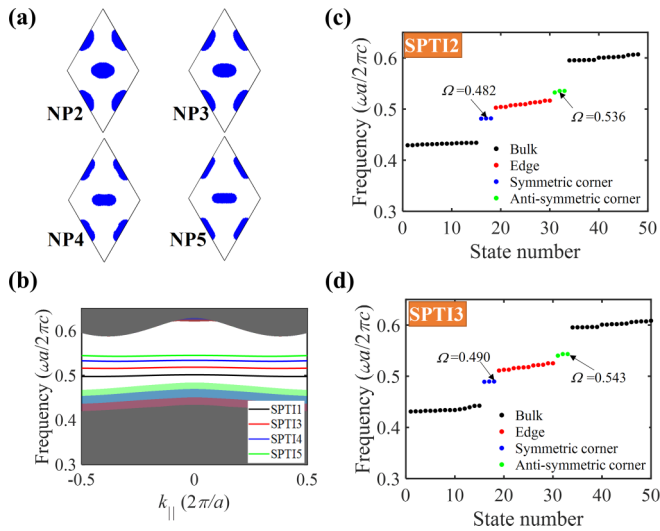


FIG. 3. (a) Optimized nontrivial PCs by specifying the normalized frequency of quadrupolar modes at 0.440, 0.450, 0.470, and 0.485, respectively; (b) superimposed projected band diagrams for SPTI1, SPTI3, SPTI4, and SPTI5; (c) eigenfrequencies of the metastructure shown in Fig. 2(c) by replacing NP1 with NP2; and (d) eigenfrequencies of the metastructure shown in Fig. 2(c) by replacing NP1 with NP3.

that the SPTI1, SPTI2, and SPTI3 metastructures have corner states at different frequencies.

V. ROUTING TOPOLOGICAL EDGE AND CORNER STATES

Through the analysis of topological edge and corner states in the previous section, routing topological edge and corner states with different frequencies can be achieved by purposely programming these optimized NPs and TPs. As an example, Fig. 4(a) shows the programming diagram of a photonic device composed of NP1, NP3, NP4, NP5, and two pieces of TPs with six channels (boundaries), which aims at manipulating waves with normalized frequencies ranging from 0.485 to 0.588. Since the supercell of two nontrivial PCs exhibits the band gap without any edge states (see the Supplemental Material [59], Sec. 2), two channels between NP1 and NP5, and NP3 and NP4, are blind. The other four channels between nontrivial and trivial PCs, as highlighted by the colored lines, will be utilized for the propagation of light waves due to the existence of edge states. As shown in Fig. 3(b), these four channels support edge states at different frequencies, which do not interfere with each other. As a result, light waves with different frequencies emitting from a line source at the center of the device (denoted by the red asterisk) may propagate along different channels. Figures 4(b)–4(e) show the distributions of the electric field when the source has the normalized frequency 0.502, 0.517, 0.536, and 0.546, respectively. It can be seen that topological edge states with different frequencies are routed into different channels. Figure 4(f) presents the spectrum of normalized transmission for the ports at the end of each channel. It can be seen that, when the

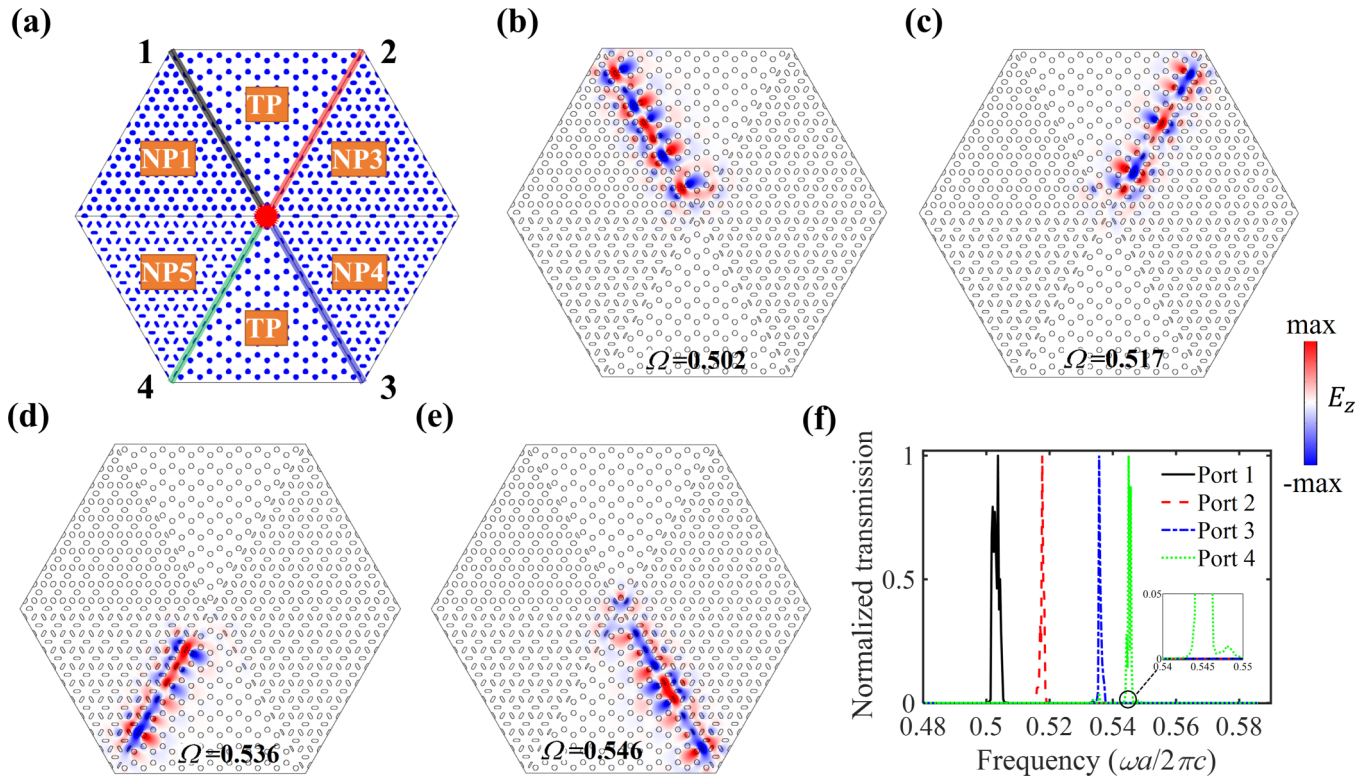


FIG. 4. Routing topological edge states with different frequencies. (a) Schematic of the photonic device for routing topological edge states. Color lines mark four channels formed between nontrivial PCs and trivial PCs. (b), (c), (d), and (e) Simulated electric-field distribution for the point source with the normalized frequency being 0.502, 0.517, 0.536, and 0.546, respectively. (f) Spectrum of normalized electric field for the ports at the end of each channel.

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