## Single-Charge Transistor Based on the Charge-Phase Duality of a Superconducting Nanowire Circuit

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We propose a transistorlike circuit including two serially connected segments of a narrow superconducting nanowire joint by a wider segment with a capacitively coupled gate in between. This circuit is made of amorphous NbSi film and embedded in a network of on-chip Cr microresistors ensuring a sufficiently high external electromagnetic impedance. Assuming a virtual regime of quantum phase slips (QPS) in two narrow segments of the wire, leading to quantum interference of voltages on these segments, this circuit is dual to the dc SQUID. Our samples demonstrated appreciable Coulomb blockade voltage (analog of critical current of the SQUIDs) and periodic modulation of this blockade by an electrostatic gate (analog of flux modulation in the SQUIDs). The model of this QPS transistor is discussed.

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The discovery of the Josephson effect [1] triggered the development of superconductive electronic devices operating on the principle of the classical behavior of the collective quantum variable, i.e., the superconducting phase difference. These include rf and dc SQUIDs [2], single-flux-quantum logic circuits [3], Josephson voltage standard arrays [4], microwave receivers [5], etc. The operation of these devices is based on the nonlinear ( $2\pi$  periodic) dependence of the supercurrent  $I = I_c \sin\varphi$  on the phase difference  $\varphi$  on the Josephson junction (JJ), where  $I_c$  is the critical current. The electromagnetic impedance of the JJ for a small harmonic signal is therefore expressed in terms of Josephson inductance  $L_J$  which varies with  $\varphi$  also in a periodic fashion,  $L_J^{-1}(\varphi) = (2\pi I_c/\Phi_0)\cos\varphi$ , where  $\Phi_0 = h/2e$  is the flux quantum.

Recently, a new type of nonlinear superconducting element, dual to the JJ-the quantum phase slip (QPS) nanowire (see, for example, the review on the physical properties of superconducting nanowires Ref. [6], and references therein)—was proposed in Refs. [7,8]. Being included in a superconducting ring with flux bias near degeneracy points,  $\Phi_b = (m + 0.5)\Phi_0$ , m is an integer number, the QPS element enables coupling of quantum states  $|m\rangle$  and  $|m+1\rangle$ . This leads to anticrossing of corresponding levels with the gap  $\Delta E = E_{\text{OPS}}$ . For a uniform nanowire of length l, the characteristic coupling energy  $E_{\text{OPS}} = l\epsilon_S$  with  $\epsilon_S \approx (k_B T_c / \xi) (R_O / R' \xi)^{1/2} / k_C / \xi$  $\exp(0.3\alpha R_O/R'\xi)$ , where  $T_c$  is the critical temperature,  $\xi$ the coherence length, R' the normal-state resistance per unit length,  $R_{\Omega} = h/4e^2 \approx 6.5 \text{ k}\Omega$ , the resistance quantum, and  $\alpha \sim 1$  [6,7]. The two-level spectrum of this circuit was proposed [7] for use in a flux qubit. The experimental observation of such a two-level system in the superconducting loop with a nanowire segment made of highly disordered  $InO_x$  has been recently reported by Astafiev *et al.* [9].

Coupling of the nanowire to a high-Ohmic environment through the appropriate boundary conditions imposes dramatic changes in its dynamics [10]. Based on the quantum duality of the Josephson and QPS effects stemming from conjugation of the phase and charge variables, Mooij and Nazarov had further predicted [8] that in the charge-bias regime, the QPS element should demonstrate the Coulomb blockade behavior with periodic dependence of voltage on the injected charge q, viz.,  $V_{\text{QPS}}(q) = V_c \sin(\pi q/e)$  with  $V_c = 2\pi E_{\text{OPS}}/2e$ . Thus, this element should behave as a nonlinear capacitor,  $C_{\text{OPS}}^{-1}(q) = dV_{\text{QPS}}/dq = (\pi V_c/e) \times$  $\cos(\pi q/e)$  with a 2*e*-periodic dependence on *q*. It is clear that such a QPS element could be exploited in the engineering of electronic devices controlled by the charge in the classical manner. These devices are dual to the Josephson ones controlled by the classically behaved phase. For example, applying microwave irradiation of frequency f should lead to the formation of the current steps in the *IV* curve at I = 2nef (*n* is integer), enabling a fundamental standard of current [8]. These steps are dual to Shapiro voltage steps in a JJ IV curve at  $V = n\Phi_0 f$  [11]. Developing such an analogy, in this Letter we proposed and realized a QPS-based single-charge transistor which is dual to dc SQUID and can be operated as an electrometer.

Our device [see schematic diagram in Fig. 1(a)] has a symmetric inline configuration and includes two narrow pieces of superconducting nanowire joined by an island (a wider middle part of the same nanowire) with a capacitively coupled gate. This transistor is embedded in the network of compact high-Ohmic resistors,  $R \sim 0.4 \text{ M}\Omega \gg R_Q$ , which ensure a sufficiently high electromagnetic impedance seen by the transistor. Because of this improvement, a classical charge regime of operation with significant damping of dynamics is realized. (Note that the recent proposal by Hriscu and Nazarov [12] deals with a complementary QPS transistor, without resistors, operating

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FIG. 1 (color online). (a) The layout with three fragments of SEM image and (b) the simplified electric circuit diagram of the QPS transistor embedded, in fact, in the four-terminal network of on-chip resistors. The device includes two QPS elements denoted by diamond symbols, kinetic inductances of the nanowire segments, and a capacitive gate. Thicknesses of NbSi and Cr films are 10 and 30 nm, respectively.

in a quantum regime, which, in contrast to our circuit is phase biased and, therefore, not strictly dual to that of the dc SQUID.)

The Kirchhoff equation for the equivalent electric circuit presented in Fig. 1(b) has the form

$$V_1 + V_2 = V_b,$$
  
 $V_i = L_{ki}\ddot{q}_i + R_i\dot{q}_i + V_{\text{OPS}i}(q_i), \qquad i = 1, 2.$  (1)

Here,  $V_{\text{QPS}i}(q_i) = V_{ci} \sin(\pi q_i/e)$  are the 2*e*-periodic voltages on the QPS elements on either side of the island. The charge conservation relation takes the form

$$q_1 - q_2 = -C_g V_g - (C_g + C_0)(V_1 - V_2).$$
(2)

This equation reflects the balance of the charges injected in the island and the polarization charges induced by the gate and floating potential of the island having a selfcapacitance  $C_0$  with respect to ground. In the case of sufficiently small gate- and self-capacitances,

$$C_g, C_0 \ll e/V_{c1,c2},$$
 (3)

and small injected currents (viz.,  $\pm \dot{q}_{1,2}$ ), the difference charge  $q_1 - q_2 = -C_g V_g \equiv -Q_g$  is totally controlled by the gate voltage  $V_g$ .

One can easily find that Eqs. (1)–(3) are dual to those of the dc SQUID; see, for example, Ref. [13]. The charges  $q_1$ and  $q_2$  are dual to the phases across individual JJs, whereas amplitudes  $V_{c1,c2}$  play the roles of the critical currents. To continue this analogy we can put in correspondence the pairs of parameters  $L_{k1,k2}$  and JJ capacitances,  $R_{1,2}$  and shunting resistors, etc., for the QPS transistor and dc SQUID, respectively. The condition of small gate- and self-capacitances, Eq. (3), is dual to the requirement of a small inductance of the SQUID loop ( $L \ll L_J$ ). Similar to external flux control of quantum interference in the SQUID [13], the voltage on the QPS transistor results from quantum interference of voltages on individual QPS elements and is controlled by the gate charge  $Q_g$ ,

$$V = V_{\text{QPS1}} + V_{\text{QPS2}} = V_m(Q_g) \sin(\pi Q/e),$$
 (4)

where  $Q = q_1 + e\eta/\pi = q_2 + e\eta/\pi - Q_g$  is the "average" charge,

$$\tan \eta = \frac{2V_{c2}\tan(\pi Q_g/2e)}{V_{c+} + V_{c-}\tan^2(\pi Q_g/2e)},$$
 (5)

 $V_{c\pm} = V_{c1} \pm V_{c2}$ , and the Coulomb blockade voltage

$$V_m^2 = V_{c1}^2 + V_{c2}^2 + 2V_{c1}V_{c2}\cos(\pi Q_g/e).$$
 (6)

In the case of a symmetrical circuit,  $V_{c1} = V_{c2} = V_c$ ,  $Q = q_{1,2} \pm Q_g/2$ , Eq. (6) yields the maximum modulation of the blockade voltage,  $V_m = 2V_c |\cos(\pi Q_g/2e)|$ . In the case of a highly asymmetrical circuit, say  $V_{c2}/V_{c1} = a \ll 1$ , charge  $Q \approx q_1$  and the blockade voltage is only slightly modulated by the gate charge  $Q_g$ ,

$$V_m = V_{c1} [1 + a \cos(\pi Q_g/e)].$$
(7)

Thus the equation of motion Eq. (1) takes a form of the resistively shunted junction (RSJ) model for dc SQUID,

$$L_k \ddot{Q} + R \dot{Q} + V_m (Q_g) \sin(\pi Q/e) = V_b, \qquad (8)$$

with  $L_k = L_{k1} + L_{k2}$  and  $R = R_2 + R_2$ .

Equation (8) describes the dynamics of a nonlinear oscillator with finite damping. It yields a dc IV curve with both the static and running charge (the oscillating regime,  $\omega = \pi \langle \dot{Q} \rangle / e = \pi \langle I \rangle / e$ ) branches, dependent on  $Q_g$  in a periodic fashion. The dimensionless parameter, crucial for the dynamics, introduced earlier in Ref. [8],

$$\beta_{\rm OPS} = \omega_c L_k / R, \tag{9}$$

where  $\omega_c = \pi V_m(Q_g)/eR$  is a characteristic circular frequency. Eventually,  $\beta_{\text{QPS}}$  is the analog of the Stewart-McCumber parameter in the Josephson dynamics. In the most realistic case of large damping,  $\beta_{\text{QPS}} < 1$ , the *IV* curve has a shape with characteristic back bending,

$$\langle V \rangle = [V_m^2(Q_g) + R^2 \langle I \rangle^2]^{1/2} - R \langle I \rangle, \qquad (10)$$

where  $\langle V \rangle$  is the average voltage on the nanowire.

The fabrication method (some modification of the method developed in Ref. [14]) included the following steps. The samples were fabricated on a silicon substrate, having 300 nm thick thermal oxide layer, combining two processes: shadow evaporation and the sputter deposition with etching step. The Cr resistors were fabricated in a single vacuum cycle with the AuPd contact wires and micropads utilizing the shadow evaporation technique

through a bilayer polymethyl methacrylate (PMMA) and copolymer stencil mask. The 30 nm thick Cr resistors were evaporated first at a low residual pressure of oxygen ( $\sim 10^{-6}$  mbar) followed by a 50 nm thick layer of AuPd from an angle for which the narrow stencil openings for Cr resistors were overshadowed by the mask. Using this trick, the formation of AuPd shadows parallel to the Cr resistors was avoided.

The Cr resistors and other parts of the circuit were protected by a PMMA mask while 10 nm thick amorphous  $Nb_x Si_{1-x}$  film (x = 0.45, defined by calibrating the sputter rate for each element at a given sputter power and periodically confirmed by energy-dispersive x-ray (EDX) measurement using separately deposited films on Ge substrates) was cosputtered on the substrate making contact only with the AuPd micropads. These micropads allowed making an electrical contact of the NbSi film with the Cr resistor circuitry. To remove organic residue and water for the purpose of forming reliable contacts, we cleaned the micropads in the reactive ion etching system with oxygen plasma and baked the sample overnight in  $N_{\rm 2}$  atmosphere at a temperature of 120 °C before NbSi deposition. The substrate was rotated 20 times during a one-minute-long deposition. After the liftoff, the wafer was coated with inorganic negative tone hydrogen silsesquioxane (HSQ) resist (XR-1541, Dow Corning) patterned with an electron beam. After exposure, the HSQ resist had characteristics of thermal silicon oxide and could be used as an etch mask to define the NbSi nanowires along with the island and the gate structure made out of the same NbSi film. In a final step, the inductively coupled plasma etching process with SF<sub>6</sub> gas was used for etching.

The samples were measured in a dilution refrigerator with electrical lines equipped with microwave frequency filters made of pieces of Thermocoax<sup>TM</sup> cable. We used battery power sources and homemade electronics for either a voltage or a current bias of our circuits. Most of the measurements were performed at the lowest temperature of the fridge, T = 15 mK. The layout allowed independent characterization of both pairs of on-chip resistors. Their IV curves at millikelvin temperatures were practically linear with resistance about 20% above the values measured at room temperature. The NbSi films had the superconducting transition temperature  $T_c \approx 1$  K and the normal-state resistance per square about 550  $\Omega$  (approaching the level of ca. 600  $\Omega$  of the thickness-driven superconductor-to-insulator transition in somewhat thicker films of Nb<sub>x</sub>Si<sub>1-x</sub> with stoichiometry x = 0.15 [15]), yielding for the narrow segments (w = 18 nm) the specific resistance  $R' \approx 31 \text{ k}\Omega/\mu\text{m}$ . The NbSi film parameters were measured at T = 3.4 K on the stand-alone test wires 20 nm in width, fabricated on the same chips.

Here we report on two samples (A and B, see their characteristics in Figs. 2 and 3, respectively). In comparison with sample B whose dimensions are shown in



FIG. 2 (color online). The charge-modulated *IV* curves of sample *A* recorded in a current bias regime for two gate voltages shifted by a half-period. In the region of small currents (blown up in inset) one can see the modulation with period  $\Delta I = 13.5$  pA, which is due to the asymmetry of off-chip biasing circuitry resulting in the current dependence of the electric potential of the transistor island,  $\delta V = (R_{\text{biasl}} - R_{\text{bias}2})I$  and, therefore, of the effective gate charge,  $\delta Q_g = (C_g + C_0)\delta V$  [cf. Eq. (2)]. The green dashed line shows the shape of the bare *IV* curve given by the RSJ model Eq. (10) with fixed  $Q_g$ .

Fig. 1(a), sample A had a twice longer island (25  $\mu$ m) and somewhat shorter QPS sections (4.5  $\mu$ m each). The layout of the gate was identical in both samples. Both samples showed an appreciable Coulomb blockade,  $\langle I \rangle = 0$ , and a gate effect at small bias. An increase in current  $\langle I \rangle \gtrsim 1$  nA was accompanied by a steep rise in voltage  $\langle V \rangle$  indicating a gradual turning of the wire into the normal state. On a large scale, the *IV* curves of



FIG. 3 (color online). The *IV* curves of sample *B* measured in the voltage bias regime at different values of gate voltage  $V_g$ . The bottom right inset shows details of the Coulomb blockade corner. Upper left inset: the gate voltage dependence of the transistor current measured at different bias voltages  $V_b$  providing a steady increase of  $\langle V \rangle$  from 0.321 up to 0.481 mV in 20  $\mu$ V steps (from bottom to top).

both samples exhibited a positive excess current, i.e., had a shifted linear asymptote,  $I_{off} = \langle I \rangle - \langle V \rangle \times (d\langle I \rangle / d\langle V \rangle)_{\langle V \rangle \to \infty} \approx 1.5$ –2.0 nA, indicating that part of the current at large bias was still transported by a superconducting component [16]. This shape of the *IV* curve argues that the observed gate effect has a superconductive origin and allows us to rule out the strikingly different effect of the Coulomb blockade in a normal single electron transistor, characterized by a pronounced voltage offset [17] or, equivalently, a negative "excess" current. Moreover, the measured *IV* curves never showed characteristic gap-voltage knees typical for Josephson tunnel junctions.

On a small scale, one can see the clear gate modulation of IV curves, although with very different periods,  $\Delta V_g \approx 2$  and 150 mV, for samples A and B, respectively. Assuming the 2e-periodic charge dependence in sample A (see inset to Fig. 2) capacitance  $C_g = 2e/\Delta V_g$  takes the value of about 160 aF. This value is not far from the modeled value of 200 aF. The depth of the modulation of about 40% with practically harmonic shape suggestsaccording to our model Eq. (7)-the ratio of QPS energies,  $a = V_{c2}/V_{c1} = E_{\text{QPS2}}/E_{\text{QPS1}} \approx 0.2$ , whereas the average value is about 20  $\mu$ eV. Inserting this value in the formula for the QPS energy and assuming the length of the QPS junction of the order of the coherence length  $\xi$ , we obtain the ratio  $R_Q/R'\xi \approx 9$  and, therefore, a rough estimate  $\xi \approx 20$  nm. This value is close to the values  $\xi = 10-15$  nm found in Ref. [14].

The shape of a bare (not modulated) IV curve is only qualitatively similar to that given by Eq. (10) and shown by the dashed line. A back bending weaker than in theory can be attributed to the effect of stray capacitance of the resistors, resulting in a roll-off of effective impedance R in Eq. (10) with the rise of current ( $\propto \omega$ ). Rounding of the Coulomb blockade corner is attributed to the effect of noise, omitted in our model.

Sample *B* demonstrated an even stronger gate effect (see upper left inset of Fig. 3), but also the peculiar properties. A huge period  $\Delta V_g \approx 150 \text{ mV}$  of the gate voltage dependence corresponds to coupling capacitance  $C_{g} \approx 2$  aF. This value is 2 orders smaller than the designed value, so we conclude that the actual island in this sample had much smaller dimensions (very crudely, on the order of 100 nm) and located presumably in one of the narrower segments of the nanowire. The available single gate did not allow any conclusions about exact size and location of this island, whereas the depth of modulation of about 20% yielded the ratio of QPS energies,  $E_{\text{QPS2}}/E_{\text{QPS1}} \approx 0.1$  with  $E_{\text{QPS1}} \approx$ 150  $\mu$ eV. A small hysteresis and almost vanishing back bent part in the IV curve, seen in the lower bottom inset of Fig. 3, may indicate that the value of the damping parameter  $\beta_{\text{QPS}}$  Eq. (9) is between 1 and 2 [13], whereas our estimation yields a value of  $\beta_{OPS}$  smaller by an order of magnitude, ensuring a heavily overdamped regime.

Interestingly, the observed periodic pattern in sample Bwas superimposed on another one having a substantially larger period and a much weaker modulation, discernible in the traces at smaller bias in the upper left inset of Fig. 3. This behavior can be due to the emergence of an additional (very small) island neighboring inline with the actual island. Thus, in spite of a rather high homogeneity in thickness and width, the entire nanowire can be considered as a circuit including several weaker sections with a local increase in QPS energies per unit length,  $\epsilon_{\rm S}(x)$ . In the spirit of the single-charge interferometer operation, Eqs. (4)-(6), the contributions of these weak sections to the total QPS energy should be summed up, taking into account phases proportional to the charges induced by the gate on all intermediate islands. The resulting quasiperiodic dependencies can be also interpreted in terms of the Aharonov-Casher effect demonstrated in the experiments with phase-biased arrays of small JJs (i.e., in a lumped-element analog of superconducting nanowire) [18,19]. Recent calculations by Vanević and Nazarov [20] support the hypothesis that apparently homogeneous nanowires may naturally have a strongly inhomogeneous distribution of specific QPS energy  $\epsilon_{S}(x)$  because of the exponential dependence on the local parameters. Large spatial fluctuations of the local energy gap detected by scanning tunneling methods were also reported for thin disordered films of TiN [21], NbN [22], and InO [23].

In conclusion, we have demonstrated the single-charge effect in superconducting nanowires having a transistor configuration with a capacitively coupled gate, embedded in a high-impedance environment. A deeper understanding and better control of the nanowire parameters determining the characteristics of these transistors and other possible circuits is urgently needed and motivates us to conduct further research. Generally, the demonstrated duality of the QPS transistor and the dc SQUID may open the way towards interesting applications of QPS nanocircuits in electronics and metrology.

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