# Probing charge- and heat-current noise by frequency-dependent fluctuations in temperature and potential

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The energetic properties of electron transport in mesoscopic and nanoscale conductors are of considerable interest at present. Here, we theoretically investigate the possibility of probing charge- and heat-current fluctuations as well as their mixed correlations via the temperature and electrochemical potential fluctuations of a probe coupled to the conductor. Our particular interest is devoted to the charge and energy noise stemming from time-dependently-driven nanoelectronic systems designed for the controlled emission of single electrons, even though our setup is appropriate for more general ac-driving schemes. We employ a Boltzmann-Langevin approach in order to relate the bare charge- and energy-current fluctuations emitted from the electron source to frequency-dependent electrochemical potential and temperature fluctuations, which the former induce in the probe. We apply our findings to the prominent example of an on-demand single-electron source, realized by a driven mesoscopic capacitor in the quantum Hall regime. We show that neither the background fluctuations of the probe in the absence of the working source, nor the fluctuations induced by the probe hinder the access to the sought-after direct source noise for a large range of parameters.

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# I. INTRODUCTION

Charge transport in mesoscopic and nanoscale structures can nowadays be manipulated at the single-electron level. Of particular interest for different fields, ranging from metrology to quantum information and quantum optics with electrons, are dynamically driven singleelectron sources injecting particles into a conductor *on demand* [1–7]. A way to access the precision of these sources, which is not accessible by the average charge current alone, is by detecting their charge-current statistics. Indeed, already the low-frequency noise (the second moment of the statistics) is known to provide additional information about the number of particles that take part in transport, thereby giving access to electron-hole pair creation detrimental for the source precision [6,8,9].

Not only do the transported charged particles carry energy, but so do the neutral electron-hole pair excitations in general [10,11]. This fact has, in recent years, motivated theoretical investigations on the fluctuations of charge currents, heat currents, and also their mixed correlations for driven systems [12–15], see also Refs. [16–18] for the stationary bias case. The energetic properties revealed in

this way are of key importance: single-electron sources, while injecting single particles into a conductor in a wellcontrolled manner, can also be used for quantum-state preparation of electronic flying qubits. Therefore, in addition to the precision, knowledge about the energy spectrum of the injected signal is required.

A pressing question is hence how charge and energy currents, as well as their fluctuations, can actually be accessed experimentally. Recently, energy-resolved charge currents of single-electron sources have been measured with the help of an energy-dependent detection barrier [7,19]. Even a highly complex state tomography [20,21] has been performed to access the quantum state of the single particles [22]. In order to detect fluctuations in the heat current, the detection of power fluctuations has been suggested [13], which have a straightforward interpretation when the source is voltage-bias driven.

In this paper, we analyze the possibility of extracting charge-current and heat-current noise—injected from an arbitrary ac electronic current source—as well as their mixed correlations using a floating probe [23,24]. The incoming fluctuating charge and heat currents modify the probe's electronic energy distribution, which can experimentally be readout via effective temperature and electrochemical potential fluctuations [25–27]. We emphasize

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that, for the detection of electrochemical potentials and their fluctuations in a system fed by a steady-state charge current, in general, conductors with three or more terminals are required [24]. Instead, for the analysis of the chargeand heat-current fluctuations of a *pure ac current source*, where the time-averaged injected charge current is always zero, the simple setting suggested here suffices. Probes of similar type have been used for stationary heat-current detection [28]. Moreover, fast temperature measurement techniques are nowadays available [29] and motivate our theoretical proposal.

Using a Boltzmann-Langevin approach [30], we directly relate the low-frequency fluctuations of charge and energy currents flowing into the probe to the finite-frequency probe temperature and potential fluctuations that these induce. The frequency dependence of the fluctuations sets a fundamental constraint on the bandwidth within which the fluctuations can be detected. In order to furthermore extract the *bare-source* fluctuations, which we are interested in, from the total noise detected with the help of the probe, one has to proceed as follows: First, subtract the *pure-probe* fluctuations (obtained when the source is not operating) and, second, disentangle the remaining data from the fluctuations *induced* by the temperature and potential measurements.

As a proof of principle, we apply our findings to a case where a time-dependently-driven mesoscopic capacitor [1] in the quantum Hall regime is taking the role of the current injector. The noise [30] stemming from the source and serving as the bare fluctuations entering the Boltzmann-Langevin approach is derived for this explicit example by employing (Floquet) scattering theory [31]. We can thereby demonstrate the possibility of determining the source fluctuations from the potential and temperature fluctuations of the probe.

Finally, we want to point out that the relation between current noise and detectable dissipated energy is much more relevant. The relation between higher moments of charge-current statistics and power fluctuations [25,32] has, for example, been exploited in the detection of photons emitted from the noisy current through a tunneling barrier [33–35]. Vice versa, heat currents are presently detected via noise measurements [36]

This paper is organized as follows: In Sec. II we introduce the setup, consisting of an injector to be analyzed and the detector being made of a probe contact and a quantum point contact (QPC). The theoretical Boltzmann-Langevin approach is introduced in Sec. III and employed to the calculation of the probe fluctuations in Sec. IV A. We then, in Sec. IV B, separate the fluctuations into pureprobe fluctuations and fluctuations arising when the source is switched on, including the sought-after bare-source fluctuations. In Sec. V, we analyze the specific case of a time-dependently-driven mesoscopic capacitor playing the role of the injector.

## **II. MODEL**

In this section, we introduce the model setup for the generation and, in particular, for the detection of fluctuating charge and energy currents and their correlations. Next, we discuss the relevant timescales of the problem and the fundamental requirements for an experimental realization of our proposal.

#### A. Description of the setup

A sketch of the mesoscopic noise detection system, implemented in a ballistic, two-terminal conductor, is presented in Fig. 1. The system comprises two key components referred to as an *injector* and a *detector*. The injector part consists of an electronic reservoir (denoted by L), electrically grounded ( $\mu_L = 0$ ) and kept at a fixed temperature  $T_L$ , and a scattering region (denoted by S) representing an ac current source-a generic device injecting electrons (and/or holes) into the conductor. Such a current source can be analyzed via its current and noise properties. Our goal is to investigate how the latter can be accessed by measuring fluctuations of the electrochemical potential  $\mu_p(t)$  and temperature  $T_p(t)$  in the probe (denoted by p) belonging to the detector part of the setup. We note that the probe scheme under consideration is applicable for any rapidly fluctuating ac current source. In fact, the only important assumptions required at this point are: (i) the lack of backreflection between the source-scattering region and a quantum point contact of the detector region, which could otherwise lead to the creation of resonant states at the injector-detector interface [37]; (ii) the charge- and



FIG. 1. Schematic of the ballistic, two-terminal noise detection system, with the injector and detector parts indicated. The injector part consists of an electronic reservoir, *L*, kept at an electrochemical potential  $\mu_L$  and temperature  $T_L$ , and a timedependently-driven forward-scattering potential, *S* (red). On the detector side, the electrons emitted from the injector are scattered at a quantum-point contact (QPC) with transparency *D*. The transmitted electrons are absorbed at the electrically and thermally floating probe, *p*, inducing fluctuations of the electrochemical potential  $\mu_p(t)$  and temperature  $T_p(t)$ .

energy-current source is characterized by some timescale T associated with the rate of charge (and energy) emission.

An illustrative example of the ac current source that we consider in this paper is a time-dependently-driven mesoscopic capacitor injecting on-demand single particles into the conductor. The operation of such a source, experimentally demonstrated by Fève *et al.* [1], can be theoretically described by a Floquet scattering matrix approach [38]—for details see Sec. V. Importantly, in the case of such a periodically driven current source, the natural characteristic timescale associated with the source is the driving period T, related to the driving frequency as  $\Omega = 2\pi/T$ . For the energetic properties of the source, the width in time of the injected current pulse,  $\sigma$ , will also turn out to be important, see Sec. V.

On the other hand, the *detector* part of the setup contains a QPC with a single active transport mode, and a thermally floating reservoir acting as a probe, indicated by p in Fig. 1. The electrical conductance of the QPC is gD, where D denotes the tunable, energy-independent transparency and  $g = e^2/h$  is the conductance quantum per spin, with the electron charge  $-e = -1.602 \times 10^{-19}$  C. Having in mind an injector system that operates in the fully spin-polarized quantum Hall regime, we discard here the electronic spin degree of freedom. At the QPC, the fluctuations of the injector currents are modified due to elastic scattering. This modification, depending on D, is used to tune the different components of the probe noise, as further discussed in Sec. VC. The core element of the detector is, however, the *floating probe*. We assume it to be metallic and large enough to avoid Coulomb-blockade effects. The probe works in the hot-electron regime, meaning that particles entering the probe thermalize quickly via electron-electron scattering, see, e.g., Refs. [28,29,36,39]. For this reason, a well-defined electrochemical potential  $\mu_p(t)$  and electronic temperature  $T_p(t)$  can be associated with the probe at all times t. Moreover, we assume that the statistics of the energy dissipated in the probe are determined by the energy statistics of the signal entering the probe [40].

Importantly, in a realistic setup, energy can be evacuated from the probe not only via electrons, but it can also be transferred to the lattice phonons, which are in thermal equilibrium at a constant temperature  $T_0$ . For the electron-phonon coupling strength  $\Sigma V$ , with the probe volume V, the energy current flowing from the electron to the phonon subsystem is given by the general expression  $\Sigma V[T^n - T_0^n]$ , where T is a temperature that can later be replaced by the probe temperature. Depending on the material, probes may be characterized by different values of the (possibly noninteger) temperature exponents n in the electron-phonon coupling [28,36,41–44]. In the following, we neglect fluctuations of the energy current to the phonon bath. As a result, the total thermal conductance  $\kappa$ of the probe involves both the electronic and the phonon contribution,  $\kappa(T) = \kappa_e(T) + \kappa_{\rm ph}(T)$ ,

$$\kappa_e(T) = DgLT$$
 and  $\kappa_{ph}(T) = n\Sigma \mathcal{V}T^{n-1}$ , (1)

with the Lorenz number  $L = (\pi k_B/e)^2/3$ .

In addition, the readout of the probe temperature and electrochemical potential would also, in general, lead to energy losses or heating. Nonetheless, for the sake of simplicity, here we assume this readout to be *noninvasive*. The unavoidable energy changes due to the readout can, however, be included in our description on the same footing as the energy outflow to the phonon bath.

Finally, the dynamic properties of the probe are ruled by its charge and heat capacitances. Specifically, the charge capacitance *C* corresponds to the total (electrochemical) capacitance,  $C = (1/C_g + 1/C_q)^{-1}$ , containing both a geometrical,  $C_g$ , and a quantum component,  $C_q = e^2 v$ , with v standing for the probe density of states at the Fermi level. Moreover, the electronic heat capacity is temperature dependent,  $C_E(T) = v(\pi k_B)^2 T/3$ .

### **B.** Relevant timescales

Relations between the different characteristic timescales occurring in the problem are crucial for the understanding of the physics underlying the proposed detection scheme, and set the validity of the theoretical approach to be applied. Below, we introduce all relevant timescales, and discuss relations between these timescales which need to be satisfied, so that the readout of the temperature and electrochemical potential fluctuations in the probe are feasible.

To begin with, the first important timescale to be considered is the electron-electron interaction time  $\tau_{e-e}$ , associated with the internal thermalization of the probe electron gas. As mentioned above, a fast thermalization guarantees a Fermi distribution of the probe electrons, with a welldefined temperature  $T_p(t)$  at all times t. Since this is a crucial condition for the readout of the probe temperature, here we assume  $\tau_{e-e}$  to be of the order of the timescale at which electrons are emitted from the source or even shorter [28]. Electron-electron relaxation times are found to be of the order of 1–10 ns in metals [45] and 1 ns in semiconductors [46,47].

Next, we recall that the fluctuations in the electrical charge and energy current from the source occur on the timescale  $\mathcal{T}$ . When the fluctuating charge and energy currents flow into the probe, they give rise to fluctuations of the electrochemical potential  $\mu_p(t)$  and of the temperature  $T_p(t)$  in the probe. These electrochemical potential and temperature fluctuations occur on timescales set by the probe charge and energy relaxation times,  $\tau_{RC} = C/(gD)$  and  $\tau_E = C_E/\kappa$ , respectively. While single-electron sources are typically operated in the GHz regime [1,6], it is expected that metallic probes can be tuned to

exhibit much slower dynamics in the MHz regime. Modifications of this time can be reached by changing the probe capacitance via the area and/or the volume of the probe, or by tuning the transmission of the QPC.

This motivates the choice of our theoretical approach the Boltzmann-Langevin approach as described in Sec. III. It is applicable as long as the characteristic time of the current source,  $\mathcal{T}$ , is much smaller than the charge and energy relaxation times  $\tau_{RC}$  and  $\tau_E$  of the probe, implying that  $\mu_p(t)$  and  $T_p(t)$  change only by a small amount per drive cycle. Taken together, the required hierarchy of timescales can be established

$$\tau_{e-e} \lesssim \mathcal{T} \ll \tau_{RC}, \tau_E.$$
 (2)

Finally, in order to have detectable temperature fluctuations, it is important that the energy relaxation is not completely dominated by the coupling to the lattice phonons. To meet this condition,  $\kappa_e$  must be comparable to or larger than  $\kappa_{ph}$ . In terms of the relaxation time due to the electron-phonon scattering,  $\tau_{e-ph} = C_E/\kappa_{ph}$ , the optimal regime for detecting the temperature fluctuations is reached for  $\tau_{e-ph} \gg \tau_E$ . This condition basically demands that the energy relaxation due to phonons is sufficiently slow to enable detection of energy fluctuations. Indeed, at subkelvin temperatures, energy relaxation times due to electron-phonon coupling in metals reach tens of microseconds [29].

# **III. METHOD AND FORMALISM**

Our aim is to acquire knowledge about the average charge and energy currents,  $I_s(t)$  and  $I_s^E(t)$ , as well as their fluctuations generated by the source. This is achieved by investigating the average electrochemical potential  $\mu_p(t)$ and the temperature  $T_p(t)$  in the probe along with their fluctuations, which are, in turn, connected to the average charge and energy currents,  $I_p(t)$  and  $I_p^E(t)$ , as well as to their fluctuations *in the probe*. Furthermore, it turns out that it is actually the heat current in the probe,  $J_p(t) =$  $I_p^E(t) + \mu_p(t)I_p(t)/e$ , rather than the energy current,  $I_p^E(t)$ , that is naturally associated with the temperature fluctuations in the probe.

In general, the procedure allowing us to relate the quantities of interest mentioned above consists of three steps: First, we relate the electrochemical potential  $\mu_p(t)$  and the temperature  $T_p(t)$  to the charge and the energy stored in the probe and formulate the dynamical equations for the two latter quantities by resorting to general conservation laws. Next, we determine the averages  $\overline{\mu}_p$  and  $\overline{T}_p$  in terms of the average source currents,  $\overline{I}_s$  and  $\overline{I}_s^E$ . Finally, using the Boltzmann-Langevin approach, we establish a relation between the fluctuations of these currents and the much slower induced fluctuations of the *macroscopic probe* variables. The *bare*-source currents and their fluctuations, which in this section enter as independent variables, will later be evaluated for a specific system by employing a Floquet scattering theory, see Sec. V.

# A. General probe conservation equations

As a first step, we express the time-dependent charge Q(t) and energy E(t) in the probe in terms of  $\mu_p(t)$  and  $T_p(t)$ . Considering the probe potential with respect to the electrochemical potential of the left reservoir ( $\mu_L \equiv 0$ ), the relations take the form

$$Q(t) = -\frac{C}{e}\mu_p(t),$$
(3a)

$$E(t) = \frac{C}{2e^2}\mu_p^2(t) + \frac{C_E(T_p)T_p(t)}{2}.$$
 (3b)

Importantly, the charge and energy in the probe can change when the charge and energy currents,  $I_p(t)$  and  $I_p^E(t)$ , flow into the probe from the current source. In this case, the overall charge and energy conservation imposes the following continuity equations:

$$\frac{dQ(t)}{dt} = I_p(t), \tag{4a}$$

$$\frac{dE(t)}{dt} = I_p^E(t) - \Sigma \mathcal{V}[T_p^n(t) - T_0^n].$$
(4b)

Here, the term  $\Sigma \mathcal{V}[T_p^n(t) - T_0^n]$  captures the fact that an additional time-dependent energy outflow also occurs due to the coupling of probe electrons to lattice phonons.

#### **B.** Time-averaged quantities

In the next step, we evaluate the time-averaged electrochemical potential  $\overline{\mu}_p$  and temperature  $\overline{T}_p$  in the probe. Note that the time with respect to which these quantities are averaged must be much longer than the response time of the probe. Since there is no long-time charge buildup in the probe, we conclude from the charge-conservation relation (4a) that  $\overline{I}_p = 0$ . In consequence, the electrochemical potential in the probe must adjust so that this constraint is fulfilled. This is the detection principle of the probe: the buildup of the potential allows us to make conclusions about the charge current into the probe (and analogously from the temperature for the conservation of energy).

The total, time-averaged current flowing through the QPC into the probe is derived from well-known expressions [31,48] and has the form

$$\begin{split} \bar{I}_p &= D\left[\bar{I}_s + \frac{e}{h}\int dE\{f_p(E) - f_L(E)\}\right] \\ &= D\left[\bar{I}_s + g\frac{\overline{\mu}_p}{e}\right], \end{split} \tag{5}$$

with  $f_{\alpha}(E) = \{1 + \exp[(E - \mu_{\alpha})/(k_B T_{\alpha})]\}^{-1}$  denoting the Fermi function for the left reservoir ( $\alpha = L$ ) and the probe ( $\alpha = p$ ). Moreover, the total current (5) consists of a direct contribution from the source,  $\bar{I}_s$ , and a current flowing out of the probe, due to the probe potential  $\overline{\mu}_p$ . The prefactor D in Eq. (5) stems from the conservation of the particle current at the QPC. Together with the condition of zero average charge current in the probe ( $\bar{I}_p = 0$ ) discussed above [23,24], this yields

$$\overline{\mu}_p = -\frac{e}{g}\overline{I}_s.$$
 (6)

As a result, one can see that the time-averaged probe electrochemical potential  $\overline{\mu}_p$  is solely determined by the time-averaged source current  $\overline{I}_s$ , and is independent of the QPC transparency. In general, the explicit form of this current relies on a specific implementation of the current source. In this paper, it is illustrated for the example of a time-dependently-driven mesoscopic capacitor in Sec. V, derived by means of the Floquet scattering matrix approach (see Appendix A). Importantly, in the present section, the current  $\overline{I}_s$  enters the formalism as a general variable.

Similar to the charge, there is also no long-time accumulation of energy in the probe. From the time average of the energy-current conservation law, Eq. (4b), we thus obtain

$$\bar{I}_p^E - \Sigma \mathcal{V}[\bar{T}_p^n - T_0^n] = 0.$$
<sup>(7)</sup>

Furthermore, the energy-current conservation at the QPC implies that

$$\bar{I}_{p}^{E} = D\left[\bar{I}_{s}^{E} - \frac{1}{h}\int dEE\{f_{p}(E) - f_{L}(E)\}\right]$$
$$= D\left[\bar{I}_{s}^{E} - \frac{\bar{I}_{s}^{2}}{2g} - \frac{Lg}{2}(\bar{T}_{p}^{2} - T_{L}^{2})\right].$$
(8)

Three different contributions can be identified in the equation above: the bare contribution from the current source  $(\bar{I}_s^E)$ , a contribution stemming from Joule heating  $(\propto \bar{I}_s^2)$ , and the last term representing the energy current induced by the temperature difference between the left reservoir and the probe. Again, the energy current  $\bar{I}_s^E$  associated with the source enters here as a general input variable, and it is calculated for the example of a time-dependently-driven mesoscopic capacitor in Sec. V. With Eqs. (7) and (8), we write the equation from which the average probe temperature  $\bar{T}_p$  can be found [49],

$$\overline{T}_{p}^{n} + \frac{D}{\Sigma \mathcal{V}} \frac{Lg}{2} \overline{T}_{p}^{2} = T_{0}^{n} + \frac{D}{\Sigma \mathcal{V}} \left[ \overline{I}_{s}^{E} - \frac{\overline{I}_{s}^{2}}{2g} + \frac{Lg}{2} T_{L}^{2} \right].$$
(9)

Since this expression shows that  $\overline{T}_p$  is determined by the average source charge and energy currents,  $\overline{I}_s$  and  $\overline{I}_s^E$ , the

readout of this temperature—together with the readout of the probe electrochemical potential  $\overline{\mu}_p$ —serves as a means for the detection of the average source energy current. Furthermore, Eq. (9) also depends on the temperatures of the left reservoir,  $T_L$ , and of the phonon bath,  $T_0$ , as well as on the ratio  $D/(\Sigma V)$ . Importantly, the QPC transmission D is easily tunable in experiments via gate voltages, and also the electron-phonon coupling strength can, in principle, be tuned, e.g., by adjusting either the electron density [28,42,50] or the background temperature [50].

#### C. Fluctuations, Boltzmann-Langevin approach

Finally, in the last step, we consider the fluctuations *in* time using the Boltzmann-Langevin approach [30]. For this purpose, we begin by dividing the probe currents  $I_p(t)$  and  $I_p^E(t)$  into time-averaged parts and fluctuating time-dependent parts,

$$I_p(t) = \bar{I}_p + \Delta I_p(t), \qquad (10a)$$

$$I_p^E(t) = \bar{I}_p^E + \Delta I_p^E(t).$$
(10b)

The time-averaged parts,  $\bar{I}_p$  and  $\bar{I}_p^E$ , have been discussed in Sec. III B. On the other hand, for the fluctuating parts,  $\Delta I_p(t)$  and  $\Delta I_p^E(t)$ , we distinguish two distinct physical origins: (i) bare-injector fluctuations,  $\delta I_p(t)$  and  $\delta I_p^E(t)$ , which arise from the source currents or from scattering at the QPC; (ii) the fluctuations in the probe currents induced by fluctuations of the probe electrochemical potential  $\mu_p(t)$  and of the probe temperature  $T_p(t)$ . As a result, it becomes convenient to decompose also the potential and temperature fluctuations into a constant and a fluctuating contribution,

$$\mu_p(t) = \overline{\mu}_p + \Delta \mu_p(t), \qquad (11a)$$

$$T_p(t) = \overline{T}_p + \Delta T_p(t).$$
(11b)

Here we make the crucial assumption that the fluctuations of  $\delta I_p(t)$  and  $\delta I_p^E(t)$  are much faster than the fluctuations of  $\Delta \mu_p(t)$  and  $\Delta T_p(t)$ , see Sec. II B, which enables a separation of timescales. In other words, the probe is sufficiently large, so that the change in  $\mu_p(t)$  and  $T_p(t)$  is small on the timescale of the particle injection, which allows us to expand all fluctuation-dependent quantities up to the linear order in  $\Delta \mu_p(t)$  and  $\Delta T_p(t)$ —note that the long-time changes in  $\mu_p(t)$  and  $T_p(t)$  do not necessarily need to be small for this expansion to be valid. In consequence, we obtain

$$\begin{pmatrix} \Delta I_p(t) \\ \Delta I_p^E(t) \end{pmatrix} = \begin{pmatrix} \delta I_p(t) \\ \delta I_p^E(t) \end{pmatrix} + \begin{pmatrix} \frac{\partial I_p}{\partial \overline{\mu}_p} & \frac{\partial I_p}{\partial \overline{T}_p} \\ \frac{\partial \overline{I}_p^E}{\partial \overline{\mu}_p} & \frac{\partial \overline{I}_p^E}{\partial \overline{T}_p} \end{pmatrix} \begin{pmatrix} \Delta \mu_p(t) \\ \Delta T_p(t) \end{pmatrix}.$$
(12)

The bare fluctuations,  $\delta I_p(t)$  and  $\delta I_p^E(t)$ , constitute the *Langevin sources* of fluctuations in our theoretical approach, and they are explicitly derived for the case of a time-dependently-driven mesoscopic capacitor in Sec. V. The prefactors of the fluctuations  $\Delta \mu_p(t)$  and  $\Delta T_p(t)$  in Eq. (12) can be straightforwardly calculated from Eqs. (5) and (8) for  $\bar{I}_p$  and  $\bar{I}_p^E$ , respectively,

$$\begin{pmatrix} \frac{\partial I_p}{\partial \overline{\mu}_p} & \frac{\partial I_p}{\partial \overline{T}_p} \\ \frac{\partial \overline{I}_p^E}{\partial \overline{\mu}_p} & \frac{\partial \overline{I}_p^E}{\partial \overline{T}_p} \end{pmatrix} = Dg \begin{pmatrix} 1/e & 0 \\ -\overline{\mu}_p/e^2 & -L\overline{T}_p \end{pmatrix}.$$
 (13)

Note that we are considering a conductor which does not break electron-hole symmetry, which essentially means that the thermoelectric linear-response coefficients vanish, that is,  $\partial \bar{I}_p / \partial \bar{T}_p = 0$  and  $\partial \bar{J}_p / \partial \bar{\mu}_p = 0$ .

Next, employing Eqs. (3) and (4), we relate the barecurrent fluctuations,  $\delta I_p(t)$  and  $\delta I_p^E(t)$ , to the electrochemical potential  $\Delta \mu_p(t)$  and temperature  $\Delta T_p(t)$  fluctuations of the probe. It turns out that the sought relations are especially convenient to derive in frequency space. By combining the Fourier transforms of Eqs. (3) and (4), and inserting these expressions into Eq. (10), we obtain formulas relating the macroscopic probe variables  $\Delta \mu_p(\omega)$  and  $\Delta T_p(\omega)$ to the bare fluctuations of the charge current  $\delta I_p(\omega)$  and the heat current,  $\delta J_p(\omega) = \delta I_p^E(\omega) + \overline{\mu}_p \delta I_p(\omega)/e$ , which have the form

$$\frac{\Delta\mu_p(\omega)}{-e} = \frac{1}{Dg} \frac{\delta I_p(\omega)}{1 + i\omega\tau_{RC}},$$
(14)

$$\Delta T_p(\omega) = \frac{1}{\kappa} \frac{\delta J_p(\omega)}{1 + i\omega\tau_E}.$$
(15)

Importantly, Eqs. (14) and (15) are the *key input* for the realization of heat- and charge-current noise detection via potential and temperature fluctuations. Here, we define  $\kappa \equiv \kappa(\overline{T}_p)$  as the thermal conductance at the probe temperature  $\overline{T}_p$ . Similarly, we define  $C_E \equiv C_E(\overline{T}_p)$ , which, together with  $\kappa$ , enters the energy relaxation time  $\tau_E = C_E/\kappa$ .

The analysis of Eqs. (14) and (15) indicates that there is an effective timescale on which the potential and temperature fluctuations have to be small in order for the linearization performed in Eq. (12) to hold. We find these effective timescales to be  $\sqrt{1/\omega^2 + \tau_{RC}^2}$  [with the charge relaxation time  $\tau_{RC} = C/(Dg)$ ] and  $\sqrt{1/\omega^2 + \tau_E^2}$ . In particular, this means that the fluctuations in  $T_p(t)$  and  $\mu_p(t)$ have to be small either on the timescales set by the probe response—given by  $\tau_{RC}$  and  $\tau_E$ —or on the timescale on which the fluctuations are probed, namely,  $1/\omega$ . From a practical point of view, this implies that even when the probe-response times diverge (for example,  $\tau_{RC}$  for  $D \rightarrow 0$ ), linear relations between source-current fluctuations and electrochemical potential and temperature fluctuations, Eqs. (14) and (15), can still be found at finite frequencies  $\omega$ .

## IV. ELECTROCHEMICAL POTENTIAL AND TEMPERATURE FLUCTUATIONS

## A. Relation to probe-current fluctuations

With the use of Eqs. (14) and (15), we can now express the charge-current noise, the heat-current noise and the mixed noise in terms of the (symmetrized) frequencydependent correlators of electrochemical potential and temperature. The latter are given by

$$\pi \delta(\omega + \omega') \mathcal{P}_p^{XY}(\omega)$$
  
=  $\langle \Delta X_p(\omega) \Delta Y_p(\omega') + \Delta X_p(\omega') \Delta Y_p(\omega) \rangle_{\text{stat}},$  (16)

with  $X, Y = T, \mu$ , and  $\langle \cdots \rangle_{\text{stat}}$  denoting the statistical average over Langevin sources. Note that the delta function  $\delta(\omega + \omega')$  stems from the invariance of fluctuations with respect to translations in time. This is a valid assumption because the time-dependent driving of the source takes place on a much faster timescale than the fluctuations of the probe quantities [51]. The two autocorrelators  $\mathcal{P}_p^{\mu\mu}(\omega)$ and  $\mathcal{P}_p^{TT}(\omega)$ , as well as the mixed correlator  $\mathcal{P}_p^{\mu T}(\omega)$ , all defined by Eq. (16), depend on the correlators of the bare fluctuations  $\delta I_p(\omega)$  and  $\delta J_p(\omega)$ , the Langevin sources, through Eqs. (14) and (15). They are equivalently written as

$$\pi \delta(\omega + \omega') \mathcal{P}_{p}^{AB}(\omega)$$
  
=  $\langle \delta A_{p}(\omega) \delta B_{p}(\omega') + \delta A_{p}(\omega') \delta B_{p}(\omega) \rangle_{\text{stat}},$  (17)

for A, B = I, J. These bare-source correlators, in the frequency regime relevant for Eq. (16), are just given by the low-frequency, period-averaged limit,

$$\mathcal{P}_p^{AB}(\omega) \approx \mathcal{P}_p^{AB}(\omega=0) \equiv \mathcal{P}_p^{AB}.$$
 (18)

This is because the macroscopic probe quantities,  $T_p(\omega)$ and  $\mu_p(\omega)$ , fluctuate on much smaller frequencies than the probe currents,  $\delta A_p(\omega)$  and  $\delta B_p(\omega)$ . The probe currents,  $I_p(t)$  and  $J_p(t)$ , are typically of quantum-mechanical nature. Thus, for the evaluation of the correlators  $\mathcal{P}_p^{AB}$  of Eq. (17), these currents have to be replaced by the corresponding operators  $\hat{I}_p(t)$  and  $\hat{J}_p(t)$ ,

$$\mathcal{P}_{p}^{AB} = \frac{1}{2} \int_{0}^{\mathcal{T}} \frac{dt}{\mathcal{T}} \int_{-\infty}^{\infty} dt' \langle \{\delta \hat{A}_{p}(t), \delta \hat{B}_{p}(t+t')\} \rangle, \quad (19)$$

with  $\{\cdots, \cdots\}$  standing for the anticommutator and  $\langle\cdots\rangle$  representing the quantum-statistical average. The explicit expressions for  $\mathcal{P}_p^{AB}$  are calculated using the quantum-coherent Floquet scattering theory, see Sec. V and Appendixes A and B for general expressions and further details of the derivation.

Combining Eqs. (16) and (17), and then employing relations (14) and (15), we can write the frequency-dependent probe electrochemical potential and temperature correlators,  $\mathcal{P}_p^{\mu\mu}(\omega)$  and  $\mathcal{P}_p^{TT}(\omega)$ , which are directly proportional to the corresponding charge- and heat-current correlators,  $\mathcal{P}_p^{II}$  and  $\mathcal{P}_p^{JJ}$ ,

$$\mathcal{P}_{p}^{\mu\mu}(\omega) = \frac{e^{2}}{(Dg)^{2}} \frac{1}{1 + (\omega\tau_{RC})^{2}} \mathcal{P}_{p}^{II}, \qquad (20)$$

$$\mathcal{P}_p^{TT}(\omega) = \frac{1}{\kappa^2} \frac{1}{1 + (\omega \tau_E)^2} \mathcal{P}_p^{JJ}.$$
 (21)

An analogous expression can be found also for the mixed correlator,

$$\mathcal{P}_{p}^{\mu T}(\omega) = \frac{-e}{Dg\kappa} \frac{1 + \omega^{2} \tau_{RC} \tau_{E}}{[1 + (\omega \tau_{E})^{2}][1 + (\omega \tau_{RC})^{2}]} \mathcal{P}_{p}^{IJ}.$$
 (22)

Here, the order of the superscript of this symmetrized correlator is irrelevant, that is,  $\mathcal{P}_p^{\mu T}(\omega) \equiv \mathcal{P}_p^{T\mu}(\omega)$  and  $\mathcal{P}_p^{IJ} \equiv \mathcal{P}_p^{JI}$ . The three equations above establish the direct relation between charge- and *heat*-current noise in the probe and the frequency-dependent fluctuations of its macroscopic quantities—the temperature and electrochemical potential. These equations provide the basis for the proposed detection scheme, making the properties of charge and energy currents emitted from the source experimentally accessible.

Nevertheless, in Eqs. (21) and (22), information about the *energy*-current fluctuations is still obscured. To extract the equivalent correlators for the (mixed) *energy*-current fluctuations ( $\mathcal{P}_p^{IE}$ )  $\mathcal{P}_p^{EE}$ , we invoke the relation between the heat- and energy-current fluctuations,  $\delta J_p(\omega) = \delta I_p^E(\omega) + \overline{\mu}_p \delta I_p(\omega)/e$ . (To keep the notation simple, here we write E instead of  $I^E$  in the superscripts of  $\mathcal{P}$ .) This allows us to find  $\mathcal{P}_p^{IE}$  and  $\mathcal{P}_p^{EE}$  from the three correlators  $\mathcal{P}_p^{II}$ ,  $\mathcal{P}_p^{JJ}$ , and  $\mathcal{P}_p^{IJ}$  introduced above, by applying step-by-step the relations

$$\mathcal{P}_{p}^{IJ} = \mathcal{P}_{p}^{IE} + \frac{\overline{\mu}_{p}}{e} \mathcal{P}_{p}^{II}, \qquad (23)$$

$$\mathcal{P}_{p}^{JJ} = \mathcal{P}_{p}^{EE} + \frac{\overline{\mu}_{p}}{e} \mathcal{P}_{p}^{IE} + \frac{\overline{\mu}_{p}^{2}}{e^{2}} \mathcal{P}_{p}^{II}, \qquad (24)$$

and employing the knowledge of the probe potential obtained from independent measurements.

Furthermore, the frequency-dependence of the (mixed) potential and temperature fluctuations, given in Eqs. (20)–(22), sets a fundamental constraint on the bandwidth of the measurement, that is, the smaller the relaxation times  $\tau_{RC}$  and  $\tau_E$ , the less the bandwidth is limited. From the experimental point of view, an increased bandwidth is highly desirable as it allows for a measurement speedup, which is particularly important when a large number of measurements has to be performed in order to guarantee an optimal subtraction of a fluctuating background signal.

Indeed, the subtraction of a background signal turns out to be crucial when one aims at extracting the desired fluctuations of the *source* currents from the *probe*-current fluctuations. How this can be done is shown in the following correlator decomposition, which for simplicity is performed on the correlators  $\mathcal{P}_p^{AB}$ , for A, B = I, J, since the frequency dependence of Eqs. (20)–(22) is not crucial for this decomposition.

## **B.** Correlator decomposition

In the previous section, we demonstrated how fluctuations of the temperature  $T_p(t)$  and electrochemical potential  $\mu_p(t)$  in the probe are related to the relevant charge-current ( $\mathcal{P}_p^{II}$ ) and heat-current ( $\mathcal{P}_p^{IJ}$ ) autocorrelators, as well as to the mixed-current ( $\mathcal{P}_p^{IJ}$ ) correlator *in the probe*. However, in order to acquire a complete understanding on how the sought *source fluctuations* are influenced by the detection—or in other words, how the correlators of the probe currents are linked to the correlators of the source currents—it is instructive to divide  $\mathcal{P}_p^{AB}$  into terms with physically distinct origins.

The first step involves writing  $\mathcal{P}_p^{AB}$  as a sum of two terms: the first  $(\mathcal{P}_0^{AB})$  arising already in the absence of a working source, and the second  $(\mathcal{P}_s^{AB})$  that can be attributed to the source driving,

$$\mathcal{P}_p^{AB} = \mathcal{P}_0^{AB} + \mathcal{P}_s^{AB}.$$
 (25)

While  $\mathcal{P}_0^{AB}$  is due to the nonzero, constant temperatures of the left reservoir and the probe,  $\mathcal{P}_s^{AB}$  contributes only if the source is on and is equal to zero otherwise. Consequently, when evaluating  $\mathcal{P}_0^{AB}$ , the probe effectively acts as a second, right (*R*), reservoir characterized by a distribution function  $f_R(E) = \{1 + \exp[E/(k_B T_R)]\}^{-1}$ , with the temperature  $T_R$  determined from Eq. (9). In an experiment, this contribution is obtained from a noise measurement carried out when the time-dependent driving is switched off. In order to obtain  $\mathcal{P}_s^{AB}$ , one has to subtract  $\mathcal{P}_0^{AB}$  from results of subsequent measurements of  $\mathcal{P}_p^{AB}$  with a working source. In the second step, the fluctuations attributed to the driv-

In the second step, the fluctuations attributed to the driving source  $\mathcal{P}_s^{AB}$  can be further separated into direct  $(\mathcal{P}_{s,dir}^{AB})$ and induced  $(\mathcal{P}_{s,ind}^{AB})$  correlations as

$$\mathcal{P}_{s}^{AB} = \mathcal{P}_{s,\text{dir}}^{AB} + \mathcal{P}_{s,\text{ind}}^{AB}.$$
 (26)

The direct correlations are the ones obtained, if the probe had been kept at constant temperature and electrochemical potential, with the distribution function  $f_R(E)$ . It essentially means that these correlations arise due to the source driving in the absence of voltage buildup and temperature raise induced by the detection itself. In fact, it is precisely this term that the detection scheme proposed in this paper aims to extract. In order to obtain  $\mathcal{P}_{s,dir}^{AB}$  theoretically from  $\mathcal{P}_s^{AB}$ , the temperature and electrochemical potential have to be set to their equilibrium values  $\overline{T}_p$  and  $\overline{\mu}_p$ , see Eq. (A9). On the other hand, the remaining term  $\mathcal{P}_{s,\text{ind}}^{AB}$  corresponds to the induced correlations, which originate from the fluctuations of the probe properties,  $\mu_p(t)$  and  $T_p(t)$ , induced by the source driving. Importantly, these correlations represent the back-action effect resulting from the measurement itself [52]. In an experiment, such a separation into induced and direct parts of the source fluctuations is not possible. For this reason, it is important to analyze the relative magnitude of these terms as we do for the example of a mesoscopic capacitor acting as the current source in the following section.

## V. TIME-DEPENDENTLY-DRIVEN MESOSCOPIC CAPACITOR

We now provide a proof of principle for our approach and consider a specific time-dependent injector source generating a quantized, alternating current. We assume that the total injector-detector system is implemented with edge states in a two-terminal conductor in the integer quantum Hall regime with filling factor one, as shown in Fig. 2. In particular, the injector part of our exemplary system consists of a mesoscopic capacitor, as realized experimentally by Fève et al. [1]. Due to the confinement of charge in the mesoscopic capacitor plate, the source has a discrete energy spectrum characterized by the level spacing  $\Delta$ . This source is side-coupled to the active edge state via a QPC with transparency  $D_s$ . By applying a time-periodic potential U(t) to the capacitor, under ideal conditions (specified below), the capacitor emits exactly one electron and one hole per driving period  $\mathcal{T}$ . Moreover, both the left reservoir ( $\alpha = L$ ) and the probe ( $\alpha = p$ ) are described by Fermi



FIG. 2. Edge-state implementation of the injector-detector system with the single-particle source (SPS) that consists of a mesoscopic capacitor slowly driven by a time-periodic potential U(t). The capacitor is connected to the edge state via a QPC with transparency  $D_s$ . The transport along the edge states is chiral, with the direction of transport shown by arrows.

functions  $f_{\alpha}(E)$ , with electrochemical potentials  $\mu_{\alpha}$  and temperatures  $T_{\alpha}$ . Recall that the left reservoir—namely, the terminal at the injector side—is assumed throughout the paper to be grounded ( $\mu_L = 0$ ) and its temperature equals the phonon temperature ( $T_L = T_0$ ), and so does the probe in the case when the source driving is switched off. This means, in turn, that there is no direct thermal bias, except the one induced by the detection.

We describe the coherent conductor shown in Fig. 2 by the time-dependent (Floquet) scattering matrix approach. Since the system is periodically driven in time, the scattering matrix elements,  $S_{\alpha\beta}(E,t)$  for  $\alpha, \beta = L, p$ , depend, in general, both on the initial time and the final energy of the scattering process (or equivalently on two times or on two energies [31]). For the scatterer considered in Fig. 2, with a forward-scattering mesoscopic capacitor coupled to one edge channel, connected in series with a QPC with energyindependent transparency *D*, one can write the components of the time- and energy-dependent scattering matrix as

$$S_{LL}(E, t) = \sqrt{1 - D}S(E, t),$$

$$S_{pL}(E, t) = \sqrt{D}S(E, t),$$

$$S_{Lp}(E, t) = \sqrt{D},$$

$$S_{pp}(E, t) = -\sqrt{1 - D}.$$
(27)

Here, S(E, t) is the scattering matrix of the mesoscopic capacitor only.

#### A. Adiabatic response to slow driving

Let us consider the adiabatic response to a slow driving potential U(t) of the mesoscopic capacitor. This implies that the period of the drive  $T = 2\pi/\Omega$  is much larger than the temporal width  $\sigma$  of the wave packets associated with

the particles emitted by the source. The width  $\sigma$  is a function of the transparency  $D_s$ , the level spacing  $\Delta$  and the driving potential U(t) [31], and is treated here as an input parameter. In addition, we consider the regime where the thermal energies of injector and probe terminals,  $k_B T_0$  and  $k_B \overline{T}_p$ , are much smaller than the energy scale  $\hbar/\sigma$  over which the scattering properties vary appreciably [53]. Consequently, we define the driving regime of interest by the following condition:

$$\hbar\Omega, k_B T_0, k_B \overline{T}_p \ll \frac{\hbar}{\sigma}.$$
 (28)

In this adiabatic-response limit the scattering properties of the mesoscopic capacitor are characterized by the *frozen* scattering matrix  $S_0(E, t)$ , where the time t only enters parametrically. The low-temperature condition further allows us to evaluate the scattering matrix at the Fermi energy of the left terminal, E = 0. This gives for  $S(t) \equiv S_0(0, t)$ , up to irrelevant phase factors [31],

$$S(t) = \begin{cases} \frac{t - t_e + i\sigma}{t - t_e - i\sigma} & \text{for } 0 \le t < \mathcal{T}/2, \\ \frac{t - t_h - i\sigma}{t - t_h + i\sigma} & \text{for } \mathcal{T}/2 \le t < \mathcal{T}, \end{cases}$$
(29)

where the driving is such that only a single energy level participates in the emission process. Here, the times for electron  $(t_e)$  and hole  $(t_h)$  emission, defined by the capacitor and drive potential properties [31], are assumed to be well inside the respective halves of the period, so that temporal overlap of the emitted particle wave functions can be neglected.

In the energy description, the scattering matrix takes the general form

$$S(E_n, E_m) = \int \frac{dt}{\mathcal{T}} e^{i(n-m)\Omega t} S(E_n, t), \qquad (30)$$

with  $E_n = E + n\hbar\Omega$  and  $S(E_n, E_m)$  denoting the Floquet scattering matrix. Here,  $n - m \in \mathbb{Z}$  is the number of Floquet energy quanta picked up during a scattering process. In the energy description, the adiabatic condition corresponds to the assumption that the scattering properties are energy independent on the scale of  $N_{\max}\hbar\Omega$ , where  $N_{\max}$ stands for the maximum number of quanta absorbed or emitted in the scattering process at the capacitor. The corresponding Floquet scattering matrix amplitudes then only depend on the difference in energies,  $S(E_n, E_m) = S(E_n - E_m) \equiv S_{n-m}$ , and up to the leading order in  $\sigma \Omega$ , one finds [54]

$$S_n = \begin{cases} -2\Omega\sigma e^{-n\Omega\sigma} e^{in\Omega t_e} & \text{for } n > 0, \\ -2\Omega\sigma e^{n\Omega\sigma} e^{in\Omega t_h} & \text{for } n < 0, \\ \delta_{n,0} & \text{for } n = 0. \end{cases}$$
(31)

With this expression at hand, we can evaluate all quantities of interest, that is, the average values of charge and heat currents, as well as the relevant current correlators. See Appendix A for the outline of the derivation and specific formulas.

In the following, we assume a metallic probe where the heat flow into the phonon bath is given by  $\Sigma \mathcal{V}[T_p^5 - T_0^5]$ , namely, we now set the temperature exponent to n=5. Furthermore, whenever we calculate specific numerical example values, we take the following set of parameters, corresponding to current experiments as discussed in Sec. II: We assume the Coulomb interaction in the probe to be weak,  $E_C/k_B \approx 3$  mK, much smaller than the energy scale given by the base (phonon) temperature, which we set to  $T_0 = 10$  mK. For a QPC transmission of D = 0.1, this would lead to an *RC* time of the order of  $\tau_{RC} = 100$ ns. Taking a source-driving frequency of  $\Omega/(2\pi) \approx 1-10$ GHz and assuming the electron-electron relaxation time  $\tau_{e-e} = 1$  ns, this fulfills the requirements set by Eq. (2). The energy relaxation time of the probe depends on the probe temperature and other tunable parameters and we will not fix it for every example; however, taking for instance v = $3 \times 10^{29} \text{ J}^{-1}$ ,  $\kappa = 10 \text{ fW/K}$  and  $T_p = 50 \text{ mK}$ , we would achieve a probe energy relaxation time  $\tau_E$  of the order of several 100  $\mu$ s, which easily fulfills the requirements of Eq. (2).

#### **B.** Average potential and temperature

In order to determine the average temperature  $T_p$  and electrochemical potential  $\overline{\mu}_p$  arising in the probe due to the driving, the average charge and energy currents have to be explicitly calculated using Eqs. (A5) and (A6) from Appendix A for the scattering matrix given in Eqs. (27) and (31). The average source-charge current is  $\overline{I}_s = 0$ , which comes as a consequence of the fact that the same number of electrons and holes is emitted by the source per period. Therefore, the average electrochemical potential of the probe, given by Eq. (6), is also zero,

$$\overline{\mu}_p = 0. \tag{32}$$

Furthermore, the average energy current emitted by the source is  $\bar{I}_s^E = 2\langle \epsilon \rangle / \mathcal{T}$ , where the average energy of each emitted particle is  $\langle \epsilon \rangle = \hbar/(2\sigma)$  and the factor 2 stems from the fact that both an electron and a hole are emitted per period. Inserting this expression together with  $\bar{I}_s = 0$  into Eq. (9), we obtain an equation allowing us to find the average probe temperature,

$$\left(\frac{\overline{T}_p}{\overline{T}_0}\right)^5 - 1 = \frac{5}{2} \frac{\kappa_{e,0}}{\kappa_{\text{ph},0}} \left[ 1 - \left(\frac{\overline{T}_p}{\overline{T}_0}\right)^2 + \frac{4\langle\epsilon\rangle}{\mathcal{T}gLT_0^2} \right]. \quad (33)$$



FIG. 3. (a) Renormalized probe temperature  $\overline{T}_p/T_0$  shown on a logarithmic scale as a function of the ratio of the electron and phonon thermal conductances  $\kappa_{e,0}/\kappa_{\rm ph,0}$  for selected values of the average energy of an emitted particle per period,  $\langle \epsilon \rangle/T$ , scaled to the electronic heat-current response of the probe,  $gLT_0^2$ . (b) Dependence of the renormalized probe temperature  $\overline{T}_p/T_0$  (plotted logarithmically) on the ratio  $\langle \epsilon \rangle/(TgLT_0^2)$  for different values of  $\kappa_{e,0}/\kappa_{\rm ph,0}$ .

Here,  $\kappa_{e,0} \equiv \kappa_e(T_0)$  and  $\kappa_{ph,0} \equiv \kappa_{ph}(T_0)$  are the electronic and the phonon thermal conductivity at the base temperature. The expression  $gLT_0^2$  is the pure electronic heat response of the probe at the base temperature; it is equal to  $\kappa_{e,0}T_0$  when D = 1. The numerical solution for Eq. (33) is presented in Fig. 3. Specifically, (a) shows the (renormalized) average probe temperature  $\overline{T}_p/T_0$  as a function of the ratio between thermal conductances  $\kappa_{e,0}$  and  $\kappa_{ph,0}$ . When the source is switched off (solid line), which effectively corresponds to  $\langle \epsilon \rangle = 0$ , the probe temperature is naturally always given by the phonon temperature  $T_0$ . On the other hand, the larger the energy emitted from the source, the stronger the deviation of the probe temperature from the phonon temperature. In addition, the monotonic increase of  $\overline{T}_p/T_0$  shown in Fig. 3(a) is intuitively clear, since the increase of the temperature is expected to be larger in the limit when the energy loss to the phonon bath is small, that is, for  $\kappa_{e,0}/\kappa_{\text{ph},0} \gg 1$ .

Employing the example values given above, with  $T_0 = 10 \text{ mK}$  and  $\Omega/(2\pi) = 1 \text{ GHz}$ , we find a probe temperature of  $\overline{T}_p \approx 5T_0 = 50 \text{ mK}$  at a pulse width  $\sigma \approx 30 \text{ ps}$  (corresponding to the energy ratio  $\langle \epsilon \rangle/(TgLT_0^2) \approx 10$ ) and  $\kappa_{e,0}/\kappa_{\text{ph},0} = 10^2$ .

The dependence of the probe temperature on the energy emitted per particle in a period,  $\langle \epsilon \rangle / \mathcal{T}$  (normalized with respect to the electronic probe heat current  $gLT_0^2$ ), is shown in (b) of Fig. 3. As mentioned above, the monotonic increase of the temperature fits in with the expectation that the value of the probe temperature is a signature of the energy transported into the probe by the particles emitted from the source. This is, however, true only as long as the coupling to the phonon bath is not too strong: in the opposite limit of very strong coupling ( $\kappa_{e,0}/\kappa_{ph,0} \rightarrow 0$ ), the probe temperature is always equal to the phonon temperature independently of the source driving, see the solid line in Fig. 3(b). In contrast, if the coupling to the phonon bath is negligibly small, the average temperature in the probe is given by

$$\frac{\overline{T}_p}{T_0} = \sqrt{1 + \frac{4\langle\epsilon\rangle}{\mathcal{T}gLT_0^2}} \quad \text{(for } \kappa_{\text{ph},0} \ll \kappa_{e,0}\text{)}. \tag{34}$$

Finally, we note that while the solution of Eq. (33) for temperatures  $\overline{T}_p$  can only be obtained numerically, we can still estimate the linear response, corresponding to the limit of  $\overline{T}_p - T_0 \ll T_0$ . In this regime, the drive has a small effect on the probe temperature, such that the analytic formula for the renormalized deviation from the bath temperature is given by

$$\frac{\overline{T}_p - T_0}{T_0} = \frac{2\langle \epsilon \rangle}{\mathcal{T}gLT_0^2} \left(1 + \frac{\kappa_{\text{ph},0}}{\kappa_{e,0}}\right)^{-1}.$$
 (35)

### C. Potential and temperature fluctuations

We now turn our attention to the quantities of main interest in this paper, that is, the fluctuations of the probe temperature and electrochemical potential,  $\mathcal{P}_p^{XY}(\omega)$ , with  $X, Y = \mu, T$ . Their frequency-dependent relation to the correlators of the probe charge and heat currents has been established in Sec. IV A, see Eqs. (20)–(22). In order to obtain explicit expressions for the example of a slowly driven mesoscopic capacitor discussed in this section, we have to derive correlators  $\mathcal{P}_p^{AB}$  for A, B = I, J. Starting from the general expression given in Appendix A, this means that the following formula needs to be evaluated:

$$\mathcal{P}_{p}^{AB} = D(1-D)\frac{1}{h}\int dE x_{A}x_{B}\sum_{n}|S_{n}|^{2}\{f_{L}(E_{n})[1-f_{p}(E)] + f_{p}(E)[1-f_{L}(E_{n})]\} + D^{2}\frac{1}{h}\int dE \left\{x_{A}x_{B}f_{p}(E)[1-f_{p}(E)] + \sum_{nkq}\frac{x_{A}x_{B,n} + x_{A,n}x_{B}}{2}S_{-q}^{*}S_{n-q}S_{-k}S_{n-k}^{*}f_{L}(E_{q})[1-f_{L}(E_{k})]\right\},$$
(36)

where the notation  $x_{A,n}$  should be read as  $x_{I,n} = x_I \equiv e$  and  $x_{J,n} \equiv E_n - \overline{\mu}_p$  (with  $x_J \equiv x_{J,0}$ ). Straightforward calculations employing the scattering matrix given in Eq. (31) yield the desired expressions for the correlator components  $\mathcal{P}_0^{AB}$ ,  $\mathcal{P}_{s,dir}^{AB}$ , and  $\mathcal{P}_{s,ind}^{AB}$ , introduced in Sec. IV B, which will be discussed in the remaining part of this section. For explicit derivations of  $\mathcal{P}_0^{AB}$  and  $\mathcal{P}_{s,dir}^{AB}$ , we refer the reader, e.g., to Ref. [14,17], where the equilibrium correlations, corresponding to  $\mathcal{P}_0^{AB}$ , are calculated, while more details regarding the derivation of the direct contribution  $\mathcal{P}_{s,dir}^{AB}$  will be presented in Ref. [38]. In contrast, the evaluation of the induced part  $\mathcal{P}_{s,ind}^{AB}$ , which is unique for the type of measurement we are proposing here, is described in more detail in Appendix B. For the sake of compactness of the following discussion, we introduce the auxiliary function  $\mathcal{H}_q(\overline{T}_p/T_0)$  defined as

$$\mathcal{H}_q\left(\frac{\overline{T}_p}{T_0}\right) = \int \frac{dE E^q}{(k_B T_0)^{q+1}} [f_L(E) - f_p(E)] f_L(E). \quad (37)$$

# 1. Electrochemical potential correlations vs. charge-current noise

We begin with the analysis of the electrochemical potential fluctuations  $\mathcal{P}_p^{\mu\mu}(\omega)$  as a measure of the charge-current correlations, which are well known and have been measured previously [6,9,55–57]. Still, we use this correlator to demonstrate here the working principle of our injectordetector scheme for the example of a time-dependentlydriven mesoscopic capacitor acting as the current source. Starting from Eq. (20) with Eq. (36), and making use of the correlator decomposition discussed above, Eqs. (25) and (26), we can write

$$\frac{\mathcal{P}_{p}^{\mu\mu}(\omega)}{\mathcal{P}_{0}^{II}} = \frac{e^{2}}{(gD)^{2}} \frac{1}{1 + (\omega\tau_{RC})^{2}} [1 + \mathcal{F}_{s,\text{dir}}^{II} + \mathcal{F}_{s,\text{ind}}^{II}].$$
(38)

Here,  $\mathcal{P}_0^{II} = 2Dgk_BT_0$  corresponds to the well-known equilibrium thermal noise of the QPC, which is related to the QPC conductance Dg through the fluctuation-dissipation theorem, see e.g., [30]. Moreover, the direct and induced components of the renormalized, dimensionless source charge-current correlators, that is,  $\mathcal{F}_{s,\text{dir}}^{II} \equiv \mathcal{P}_{s,\text{dir}}^{II}/\mathcal{P}_0^{II}$  and

 $\mathcal{F}_{s,\text{ind}}^{II} \equiv \mathcal{P}_{s,\text{ind}}^{II} / \mathcal{P}_{0}^{II}$ , respectively, are given by

$$\mathcal{F}_{s,\text{dir}}^{II} = (1-D)\frac{\hbar\Omega}{k_B T_0},\tag{39}$$

$$\mathcal{F}_{s,\text{ind}}^{II} = \frac{D(\overline{T}_p - T_0)}{2T_0} + (1 - D)\mathcal{H}_0\left(\frac{\overline{T}_p}{T_0}\right).$$
(40)

The aim of the present analysis is to investigate the ratio between the induced and direct contributions to the fluctuations in order to demonstrate the feasibility of the detection scheme proposed in this paper. For this purpose, we fix the direct fluctuations  $\mathcal{F}_{s,dir}^{II}$  to the ideal, simple regime, described by Eq. (39). In a realistic experiment, one might instead be interested in studying the direct noise as a function of the driving signal shape [6,9,58] or other source-related properties deviating from the ideal-driving case presented here.

The direct part  $\mathcal{F}_{s,dir}^{II}$ , which in this regime only depends on the QPC transmission and the number of particles impinging on the QPC per period [59], is shown as the solid line in Fig. 4. The linear dependence of  $\mathcal{F}_{s,dir}^{II}$  on the transmission probability D is visible in Fig. 4, where the magnitude of  $\mathcal{F}_{s,dir}^{II}$  is increased by approximately one order when diminishing the transmission from D = 0.9(bottom panel) to D = 0.1 (top panel). In contrast, this part of the noise is inherently insensitive to the probe properties, quantified by the ratio  $\kappa_{ph,0}/\kappa_0$ .

The situation is different for the induced part  $\mathcal{F}_{s,ind}^{II}$ , which depends on the detector properties through the induced temperature  $\overline{T}_p$ , see Eq. (40). Since, the transmission D now enters the noise also indirectly via the fraction  $\kappa_{e,0}/\kappa_{\text{ph},0} = DgLT_0/\kappa_{\text{ph},0}$ , in Fig. 4 we choose to show different curves for selected values of the *probe* parameter  $\kappa_{\text{ph},0}/(gLT_0)$ , only, whereas different panels represent indicated values of D. It can be seen that the rise of the probe temperature  $\overline{T}_p$  with decreasing  $\kappa_{\text{ph},0}$ , demonstrated in Fig. 3(b), is also accompanied by an increase of  $\mathcal{F}_{s,ind}^{II}$ . On the other hand, the dependence of the induced part on the transmission probability D is negligible, as compared to the D dependence of the direct part. The reason for this is that the magnitudes of the two contributions,  $(\overline{T}_p - \overline{T}_p)$ .



FIG. 4. Direct  $(\mathcal{F}_{s,\text{dir}}^{II}$ , solid lines) and induced  $(\mathcal{F}_{s,\text{ind}}$ , dashed lines) components of the renormalized, dimensionless source charge-current correlators shown as a function of  $\hbar\Omega/(k_BT_0)$ for several values of the normalized electron-phonon coupling  $\kappa_{\text{ph},0}/(gLT_0)$ . Note that only  $\mathcal{F}_{s,\text{ind}}^{II}$  depends on this coupling, as can be seen from Eqs. (39) and (40). The horizontal (finely dashed) line represents the background noise, with respect to which the expressions are normalized. Different panels correspond to specific values of the QPC transparency: (a) D = 0.1, (b) D = 0.5, and (c) D = 0.9. The pulse width in all panels is  $2\Omega\sigma = 0.001$ . Finally, the lower-bound cutoff for values of  $\hbar\Omega/(k_BT_0)$  is imposed by the adiabatic condition, Eq. (28), requiring  $\hbar\Omega/(k_BT_0) \gg \sigma \Omega$ .

 $T_0)/(2T_0)$  and  $-\mathcal{H}_0(\overline{T}_p/T_0)$ , to  $\mathcal{F}_{s,\text{ind}}^{II}$ , Eq. (40), are of similar order. Therefore, the terms proportional to *D* approximately cancel each other, such that  $\mathcal{F}_{s,\text{ind}} \approx -\mathcal{H}_0(\overline{T}_p/T_0)$ . Note that  $\mathcal{F}_{s,\text{ind}}^{II}$  still weakly depends on *D* through  $\overline{T}_p$ .

Consequently, in an experiment, where the transmission probability D can be easily tuned, the relative magnitude of  $\mathcal{F}_{s,\text{dir}}^{II}$  and  $\mathcal{F}_{s,\text{ind}}^{II}$  can be adjusted as well. Indeed, the three panels of Fig. 4 clearly illustrate that, while at D = 0.9 [see Fig. 4(c)] the desired direct signal is fully covered by the induced fluctuations, a small QPC transmission D = 0.1

[see Fig. 4(a)] allows for a detection of charge-current fluctuations of the source. At D = 0.1 and  $\kappa_{ph,0}/(gLT_0) = 10^2$ , as shown in Fig. 4(a), the sought-after direct part of the charge-current fluctuations is always much larger than the induced noise, as required for a measurement of the former. A detailed discussion of the ideal choice of tunable parameters is presented in Sec. V D.

An estimate of the order of magnitude of the electrochemical potential fluctuations to be readout can be found employing the example values given above in Sec. V A. For the reference thermal charge-current noise, indicated by the gray-dashed lines in Fig. 4, we find  $\mathcal{P}_0^{II} \approx 10^{-30} \text{ A}^2 \text{ s}$ , which is in the measurable range, see, e.g., Ref. [39], and corresponds to  $\sqrt{\mathcal{P}_p^{\mu\mu}} \approx 7 \times 10^{-10} \text{ eV}/\sqrt{\text{Hz}}$ . Figure 4 shows that the sought-after direct fluctuations are typically an order of magnitude larger in the considered parameter regime.

#### 2. Temperature correlations vs heat-current noise

In order to show how to detect the behavior of the heat-current noise from the temperature fluctuations, we start from Eq. (21) together with Eq. (36), we use the decomposition introduced in Eqs. (25) and (26). This procedure leads to

$$\frac{\mathcal{P}_{p}^{II}(\omega)}{\mathcal{P}_{0}^{JJ}} = \frac{1}{\kappa^{2}} \frac{1}{1 + (\omega\tau_{E})^{2}} [1 + \mathcal{F}_{s,\text{dir}}^{JJ} + \mathcal{F}_{s,\text{ind}}^{JJ}], \quad (41)$$

where  $\mathcal{F}_{s,\text{dir}}^{JJ} \equiv \mathcal{P}_{s,\text{dir}}^{JJ}/\mathcal{P}_0^{JJ}$  and  $\mathcal{F}_{s,\text{ind}}^{JJ} \equiv \mathcal{P}_{s,\text{ind}}^{JJ}/\mathcal{P}_0^{JJ}$  stand for the direct and induced components, respectively, of the source heat-current correlator  $\mathcal{P}_s^{JJ}$ , normalized to the equilibrium heat noise  $\mathcal{P}_0^{JJ} = 2D\pi^2(k_BT_0)^3/(3h)$ . The explicit expressions for these two terms are found to be

$$\mathcal{F}_{s,\text{dir}}^{JJ} = \frac{3(2-D)}{\pi^2} \frac{1}{(2\Omega\sigma)^2} \left(\frac{\hbar\Omega}{k_B T_0}\right)^3, \qquad (42)$$

$$\mathcal{F}_{s,\text{ind}}^{JJ} = \frac{D(\overline{T}_p - T_0^3)}{2T_0^3} + \frac{3(1-D)}{\pi^2} \mathcal{H}_2\left(\frac{\overline{T}_p}{T_0}\right).$$
(43)

Both these terms are shown in Fig. 5 as a function of the driving frequency for different values of the phonon heat conductance  $\kappa_{ph,0}/(gLT_0)$ . Again, it is only the induced part that inherently depends on the probe properties entering via the ratio of thermal conductances  $\kappa_{e,0}/\kappa_{ph,0}$  in  $\overline{T}_p$ , Eq. (33). While the overall shape of the induced and direct part as a function of the driving frequencies is similar to the one for the electrochemical potential fluctuations, presented in Fig. 4, we note two striking differences.

First and foremost, for the realistic parameters presented here, the induced part  $\mathcal{F}_{s,ind}^{JJ}$  of the fluctuations is always



FIG. 5. Analogous to Fig. 4 except that the now direct (solid lines) and induced (dashed lines) components of the normalized, dimensionless source heat-current correlators are shown as a function of  $\hbar\Omega/(k_B T_0)$  for selected values of the electron-phonon coupling  $\kappa_{\rm ph,0}/(gLT_0)$ . The horizontal (finely dashed) line corresponds now to the background noise  $\mathcal{P}_0^{JJ}$ , which the noise is normalized with respect to.

smaller than the desired direct part  $\mathcal{F}_{s,\text{dir}}^{JJ}$  of the noise. This demonstrates that the proposed injector-detector setup is indeed promising for a readout of heat-current correlations via temperature fluctuations.

Second, it is worth emphasizing that for the heat-current correlator,  $\mathcal{F}_{s}^{JJ} \equiv \mathcal{P}_{s}^{JJ}/\mathcal{P}_{0}^{JJ}$ , in contrast to the chargecurrent correlator,  $\mathcal{F}_{s}^{II} \equiv \mathcal{P}_{s}^{II}/\mathcal{P}_{0}^{II}$ , discussed in Sec. 1, both contributions  $\mathcal{F}_{s,\text{dir}}^{JJ}$  and  $\mathcal{F}_{s,\text{ind}}^{JJ}$ , Eqs. (42) and (43), are basically insensitive to the transmission probability at the QPC. This can mostly be attributed to the modified *D* dependence of the direct part  $\mathcal{F}_{s,\text{dir}}^{JJ}$ , which, due to its prefactor (2 - D), is never strongly suppressed, not even for large QPC transmissions *D*. Technically, this stems from the fact that the heat-current noise, unlike the charge-current noise, contains a so-called *interference* contribution [13,60] originating from the time-dependent driving. This contribution has a linear D dependence similar to the purely thermal part, which the expression is normalized with respect to here.

The important message of this subsection is that the backaction of the measurement—resulting in the induced part of the noise—is not expected to hinder the readout of the heat-current noise, and hence of the temperature fluctuations, under realistic experimental conditions.

Also here, an estimate of the order of magnitude of the temperature fluctuations to be readout can be found employing the example values given above in Sec. V A; we obtain  $\mathcal{P}_0^{JJ} \approx 2 \times 10^{-42} \text{ W}^2$  s, while the desired direct noise is typically six orders of magnitude larger, see Fig. 5. The latter would, therefore, correspond to fluctuations of the order of  $\sqrt{\mathcal{P}_p^{TT}} \approx 10^{-4} \text{ K}/\sqrt{\text{Hz}}$ , which exceeds the noise level of current temperature measurements [29,61].

#### 3. Mixed correlator

In the discussed setup, the mixed correlators  $\mathcal{P}_p^{\mu T}$  and  $\mathcal{P}_p^{IJ}$  identically vanish. The reason for this is the absence of an electrochemical potential difference, see e.g., Ref. [14]. Note that in the detector scheme suggested here, an induced part of this mixed correlator is expected whenever the thermoelectric response-coefficients in Eq. (13) are nonvanishing.

# **D.** Optimizing parameters

In the previous section, we have seen how various tunable parameters-conductances and capacitancesinfluence the probe temperature and the detected fluctuations of temperature and electrochemical potentials in a complex way and we have seen numerical values for specifically selected sets of parameters. In this section, we give an overview of different aspects that should be taken into account when optimizing these parameters with various scopes in mind. The conductance (of importance for the potential fluctuations) is tunable via the QPC transmission D. The situation is more complicated for the thermal conductivities, which depend on temperature themselves. We, therefore, decide to lead the discussion based on the thermal conductivities  $\kappa_{e,0}$  and  $\kappa_{ph,0}$  at the base temperature  $T_0$ . Apart from tuning via the base temperature these two quantities can be tuned by changing the electron-phonon coupling  $\Sigma \mathcal{V}$  via the electron density [28,42] ( $\propto \kappa_{ph,0}$ ), or by changing the QPC transmission  $D (\propto \kappa_{e,0})$ , see Eq. (1). These two quantities influence the magnitude of the probe temperature and its fluctuations, directly or also indirectly via the probe temperature entering  $\kappa$ .

The *dynamics* of the fluctuations (important for the bandwidth) depend on the capacitances C and  $C_E$  which are tunable through the probe density of states  $\nu$  at the Fermi level.

## 1. Average quantities

While the average charge current  $\bar{I}_s$  is always zero (and so is the average electrochemical potential of the probe,  $\bar{\mu}_p = 0$ ), the average energy current  $\bar{I}_s^E$  from the source can be detected via the average probe temperature  $\bar{T}_p$ , see Eqs. (9) and (33). For this purpose, the difference between the probe temperature with and without the working source should be *maximized* for a fixed average source energy current. To meet this goal, the probe thermal conductance  $\kappa_{ph,0}$  due to the coupling to phonons should be small with respect to the electronic thermal conductance  $\kappa_{e,0}$ .

#### 2. Magnitude of source fluctuations

An increase of the probe temperature  $\overline{T}_p$ , in order to improve its detectability is naturally detrimental for the readout of the charge- and heat-current fluctuations, because it leads to a relative increase of the *induced* noise,  $\mathcal{P}_{s,\text{ind}}^{II}$  and  $\mathcal{P}_{s,\text{ind}}^{JJ}$ . From this point of view, the ratio  $\kappa_{e,0}/\kappa_{\text{ph},0}$ should be *minimized*. However, the overall prefactor of the fluctuations also depends on this quantity and needs to be optimized.

In the case of the charge-current fluctuations  $\mathcal{P}_p^{II}$ , the QPC transmission *D* entering the electronic thermal conductivity  $\kappa_{e,0}$  not only needs to be small to ensure that the magnitude of the direct part  $\mathcal{P}_{s,\text{dir}}^{II}$  is maximized with respect to the induced part  $\mathcal{P}_{s,\text{ind}}^{II}$ , but also to increase the overall value of the desired potential fluctuations, see the prefactors in both Eqs. (38) and (39).

On the other hand, for the detection of the heat-current noise  $\mathcal{P}_p^{JJ}$ , the total thermal conductance  $\kappa$  at the probe temperature needs to be small in order to maximize the overall value of  $\mathcal{P}_p^{TT}$ . The interplay of the requirement for a large  $\kappa_{ph,0}$  to reduce the induced part of the noise with respect to the direct one, together with the need for a small  $\kappa$  to maximize the total magnitude of the fluctuations, is crucial from the point of view of the detection-scheme optimization. Thus, in Fig. 6, we present the behavior of the thermal conductance at the probe temperature,  $\kappa$ , as a function of the ratio  $\kappa_{e,0}/\kappa_{ph,0}$  for different energies of the emitted particles  $\langle \epsilon \rangle/T$  (influencing the probe temperature). The explicit ratio between  $\kappa$  and the basetemperature thermal conductance  $\kappa_0 = \kappa_{e,0} + \kappa_{ph,0}$  can be written as

$$\frac{\kappa}{\kappa_0} = \left[\frac{\kappa_{e,0}}{\kappa_{\rm ph,0}} \frac{\overline{T}_p}{T_0} + \left(\frac{\overline{T}_p}{T_0}\right)^4\right] \left[\frac{\kappa_{e,0}}{\kappa_{\rm ph,0}} + 1\right]^{-1}, \quad (44)$$

using the general, temperature-dependent expressions given in Eq. (1). It shows that smaller values of the total thermal conductance  $\kappa$  can be achieved by augmenting the phonon conductance  $\kappa_{ph,0}$  with respect to the electronic heat conductance  $\kappa_{e,0}$ , that is, when  $\kappa_{e,0}/\kappa_{ph,0} \ll 1$ .



FIG. 6. Thermal conductance at probe temperature  $\kappa$ , scaled to its value  $\kappa_0$  at  $T_0$ , shown as a function of the ratio of the electron ( $\kappa_{e,0}$ ) and phonon ( $\kappa_{ph,0}$ ) thermal conductances,  $\kappa_{e,0}/\kappa_{ph,0}$ , at base temperature  $T_0$ . We show curves for selected values of  $\langle \epsilon \rangle / (\mathcal{T}gLT_0^2)$ —that is, the average energy of an emitted particle per period,  $\langle \epsilon \rangle / \mathcal{T}$ , related to the electronic heat-current response of the probe,  $gLT_0^2$ .

However, the function saturates, and even decreases again, when  $\kappa_{ph,0}$  is small with respect to  $\kappa_{e,0}$ .

## 3. Bandwidth of noise detection

For an optimization of the bandwidth, not only the (thermal) conductances, but also the (thermal) capacitances play a role, providing an additional means for the tunability.

The bandwidth of the charge-current noise detection, improving with increasing  $1/\tau_{RC} = Dg/C$  [see Eq. (38)], can be optimized by increasing the conductance  $\propto D$  or by diminishing the geometric capacitance of the probe, while keeping *D* fixed at a value to find the highest achievable magnitude of  $\mathcal{P}_{s,\text{dir}}^{II}$ .

Similarly, a large thermal conductance  $\kappa$ , with its dependence on  $\kappa_{e,0}$  and  $\kappa_{ph,0}$  at  $T_0$  as shown in Fig. 6, is beneficial for the bandwidth of the detection of the heat-current noise,  $\propto 1/\tau_E = \kappa/C_E$ . This bandwidth can, however, also be increased by diminishing the heat capacity  $C_E$ , at a fixed optimized value of  $\kappa$ .

#### **VI. CONCLUSIONS**

In this paper we have proposed and analyzed an injector-detector setup, suitable for the measurement of charge- and heat- (or energy-) current noise injected by a time-dependently-driven electronic current source. For our detection scheme, we suggest to measure the (macroscopic) temperature and voltage fluctuations arising in a probe due to the operation of the source, in order to extract the charge- and energy-current fluctuations from them. Using a Boltzmann-Langevin approach together with an appropriate correlator decomposition scheme, we show how the frequency-dependent fluctuations of the macroscopic probe properties are related to the current correlators to be detected. We carefully analyze the backaction effects of the detection scheme on the signal to be extracted.

As a proof-of-principle example, we investigate, as the current source to be analyzed, an ideally time-dependentlydriven mesoscopic capacitor emitting a sequence of wellseparated quantized charge pulses. Based on this example, we show the behavior of the different contributions to the fluctuations as a function of the setup parameters and thereby indicate regimes in which a detection of heat-current fluctuations appears feasible. The experimental detection of recent theoretically investigated energetic properties of electronic single-particle sources [11–14] and (mixed) heat-current fluctuations more generally [17] is hence made tangible by our proposed setup.

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## APPENDIX A: TIME-DEPENDENT TRANSPORT THEORY

To calculate charge and energy currents, as well as respective noise correlators in a time-periodicallydriven mesoscopic conductor, we use a scattering matrix approach, which we briefly introduce here. In general, the properties of a time-dependently-driven mesoscopic conductor are described by the Floquet scattering matrix [31]. In the considered system, that is, for a two-terminal conductor supporting a single channel, the elements of the scattering matrix are  $S^{\rm F}_{\alpha\beta}(E_n, E_m)$ . They represent the amplitude to scatter from energy  $E_m$  in terminal  $\beta$  to energy  $E_n$  in terminal  $\alpha$ , with  $E_n = E + n\hbar\Omega$  and  $\alpha, \beta =$ L, p. As a result, the annihilation operators  $\hat{b}_{\alpha}(E)$  for outgoing particles at energy E in terminal  $\alpha$  can be related to the annihilation operators  $\hat{a}_{\beta}(E_n)$  for incoming particles at energy  $E_n$  in terminal  $\beta$  by means of the scattering matrix as follows:

$$\hat{b}_{\alpha}(E) = \sum_{\beta} \sum_{n=-\infty}^{\infty} S^{\mathrm{F}}_{\alpha\beta}(E, E_n) \hat{a}_{\beta}(E_n).$$
(A1)

The operators for the *charge*,  $\hat{I}_{\alpha}(t)$ , and *energy*,  $\hat{I}_{\alpha}^{E}(t)$ , currents in terminal  $\alpha$  can be written in terms of the corresponding annihilation and creation operators as

$$\hat{I}_{\alpha}(t) = -\frac{e}{h} \iint dE dE' e^{i(E-E')t/\hbar} \\ \times \{\hat{b}_{\alpha}^{\dagger}(E)\hat{b}_{\alpha}(E') - \hat{a}_{\alpha}^{\dagger}(E)\hat{a}_{\alpha}(E')\},$$
(A2)

and

$$\hat{I}^{E}_{\alpha}(t) = \frac{1}{h} \iint dE dE' \frac{(E+E')}{2} e^{i(E-E')t/\hbar} \\ \times \{\hat{b}^{\dagger}_{\alpha}(E)\hat{b}_{\alpha}(E') - \hat{a}^{\dagger}_{\alpha}(E)\hat{a}_{\alpha}(E')\}.$$
(A3)

The *heat*-current operator can then be obtained from

$$\hat{J}_{\alpha} = \hat{I}^{E}_{\alpha} + \overline{\mu}_{\alpha}\hat{I}_{\alpha}/e.$$
 (A4)

Employing the current operators in Eqs. (A2) and (A3) together with the operator relation in Eq. (A1), the charge and energy currents averaged over one period of the driving are derived along standard lines, see e.g., Ref. [31],

$$\bar{I}_{\alpha} = -\frac{e}{h} \int dE[\tilde{f}_{\alpha}(E) - f_{\alpha}(E)], \qquad (A5)$$

$$\bar{I}^E_{\alpha} = \frac{1}{h} \int dE \, E[\tilde{f}_{\alpha}(E) - f_{\alpha}(E)], \qquad (A6)$$

with  $f_{\alpha}(E) = \{1 + \exp[(E - \overline{\mu}_{\alpha})/(k_B T_{\alpha})]\}^{-1}$  denoting the Fermi distribution in the terminal  $\alpha$ , and

$$\widetilde{f}_{\alpha}(E) = \sum_{\beta} \sum_{n=-\infty}^{\infty} |S_{\alpha\beta}^{\mathrm{F}}(E, E_n)|^2 f_{\beta}(E_n).$$
(A7)

On the other hand, the low-frequency, period-averaged correlations between currents of type A, B = I, J in terminal  $\alpha$  are defined as

$$\mathcal{P}_{\alpha}^{AB} = \frac{1}{2} \int_{0}^{\mathcal{T}} \frac{dt}{\mathcal{T}} \int_{-\infty}^{\infty} dt' \langle \{ \delta \hat{A}_{\alpha}(t), \delta \hat{B}_{\alpha}(t+t') \} \rangle, \quad (A8)$$

where  $\delta \hat{A}_{\alpha}(t) \equiv \hat{A}_{\alpha}(t) - \overline{A}_{\alpha}$ . Following the standard procedure outlined, e.g., in Refs. [12,13,31], we derive the expression for charge-, energy- and mixed-current autocorrelations, conveniently written in a general form as

$$\mathcal{P}_{\alpha}^{AB} = \frac{1}{h} \int dE \left\{ x_{A} x_{B} f_{\alpha}(E) [1 - f_{\alpha}(E)] - \sum_{n = -\infty}^{\infty} (x_{A} x_{B,n} + x_{A,n} x_{B}) |S_{\alpha\alpha}^{F}(E, E_{n})|^{2} f_{\alpha}(E_{n}) [1 - f_{\alpha}(E_{n})] + \sum_{n = -\infty}^{\infty} \frac{x_{A} x_{B,n} + x_{A,n} x_{B}}{2} \sum_{\beta \gamma} \sum_{k,q = -\infty}^{\infty} S_{\alpha\beta}^{F*}(E, E_{k}) S_{\alpha\beta}^{F}(E_{n}, E_{k}) S_{\alpha\gamma}^{F}(E, E_{q}) S_{\alpha\gamma}^{F*}(E_{n}, E_{q}) f_{\beta}(E_{k}) [1 - f_{\gamma}(E_{q})] \right\}, \quad (A9)$$

where we have introduced a shorthand notation:  $x_{I,n} = x_I \equiv -e$  and  $x_{J,n} \equiv E_n - \overline{\mu}_p$  (with  $x_J \equiv x_{J,0}$ ).

#### **APPENDIX B: MESOSCOPIC CAPACITOR—INDUCED CORRELATIONS**

#### 1. Chemical potential correlators

To obtain the induced part  $\mathcal{P}_{s,\text{ind}}^{II}$ , we start from the full expression for the charge-current correlator (36), and we omit the direct part  $\mathcal{P}_{s,\text{dir}}^{II}$ . The remaining correlator  $\mathcal{F}_{s,\text{ind}}^{II} \equiv \mathcal{P}_{s,\text{ind}}^{II}/\mathcal{P}_{0}^{II}$  (for convenience divided by the equilibrium thermal noise of the QPC,  $\mathcal{P}_{0}^{II} = 2Dgk_{B}T_{0}$ ) can then be written as

$$\mathcal{F}_{s,\text{ind}}^{II} = \frac{1-D}{2k_B T_0} \int dE \sum_{n=-\infty}^{\infty} |S_n|^2 [f_p(E) - f_L(E)] [1 - 2f_L(E_n)] + \frac{D}{2k_B T_0} \int dE \{f_p(E)[1 - f_p(E)] - f_L(E)[1 - f_L(E)]\}.$$
(B1)

Performing the last integral, and inserting the explicit expression for the frozen scattering matrix  $S_n$ , Eq. (31), we find

$$\mathcal{F}_{s,\text{ind}}^{II} = -\frac{1-D}{k_B T_0} (2\Omega\sigma)^2 \int dE[f_p(E) - f_L(E)] \sum_{n>0} e^{-2n\Omega\sigma} [f_L(E_n) + f_L(E_{-n})] + \frac{D(\overline{T}_p - T_0)}{2T_0} + (1-D)\mathcal{H}_0\left(\frac{\overline{T}_p}{T_0}\right).$$
(B2)

Making use of the adiabatic and low-temperature conditions given by Eq. (28), the first term of the equation above can be expanded up to leading order in  $\Omega \sigma \ll 1$ . One finds then that this term is negligibly small compared to the remaining two terms in the last line of Eq. (B2). With this, Eq. (40) from the main text is obtained.

## 2. Temperature correlators

The induced part  $\mathcal{P}_{s,\text{ind}}^{JJ}$  of the heat-current correlator  $\mathcal{P}_{p}^{JJ}$  can be derived in an analogous manner, as discussed in the section above. Again, starting from the full expression for the correlator (36), the normalized induced component  $\mathcal{F}_{s,\text{ind}}^{JJ} \equiv \mathcal{P}_{s,\text{ind}}^{JJ} / \mathcal{P}_{0}^{JJ}$  [with  $\mathcal{P}_{0}^{JJ} = 2D\pi^{2}(k_{\text{B}}T_{0})^{3}/(3h)$  representing the equilibrium heat noise of the QPC] can be written as

$$\mathcal{F}_{s,\text{ind}}^{JJ} = \frac{3(1-D)}{2\pi^2 (k_B T_0)^3} \int dE E^2 \sum_{n=-\infty}^{\infty} |S_n|^2 [f_p(E) - f_L(E)] [1 - 2f_L(E_n)] + \frac{3D}{2\pi^2 (k_B T_0)^3} \int dE E^2 \{f_p(E) [1 - f_p(E)] - f_L(E) [1 - f_L(E)]\}.$$
(B3)

Performing integrals and rearranging terms similarly as in Eq. (B2), we get

$$\mathcal{F}_{s,\text{ind}}^{JJ} = -\frac{3(1-D)}{\pi^2 (k_B T_0)^3} \int dE \, E^2 [f_p(E) - f_L(E)] \sum_{n>0} e^{-2n\Omega\sigma} [f_L(E_n) + f_L(E_{-n})] + \frac{D(\overline{T}_p^3 - T_0^3)}{2T_0^3} + \frac{3(1-D)}{\pi^2} \mathcal{H}_2\left(\frac{\overline{T}_p}{T_0}\right). \tag{B4}$$

Finally, using the low-temperature adiabatic expansion of the first term in the equation above, and consequently retaining only the last two terms, we obtain Eq. (43) in the main text.

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- [54] The Floquet matrix is unitary, implying the relation  $\sum_{n=-\infty}^{\infty} S^*(E_n, E_m)S(E_n, E_q) = \delta_{mq}$ , which here is required to hold up to the first order in  $\sigma \Omega$ , only.
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