Constrained multiobjective shape optimization of superconducting rf cavities considering robustness against geometric perturbations

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High current storage rings, such as the Z-pole operating mode of the FCC-ee, require accelerating cavities that are optimized with respect to both the fundamental mode and the higher order modes. Furthermore, the cavity shape needs to be robust against geometric perturbations which could, for example, arise from manufacturing inaccuracies or harsh operating conditions at cryogenic temperatures. This leads to a constrained multiobjective shape optimization problem which is computationally expensive even for axisymmetric cavity shapes. In order to decrease the computation cost, a global sensitivity analysis is performed and its results are used to reduce the search space and redefine the objective functions. A massively parallel implementation of an evolutionary algorithm, combined with a fast axisymmetric Maxwell eigensolver and a frequency-tuning method is used to find an approximation of the Pareto front. The computed Pareto front approximation and a cavity shape with desired properties are shown. Further, the approach is generalized and applied to another type of cavity.

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I. INTRODUCTION

Accelerating cavities are metallic chambers with a resonating electromagnetic field that are used to impart energy to charged particles in many particle accelerators. Over the last few decades, a lot of research has been carried out on improving the shape and material properties of accelerating cavities. Many attractive features of superconducting radio frequency (rf) cavities, such as high intrinsic quality factor, have made them a favorable choice in applications where high continuous wave voltage is required [1]. The shape of the cavity, on the other hand, determines many figures of merit, such as the normalized peak electric and magnetic field on the surface of the cavity $(E_{\rm pk}/E_{\rm acc})$ and $B_{\rm pk}/E_{\rm acc}$, respectively), geometric shunt impedance (R/Q), cell-to-cell coupling, etc. Therefore,

Published by the American Physical Society under the terms of the Creative Commons Attribution 4.0 International license. Further distribution of this work must maintain attribution to the author(s) and the published article's title, journal citation, and DOI. based on the requirements of each specific accelerator, the shape of the cavity needs to be carefully optimized.

In many accelerators it is desirable to achieve a high accelerating field ($E_{\rm acc}$) in order to improve efficiency and save on equipment cost. In superconducting rf cavities the maximum achievable $E_{\rm acc}$ is limited by the maximum electric and magnetic field on the surface of the cavity. Thus, many cavity optimization methods have focused on minimizing $E_{\rm pk}/E_{\rm acc}$ and $B_{\rm pk}/E_{\rm acc}$ in order to provide room for increasing $E_{\rm acc}$ [2–6]. In addition to the maximum achievable $E_{\rm acc}$, the surface losses of the cavity should be minimized, which can be achieved by maximizing $G \cdot R/Q$ (where G is the geometry factor). It was shown in [2] that high $G \cdot R/Q$ typically goes along with low $B_{\rm pk}/E_{\rm acc}$, so either of the two can be considered in the optimization. Since various conflicting objective functions are usually optimized simultaneously, a trade-off between these objective functions has to be considered. In [2,3] $B_{\rm pk}/E_{\rm acc}$ is minimized, or $G \cdot R/Q$ maximized, while other figures of merit such as $E_{\rm pk}/E_{\rm acc}$, the wall slope angle, and the aperture radius of the cavity are kept fixed. In [4] a Pareto front between $E_{\rm pk}/E_{\rm acc}$ and $B_{\rm pk}/E_{\rm acc}$ was obtained by analyzing more than a thousand geometries. A geometry was then selected from this Pareto front based on the machine requirements (high gradient applications favor a smaller $E_{\rm pk}/E_{\rm acc}$, while low loss applications favor a smaller $B_{\rm pk}/E_{\rm acc}$).

The optimization methods proposed in the literature mainly focus on optimizing the inner cells of a multicell

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cavity. The end half-cells are then optimized to get high field flatness in the cavity and ease the higher order mode (HOM) damping [2]. The optimization of single-cell cavities is slightly different than that of multicell cavities. The length of the inner cells of a multicell cavity is usually fixed to $\beta \lambda/2$, where λ is the wavelength of the fundamental mode (FM), and β is the ratio of the particle velocity to the speed of light ($\beta \approx 1$ for particle velocities close to the speed of light). This restriction improves acceleration, since, as a particle traverses a cell, the direction of the electric field changes and the particle receives a force in the same direction. In single-cell cavities, the particle passes through only one cell, so there is no such restriction on the length of the cavity. Consequently, in single-cell optimization there is one more degree of freedom. Furthermore, since part of the field leaks into the beam pipe, in single-cell optimization the cell and the beam pipe have to be simulated together. This is in contrast to the optimization of the inner cells of multicell cavities, where only one cell, with appropriate boundary conditions, is considered.

Many future accelerators, such as the Future Circular Collider (FCC), aim at colliding beams with unprecedented luminosities [7]. The FCC design study includes three colliders: a hadron collider (FCC-hh), a lepton collider (FCC-ee), and a lepton-hadron collider. The aim of FCC-ee is to study properties of Z, W and Higgs bosons, as well as top quark, with collision energies ranging from 90 GeV to 365 GeV. The high luminosity requirements demand a significant increase in the beam current of the machine, which in turn increases the HOM power deposited into the cavities by the traversing beam. Preliminary studies have suggested using single-cell cavities for FCC-hh and the Zpole operating mode of FCC-ee (FCC-ee-Z) [8,9]. In the case of FCC-ee-Z, the main reasons for selecting a singlecell cavity are the issues related to the beam instability and high HOM power. In the FCC-ee-Z, the operating $E_{\rm acc}$ is a few MV/m [8–10]. Reaching higher $E_{\rm acc}$ is precluded due to limitations on the fundamental power coupler in providing high input power per cavity (e.g., for FCC-ee-Z this is around 1 MW per cavity). In such low $E_{\rm acc}$ and highcurrent operations, minimizing surface peak fields should not be the primary goal. Instead, other figures of merit, in particular the HOM aspects, have to be taken into account right from the early design stages of the cavity. Enlarging the beam pipe radius is a common approach to untrap many HOMs and reduce the loss factor. However, even with an enlarged beam pipe radius, the first dipole band usually stays trapped and cannot propagate out of the cavity [11]. Additionally, the modes in the first dipole band usually have a large transverse impedance, which can cause transverse beam instability. Therefore, and in order to simplify their damping via HOM couplers, special attention should be given to the modes in the first dipole band.

In addition to the rf properties of the cavity, the cavity shape needs to be robust against geometric perturbations [12–15], which could, for example, arise from harsh operating conditions at cryogenic temperatures or manufacturing inaccuracies. Most importantly, the frequency of the FM has to remain at its nominal value in order to avoid excessive power being fed into the cavity for maintaining $E_{\rm acc}$, and the frequencies of the modes in the first dipole band should not be very sensitive to geometric changes in order to avoid unwantedly hitting a beam spectral line or harming their damping when coaxial HOM couplers are used.

The major contribution of this paper lies in finding a robust cavity shape while several properties of the FM and the first dipole band are optimized at the same time. For this purpose an optimization method for constrained multiobjective shape optimization of superconducting rf cavities is proposed. The focus of optimization is on axisymmetric single-cell cavities used in high-current accelerators, e.g., FCC-ee-Z, which unlike the multicell cavities have not been extensively studied before. As a part of the proposed optimization approach, a global sensitivity analysis is carried out in order to quantify the relative influence of each of the geometric parameters on the figures of merit of the cavity. The novel approach makes use of the most influential parameters, reduces the search space by discarding noninfluential parameters, and provides valuable insights into the optimization of rf cavities.

The structure of the paper is as follows. The parametrization of the cavity cross section and the quantities of interest are described in Sec. II. The results of the sensitivity analysis are shown in Sec. III, and used for search space reduction in Sec. IV. The constrained multiobjective optimization problem is defined in Sec. V, the algorithm described in Sec. VI, and the results presented in Sec. VII. The approach described in Secs. II—VI is generalized to a different type of cavity in Sec. VIII. Conclusions are drawn in Sec. IX.

II. QUANTITIES OF INTEREST

Elliptical cavities are a commonly used cavity shape for the acceleration of particles with $\beta \approx 1$. As shown in Fig. 1,

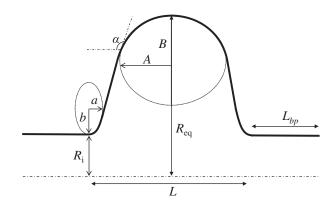


FIG. 1. The parametrization of the cross section of a single-cell elliptical cavity. The variables R_{eq} , L, A, B, a, b, and R_i parametrize the cell, and $L_{\rm bp}$ is the beam pipe length.

the cross section of the cell of such cavities can be parametrized by seven variables, i.e., geometric parameters: R_{eq} , L, A, B, a, b, and R_i . The wall slope angle α is then determined by these geometric parameters. The beam pipe length $L_{\rm bp}$ is not an optimization variable. It is shown here to fully depict the simulation setup that was considered in the optimization.

In constrained multiobjective shape optimization the aim is to find the geometric parameters that best satisfy the given objectives and constraints. The objectives and constraints that need to be satisfied are determined by the requirements of the accelerator. The first constraint is to tune the frequency of the FM ($f_{\rm FM}$), which is typically the TM₀₁₀ mode, to a desired value. Focusing on FCC-ee-Z, in this paper $f_{\rm FM}$ is tuned to 400.79 MHz, which is the proposed value for both FCC-ee-Z and FCC-hh. In order to have a sufficient distance for decaying the FM leaked into the beam pipe, the beam pipe length $L_{\rm bp}$ is set to the value of the wave length, i.e., to $\lambda=748$ mm. The second constraint is $\alpha[{\rm degree}] \geq 90$ in order to avoid reentrant cavity shapes (due to the problems associated with chemical treatment).

In comparison with the middle-cells of multicell cavities, the electric field concentration around the iris region of single-cell cavities is lower because part of the electric field leaks into the beam pipe. Due to this reason and relatively low $E_{\rm acc}$ of FCC-ee-Z, the surface peak fields are not the primary concern in the optimization. Lowering the surface losses helps to decrease the power released into the helium bath, which consequently reduces the amount of power required to maintain the cryogenic temperature. The surface losses of the FM are given by

$$P_c = \frac{(V_{\parallel(r=0)})^2}{G \cdot R/O} R_s,\tag{1}$$

where R_s is the surface resistance of the cavity, which depends on its material properties, $V_{||(r=0)}$ the longitudinal voltage calculated along the longitudinal axis, and G and R/Q the geometry factor and the geometric shunt impedance of the FM, respectively. Both G and R/Q depend on the shape of the cavity [5]. Equation (1) indicates that P_c can be minimized by maximizing $G \cdot R/Q$ [the corresponding objective function is labeled as F_4 in Eq. (3)].

In addition to the properties of the FM, the properties of the HOM spectrum also have to be taken into account. The longitudinal loss factor is inversely proportional to the aperture radius R_i of the cavity [[16] p. 372]. Therefore, increasing the aperture radius lowers the longitudinal loss factor and consequently the HOM power deposited into the cavity by the beam. Larger aperture radius also helps to untrap the dangerous higher order monopole modes. As shown in Fig. 2, an aperture radius roughly above 145 mm untraps the monopole modes which typically have a large longitudinal geometric shunt impedance, i.e., the TM_{011}

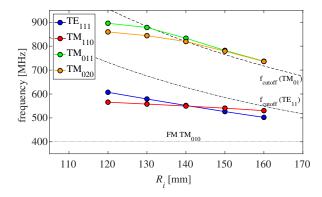


FIG. 2. The dependency of the frequency of the TE_{111} , TM_{110} , TM_{011} and TM_{020} modes on R_i . For each case, the frequency of the FM, i.e., the TM_{010} mode is tuned to 400.79 MHz by varying R_{eq} . The parameter sweep is carried out around the geometric parameters taken from [17]. An aperture radius R_i approximately above 145 mm helps to untrap the dangerous monopole HOMs. Note that the cavity in [17] is optimized such that the frequencies of the TE_{111} and TM_{110} modes are almost equal.

and TM_{020} modes. Therefore, a larger aperture radius is preferable (in Eq. (3), this is denoted by F_5). The first dipole band, however, stays trapped in the cavity, in particular the TE_{111} mode whose frequency approaches the FM with enlarged R_i .

The damping of the HOMs is usually done using coaxial couplers, waveguide (WG) couplers, beam pipe absorbers, or their combination. Coaxial and WG HOM couplers provide better damping of trapped modes as they can be placed close to the cell. In order to allow the propagation of the HOMs into the WG, the cutoff frequency of the first mode of the WG (the TE₀₁ mode) has to be between the FM and the first dipole mode. If the frequency of the first dipole mode is very close to $f_{\rm FM},$ a WG coupler with larger dimensions is required, which occupies more space in the cryomodule, in particular at 400.79 MHz. In that case the WG has to be wider in order to decrease the cutoff frequency of its TE₀₁ mode, and also longer in order to have a sufficient distance for decaying the FM leaked into the WG. Therefore, as another objective function, the distance between $f_{\rm FM}$ and the frequency of the first dipole mode (f_1) , which is typically the TE_{111} mode, has to be maximized. Since f_1 is larger than $f_{\rm FM}$, this is equivalent to minimizing the negative value $f_{\rm FM} - f_1$ (in Eq. (3), the corresponding objective function is labeled as F_1).

Coaxial HOM couplers act like 3D resonant circuits that are optimized to have a notch at the FM and resonances (with a high transmission) at certain frequencies that require strong damping, such as the frequency of the TE_{111} and TM_{110} modes. A smaller difference between the frequencies of the first two dipole modes simplifies using a narrow band coaxial HOM coupler for damping them (such as the Hook-type coupler used in LHC cavities [18–20]). Therefore, another objective is to minimize the difference between the frequencies of the two trapped

dipole modes [this is denoted by F_2 in Eq. (3)]. Additionally, the sum of the transverse impedances of the first two dipole modes is minimized [F_3 in Eq. (3)]. The following definition of the transverse impedance is used

$$rac{R}{Q_{\perp}} = rac{(V_{||(r=r_0)} - V_{||(r=0)})^2}{k^2 r_0^2 \omega U},$$

where k is the wave number, r_0 the offset from the axis, ω the angular frequency, and U the stored energy.

The optimized cavity should be robust against geometric changes that might arise due to manufacturing inaccuracies or perturbations during operation. If the FM is detuned from its nominal value, additional power has to be fed into the cavity to maintain the same $E_{\rm acc}$. Thus, tuners are used to adjust the frequency to the desired value. Frequency tuning is usually carried out by changing the length of the cells via applying a longitudinal force to the cavity [[1] p. 431]. A large tuning force should be avoided as it might plastically deform the cavity. Therefore, the cavities have to be fabricated with a high precision to avoid large tunings afterwards. On the other hand, a cavity fabricated with a higher precision is more expensive to fabricate and handle. As shown in Fig. 3, different cavity shapes can have different sensitivities with respect to geometric perturbations. E.g., cavities with a larger wall slope angle have a less sensitive $f_{\rm FM}$ against changes in R_{eq} (up to around 30% difference between different shapes).

The frequency of the trapped dipole modes could also vary due to perturbations or during the tuning of the frequency of the FM. Figure 4 shows the local sensitivity of the frequency of the TM₁₁₀ mode with respect to R_i against the geometric variable A. Sensitivity of the TM₁₁₀ mode against changes in R_i can vary up to a factor of seven between different shapes. A change in the frequency of the dipole modes could harm

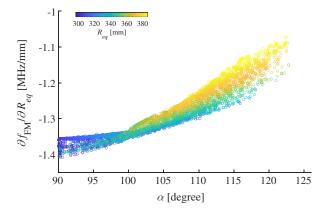


FIG. 3. The local sensitivity of $f_{\rm FM}$ with respect to R_{eq} plotted versus the wall slope angle α . Around 5'000 samples of cavities with different values of R_{eq} , L, A, B, a, b and R_i are studied. The frequency of each sample is tuned to 400.79 MHz. The local sensitivity is calculated using the forward difference method with the step h=1 mm. Generally, a higher wall slope angle yields a less sensitive $f_{\rm FM}$ against changes in R_{eq} .

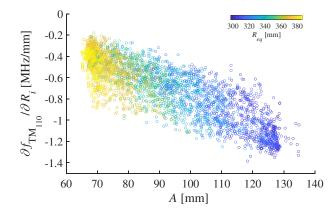


FIG. 4. The local sensitivity of the frequency of the TM_{110} mode with respect to R_i plotted versus A. Around 5'000 samples of cavities with different values of R_{eq} , L, A, B, a, b and R_i are studied, and the frequency of the FM of each sample is tuned to 400.79 MHz.

their damping when coaxial HOM couplers are used. Coaxial HOM couplers are usually tuned to have a resonance at the frequency of the targeted HOMs and a mismatch between the frequency of the HOM and the coaxial HOM coupler could result in a poor damping of that mode. Therefore, in addition to the robustness of $f_{\rm FM}$, the robustness of the frequencies of the first two dipole modes is of importance. In Eq. (3), the objective functions corresponding to the minimization of the local sensitivities of $f_{\rm FM}$, $f_{\rm TE_{111}}$ and $f_{\rm TM_{110}}$ against geometric changes are denoted by \hat{F}_6 , \hat{F}_7 and \hat{F}_8 , respectively.

To summarize, denoting by f_2 the frequency of the second dipole mode (which is typically the TM_{110} mode) and

$$\mathbf{d} = (d_1, ..., d_7) = (R_{eq}, R_i, L, A, B, a, b), \tag{2}$$

the constrained multiobjective optimization problem considered in this paper can be written as

$$\begin{split} & \min_{d} \left(\underbrace{f_{\text{FM}} - f_{1}}_{F_{1}}, \underbrace{|f_{1} - f_{2}|}_{F_{2}}, \underbrace{\frac{R}{Q_{\perp 1}} + \frac{R}{Q_{\perp 2}}}_{F_{3}}, \underbrace{-G \cdot \frac{R}{Q}}_{F_{4}}, \underbrace{-R_{i}}_{F_{5}}, \\ & \underbrace{\sum_{j} \left|\frac{\partial f_{\text{FM}}}{\partial d_{j}}\right|}_{\hat{F}_{6}}, \underbrace{\sum_{j} \left|\frac{\partial f_{\text{TE}_{111}}}{\partial d_{j}}\right|}_{\hat{F}_{7}}, \underbrace{\sum_{j} \left|\frac{\partial f_{\text{TM}_{110}}}{\partial d_{j}}\right|}_{\hat{F}_{8}}\right), \\ & \text{subject to} \quad f_{\text{FM}} = 400.79 \, \text{MHz} \quad \text{and} \quad \alpha[\text{degree}] \geq 90. \end{split} \tag{3}$$

In this formulation the local sensitivities of the FM and the dipole modes are calculated with respect to all seven geometric parameters, which entails a very expensive computation. In the following section a global sensitivity analysis is carried out in order to find the most influential geometric parameters on each of the frequencies. This information is then used to redefine objective functions \hat{F}_6 ,

 \hat{F}_7 and \hat{F}_8 , reduce the search space and, consequently, reduce the computational cost of the problem.

III. SENSITIVITY ANALYSIS

In order to determine which geometric parameters have the greatest influence on $f_{\rm FM}$, $f_{\rm TE_{111}}$ and $f_{\rm TM_{110}}$, a variancebased global sensitivity analysis is performed [21,22]. The geometric parameters are considered to be independent, uniformly distributed random variables. The first-order Sobol' indices and total Sobol' indices [23], representing the individual and total influences of these random variables on the variance of the quantities of interest (QoIs), are computed using polynomial chaos (PC) expansion [22,24–27]. The coefficients of the PC expansion are obtained nonintrusively, because nonintrusive methods allow the use of an existing solver as a black box, requiring only an evaluation of the QoIs in a set of either deterministic or random points. Since some of the deterministic points may correspond to infeasible cavity shapes, a random sample, i.e., a collection of random points, is used. According to [24], it is enough to use a sample of size

$$(N-1)\frac{(N+p)!}{N! \cdot p!},\tag{4}$$

where N is the number of geometric parameters and p the polynomial degree used. The values of Eq. (4), for a few relevant values of N and p, are given in Table I.

For this analysis, the Uncertainty Quantification Toolkit (UQTk) [25,28] is used. The random sample is evaluated in parallel, taking into account only feasible cavity shapes with $\alpha[\text{degree}] \geq 90$. The changes in the sensitivity plots obtained for p=3 (which needs 720 random points) are almost imperceptible, so only the case p=2 (216 random points) is shown. The main sensitivities (i.e., first-order Sobol' indices), with p=2 and with geometric parameters in the intervals shown in s II and III, are shown in Figs. 5 and 6, respectively.

These plots show the influence of the geometric parameters [cf. Eq. (2)] on the frequencies

TABLE I. Eq. (4) for a few relevant values of N and p.

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p	4	5	6	7
2	45	84	140	216
3	105	224	420	720

TABLE II. Wide intervals for the geometric parameters.

Variable [mm]	R_{eq}	R_i	L	A	В	а	b
Lower bound	325	145	240	65	65	10	10
Upper bound	375	160	380	140	140	60	60

TABLE III. Narrow intervals for the geometric parameters.

Variable [mm]	R_{eq}	R_i	L	A	В	а	b
Lower bound	345	145	280	70	70	45	45
Upper bound	355	155	300	80	80	55	55

$$f_{\rm FM}, f_{\rm TE_{111}}, f_{\rm TM_{110}},$$

but also on other QoIs

$$\frac{R}{Q}, \frac{R}{Q_{\perp, \mathrm{TE}_{111}}}, \frac{R}{Q_{\perp, \mathrm{TM}_{110}}}, G \cdot \frac{R}{Q}, \alpha, \frac{E_{\mathrm{pk}}}{E_{\mathrm{acc}}}, \frac{B_{\mathrm{pk}}}{E_{\mathrm{acc}}}.$$

Sobol' indices are, by definition, normalized with respect to the total variance, so they (the first-order and higher-order indices) sum up to 1. Consequently, the sum of the first-order indices, representing the individual influences of the parameters, is at most 1. The higher-order indices represent mixed influences of the parameters, so the fact that the sum of the first-order indices, i.e., the height of the bars in the plots, is close to 1 indicates a low correlation between parameters.

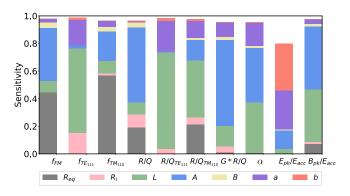


FIG. 5. Main sensitivities. The intervals for the geometric parameters R_{eq} , R_i , L, A, B, a, b are given in Table II. The polynomial degree is p = 2, so the size of the sample is 216 [cf. Table I].

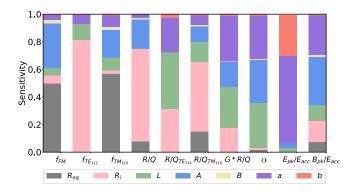


FIG. 6. Main sensitivities. The intervals for the geometric parameters R_{eq} , R_i , L, A, B, a, b are given in Table III. The polynomial degree is p = 2, so the size of the sample is 216 [cf. Table I].

The geometric parameter R_{eq} has the greatest influence on the frequency of the FM, $f_{\rm FM}$, (closely followed by A), so it is used to tune $f_{\rm FM}$ to 400.79 MHz, i.e., to enforce the first constraint [cf. Eq. (3)]. This is explained in detail in the next section.

IV. SEARCH SPACE REDUCTION

According to Figs. 5 and 6, R_{eq} has the greatest influence on $f_{\rm FM}$. Therefore, it is possible to use R_{eq} to tune the frequency to 400.79 MHz. Specifically, for a point

$$(d_2,...,d_7) = (R_i, L, A, B, a, b)$$

[cf. Eq. (2)], $d_1 = R_{eq} [\mathrm{mm}] \in [d_{\mathrm{lower}}, d_{\mathrm{upper}}]$ is found such that

$$f_{\rm FM}(d) = 400.79 \text{ MHz}.$$

The QoIs are then computed for the cavity corresponding to d (the details are given in Sec. VIB and Algorithm 2). The main sensitivities of the QoIs and R_{eq} with respect to $d_2, ..., d_7$, considering the wide (Table II) and narrow (Table III) intervals are shown in Figs. 7 and 8, respectively. The polynomial degree is again p=2, but the number of variables is now 6, so only 140 training points are needed.

It can be seen in Figs. 5–8 that the influence of B on all of the QoIs is very low, and that b significantly influences only $E_{\rm pk}/E_{\rm acc}$. Therefore, since $E_{\rm pk}/E_{\rm acc}$ is not part of any objective function in the optimization problem [cf. Eq. (3)], these two geometric parameters can be omitted. A natural way to do this is to set B=A and b=a, i.e., to consider circles instead of ellipses in the geometric parametrization of the cavity cross section [cf. Fig. 1].

V. CONSTRAINED MULTIOBJECTIVE OPTIMIZATION PROBLEM (CMOOP)

Based on the information from Figs. 5 and 6,

$$\frac{\partial f_{\mathrm{FM}}}{\partial R_{eq}}, \frac{\partial f_{\mathrm{FM}}}{\partial A}, \quad \mathrm{and} \quad \frac{\partial f_{\mathrm{TM}_{110}}}{\partial R_{eq}}$$

need to be taken into account. Similarly, from the information shown in Sec. IV (Figs. 7 and 8),

$$\frac{\partial f_{\text{TE}_{111}}}{\partial L}, \frac{\partial f_{\text{TE}_{111}}}{\partial R_i}, \frac{\partial f_{\text{TM}_{110}}}{\partial A}, \text{ and } \frac{\partial f_{\text{TM}_{110}}}{\partial R_i}$$

need to be considered as well. However, due to the influence of A on R_{eq} (Figs. 7 and 8) and their geometric connection (Fig. 1), in order to decrease the computation cost the local sensitivities with respect to A are omitted.

Therefore, the constrained multiobjective optimization problem is the following [cf. Eq. (3)]

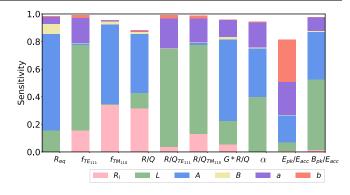


FIG. 7. Main sensitivities. For a point (R_i, L, A, B, a, b) inside the intervals from Table II, $f_{\rm FM}$ is tuned to 400.79 MHz using $R_{eq}[{\rm mm}] \in [325, 375]$. The polynomial degree is p=2, so the size of the sample is 140 [cf. Table I].

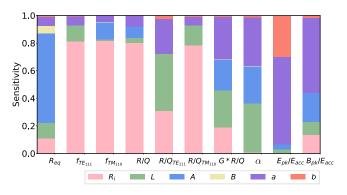


FIG. 8. Main sensitivities. For a point (R_i, L, A, B, a, b) inside the intervals from Table III, $f_{\rm FM}$ is tuned to 400.79 MHz using $R_{eq}[{\rm mm}] \in [345, 355]$. The polynomial degree is p=2, so the size of the sample is 140 [cf. Table I].

$$\begin{split} & \min_{R_i,L,A,a} \left(f_{\text{FM}} - f_1, |f_1 - f_2|, \frac{R}{Q_{\perp 1}} + \frac{R}{Q_{\perp 2}}, -G \cdot \frac{R}{Q}, -R_i, \\ & \underbrace{\left\lfloor \frac{\partial f_{\text{FM}}}{\partial R_{eq}} \right\rfloor}_{F_6}, \underbrace{\left\lfloor \frac{\partial f_{\text{TE}_{111}}}{\partial L} \right\rfloor + \left\lfloor \frac{\partial f_{\text{TE}_{111}}}{\partial R_i} \right\rfloor}_{F_7}, \underbrace{\left\lfloor \frac{\partial f_{\text{TM}_{110}}}{\partial R_{eq}} \right\rfloor + \left\lfloor \frac{\partial f_{\text{TM}_{110}}}{\partial R_i} \right\rfloor}_{F_8}, \end{split}$$

subject to $f_{\text{FM}} = 400.79 \text{ MHz}$ and $\alpha[\text{degree}] \ge 90$.

(5)

It is implied that, for each point $\mathbf{I} = (R_i, L, A, a)$, the values of B and b will be set to B = A and b = a, and that f_{FM} will be tuned to 400.79 MHz using R_{eq} as described in Sec. IV.

VI. CONSTRAINED MULTIOBJECTIVE OPTIMIZATION ALGORITHM

A. Forward solver

In order to compute the values of the objective functions from Eq. (5) in a point $\mathbf{d} = (R_{eq}, R_i, L, A, B, a, b)$ [cf. Eq. (2)] time-harmonic Maxwell's equations with perfectly electrically conducting (PEC) boundary

conditions (BC) are solved in the evacuated axisymmetric rf cavity parametrized by d. The mixed finite element method (FEM) leads to a generalized eigenvalue problem (GEVP) for each azimuthal mode number $m \in \mathbb{N}_0$ [29,30]. For monopole and dipole modes, m=0 and m=1, respectively. If the cross section of the cavity is symmetric, as is the case for the single-cell elliptical cavity from Fig. 1, it is sufficient to solve time-harmonic Maxwell's equations for one half of it, once with PEC and once with perfectly magnetically conducting (PMC) BC on the cross-sectional symmetry plane (BC_{SP}).

To compute the properties of the TM_{010} mode, the smallest eigenpair of the GEVP corresponding to m=0 and PEC BC_{SP} is found. This will be referred to as

$$MAXWELL_d(m, BC_{SP}) = MAXWELL_d(0, PEC).$$

In order to compute the properties of the TM_{110} and TE_{111} mode, the smallest eigenpair of the GEVPs corresponding to m=1 and PEC and PMC BC_{SP}, respectively, is found. This will be referred to as

 $MAXWELL_d(1, PEC)$ and $MAXWELL_d(1, PMC)$.

B. Optimization algorithm

Since the minimizers of different objective functions are usually different points, the concept of dominance is used: a point $I_1 = (R_{i,1}, L_1, A_1, a_1)$ dominates I_2 if it is not worse in any of the objectives, and it is strictly better in at least one objective. A point is called Pareto optimal if it is not dominated by any other point.

Because of conflicting objectives, the ability of an evolutionary algorithm (EA) to escape local optima, its suitability for parallelization, as well as good results in the area of particle accelerator physics [31–34] a multiobjective EA [35] is used to find an approximation of the set of Pareto optimal points, even though other methods, such as particle swarm optimization [36–38], ant colony optimization [39], simulated annealing [40], or artificial immune system [41] exist. The basic steps of an EA are shown in Algorithm 1.

Algorithm 1. Evolutionary algorithm. Geometric parameters $d_1,...,d_N, f_{\text{FM}}$ tuned using d_j , an individual is

$$I = (d_{i_1}, \ldots, d_{i_q}), \{i_1, \ldots, i_q\} \subset \{1, \ldots, N\} \setminus \{j\}.$$

- 1: random population of individuals $I_1, ..., I_M$
- 2: EVALUATE $(I_i, j), \forall i \in \{1, ..., M\}$
- 3: for a predetermined number of generations do
- 4: **for** pairs of individuals I_i , I_{i+1} **do**
- 5: crossover (I_i, I_{i+1}) , mutate (I_i) , mutate (I_{i+1})
- 6: for each new individual I_{new} , EVALUATE (I_{new}, j)
- 7: choose M fittest individuals for the next generation

In the case of the parametrization from Fig. 1, N=7, the geometric parameters d_1, \ldots, d_7 are [cf. Eq. (2)] R_{eq}, R_i, L , A, B, a, and b, and j=1, i.e., the frequency of the FM is tuned using $d_1=R_{eq}$. Since B and b are set to A and a, respectively, a design point for the CMOOP in Eq. (5), also called an individual in the context of an EA, is

$$I = (d_2, d_3, d_4, d_6) = (R_i, L, A, a).$$

The M individuals ($M \in \mathbb{N}$) comprising the first generation are chosen randomly from the given intervals (Algorithm 1: line 1). In this paper, in the case of the single-cell elliptical cavity, these intervals are given in Table II. These M individuals are then evaluated (1:2), i.e., the corresponding objective function values are computed, as shown in Algorithm 2, in the following way.

Algorithm 2. EVALUATE (I, j)**In:** $j \in \{1, ..., N\}, I = (d_{i_1}, ..., d_{i_n}),$ where

$$\{i_1, ..., i_q\} \subset \{1, ..., N\} \setminus \{j\}$$

Out: $F(d = (d_1, ..., d_N))$ s.t. $f_{FM}(d) = 400.79$ MHz

1: use

$$MAXWELL_d(0, PEC)$$

to compute $f_{\rm FM}(\boldsymbol{d})$ and TOMS 748 [42] algorithm to find

$$d_j \in [d_{\text{lower}}, d_{\text{upper}}]$$

s.t.

$$f_{\rm FM}(\mathbf{d})[{\rm MHz}] - 400.79 = 0$$

- 2: if $\alpha(d)$ [degree] < 90, **return**
- 3: compute the properties of the dipole modes using

$$MAXWELL_d(1, PEC)$$
 and $MAXWELL_d(1, PMC)$

4: for each partial derivative $\partial f_k/\partial d_l$, with the corresponding m_{f_k} and $\mathrm{BC}_{\mathrm{SP},f_k}$, and $d_{h,l}=(d_1,...,d_l+h,...,d_N)$, compute

$$MAXWELL_{d_{h,l}}(m_{f_k}, BC_{SP, f_k})$$

in order to numerically compute $\partial f_k/\partial d_l$ using the forward difference method

5: **return** $F(d) = (F_1(d), ..., F_n(d))$

First, an $R_{eq}[\mathrm{mm}] \in [325, 375]$ [cf. Table II] is found such that the frequency of the FM of the corresponding cavity is $f_{\mathrm{FM}}(\boldsymbol{d}) = 400.79$ MHz (2:1). This is done using the zero-finding method TOMS 748 [42]. Each evaluation of f_{FM} requires finding the smallest eigenpair of the GEVP (which will be referred to as "solving" the GEVP) corresponding to m=0 and PEC BC_{SP}. In case the cavity found this way is reentrant, the second constraint is violated, so this individual is declared invalid and discarded from the population (2:2). Otherwise, the rest of the objective function values are computed (2:3–5).

In order to compute the properties of the TM₀₁₀ and TE₁₁₁, two GEVPs need to be solved (2:3). In order to numerically compute the local sensitivity (i.e., the partial derivative) $\partial f_k/\partial d_l$ using the forward difference method, f_k needs to be evaluated in the point $d_{h,l} = (d_1, ..., d_l + h, ..., d_N)$, i.e., another GEVP needs to be solved for the appropriate azimuthal mode number m_{f_k} and BC on the cross-sectional symmetry plane BC_{SP,f_k}.

Once the first generation of the EA is evaluated, a predetermined number of cycles is performed, each resulting in a new generation (1:3-7). In each cycle, new individuals are created using crossover and mutation operators (1:4-5), and their objective function values are again computed (1:6) as shown in Algorithm 2. The new generation is then chosen to comprise approximately M fittest individuals (1:7).

VII. RESULTS

A. Implementation and timings

The implementation of the optimization algorithm from the previous section is based on a combination of a massively parallel implementation of an EA [31,32], written in C++ and parallelized using MPI, with the axisymmetric Maxwell eigensolver [43]. For a point \boldsymbol{d} , a mesh of half of the cross section of the corresponding cavity is created using the Gmsh [44] C++ API. In the FEM linear elements are used, and the resulting GEVPs are solved using the symmetric Jacobi–Davidson algorithm [45].

In (2:1) a cheap solve on a coarse mesh is performed first, in order to compute a good approximation for $f_{\rm FM}$, $f_{\rm FM,coarse}$, (3–4 significant digits). Whenever $f_{\rm FM,coarse}$ is further away from 400.79 MHz than a given value ε (in this paper, $\varepsilon = 1$ MHz), the value $f_{\rm FM,coarse}$ is used in order to speed up the computation. Once the zero-finding method gets closer to 400.79 MHz than ε , a more expensive solve, on a much finer mesh, is performed in order to compute five significant digits of $f_{\rm FM}$. As a zero-finding method, the Boost¹ C++ library implementation of the TOMS 748 algorithm is used. Additionally, the zero-finding method is stopped as soon as five significant digits of $f_{\rm FM}[{\rm MHz}]$ match 400.79. When the search interval [325, 375] is first subdivided into three similarly-sized parts, (usually) using coarse solves, this entire approach requires, on average, only 2.2 fine solves in (2:1). The coarse solves use a mesh with around 20'000 triangles, and the fine ones around 500'000 triangles. On one core of the Intel Xeon Gold 6150 a coarse solve (creating a mesh, computing five smallest eigenpairs and the objective function values) takes around 2 s. A fine solve takes around 95 s (18 s for creating a mesh, 74 s for computing the eigenpairs, and 3 s for computing the objective function values). Additionally, in (2:3) two GEVPs on the same mesh need to be solved, so the mesh can be reused. Similarly, for each partial derivative that needs to be computed in (2:4), the solves are grouped in such a way as to avoid remeshing (eigenproblems corresponding to the same point, i.e., the same cavity shape, are solved consecutively).

To give an impression of the computation work for the entire optimization, using 108 processes on Euler IV (Euler cluster² of ETH Zurich, three Intel Xeon Gold 6150 nodes, each with 18 cores @ 2.7 GHz and cache size 24.75 MB) it took almost 15 h 19 min to compute 60 generations with M = 100 (around 30% of the evaluated individuals were discarded from the optimization).

B. Optimization results

The trade-off between the eight objective functions [cf. Eq. (5)] in the Pareto front approximation obtained in the 60th generation of an optimization with M = 100 is illustrated in Fig. 9 using a scatter-plot matrix. The histogram at position (i, i) (i.e., on the main diagonal) shows the distribution of the objective function F_i . The graph at position (i, j) where i < j (i.e., above the main diagonal) shows the values of F_i (y axis) and F_j (x axis) for the individuals in the last generation. The number in the *i*th row and the jth column where i > j (i.e., below the main diagonal) is the correlation coefficient between F_i and F_i . ${\cal F}_1$ and ${\cal F}_2$ are positively correlated because f_1 (in most cases f_1 is $f_{\text{TE}_{111}}$) is more sensitive to the geometric changes than f_2 , and the further it gets from f_{FM} , the closer it gets to f_2 . Improving F_1 and F_2 leads to a higher transverse impedance (larger F_3) and a more sensitive $f_{\text{TE}_{111}}$ (larger F_7). The V-shaped curve of F_7 vs F_2 is created because the rise in F_2 ($F_2 = |f_1 - f_2|$) corresponds to cases where $f_{\text{TE}_{111}}$ is higher than $f_{\text{TM}_{110}}$. In order to simplify the damping of the dipole modes using coaxial HOM couplers, F_2 should be as close to zero as possible. A close-to-zero value of F_2 fixes F_7 to around 3.1 MHz/mm. However, in choosing a point from the Pareto front approximation, a low importance can be assigned to F_7 because the transverse impedance of the TE₁₁₁ mode is much smaller than that of the TM_{110} mode.

All of the individuals, i.e., rf cavities, in the last generation are compared with four other storage ring single-cell cavities: CESR-B [46], HL-LHC [20], FCC $_{\rm HO18}$ [47,48] and FCC $_{\rm IC18}$ [17] in Table IV. Due to the conflicting nature of some of the objective functions, it is not feasible to surpass the other cavities in all objectives. However, all of these cavities are equally good in the Pareto sense.

The design variable and objective function values, as well as QoIs, for a chosen cavity from the last generation, called FCC_{PR19} , are given in Table V. The shape of FCC_{PR19}

¹https://www.boost.org.

²https://scicomp.ethz.ch/wiki/Euler.

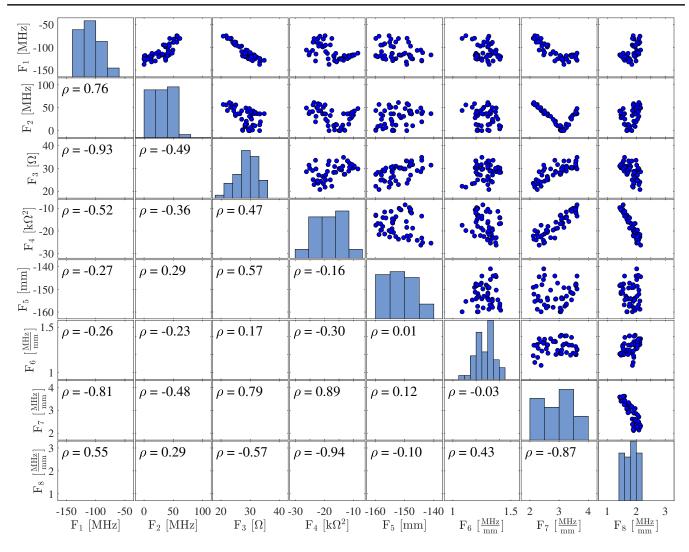


FIG. 9. A scatter plot of the objective functions for the individuals in the last generation. Only individuals with $140 \le R_i \le 160$ and $200 \le L \le 400$ are shown in order to preserve the scale. The numbers below the diagonal are the correlation coefficients between pairs of objective functions.

is shown in Fig. 10. The chosen cavity outperforms the other cavities in five or six objective functions. Since the design of FCC-ee-Z demands the cavities to be operated at low $E_{\rm acc}$ [[8] p. 120], a small value of $G \cdot R/Q$ does not lead to a significantly large dynamical loss on the surface of cavity. Therefore, the value of F_4 is sacrificed in favor of

TABLE IV. The value in ith row, jth column is the number of distinct cavities that have at least j objectives better than the cavity which corresponds to row i. The last (60th) generation contains 95 distinct cavities.

j	1	2	3	4	5	6	7	8
HL-LHC	95	95	92	79	47	10	0	0
CESR-B	95	95	92	68	40	9	0	0
FCC_{HO18}	95	94	78	51	19	1	0	0
FCC _{IC18}	95	95	93	86	35	5	0	0

other objective functions. The cavity shapes from Table V are shown and compared in Fig. 11. A small volume around the equator of FCC_{PR19} results in a small $G \cdot R/Q$. The large iris radius of FCC_{PR19} ensures that the dangerous higher order monopole modes are untrapped and propagate out of the beam pipes. Compared to other shown cavities, the frequency of the FM of FCC_{PR19} is between 9% and 15% less sensitive against changes in R_{eq} . The local sensitivity of FCC_{PR19} with respect to geometric parameters is given in Table VI. These numbers confirm that the assumptions made in Sec. V are valid for FCC_{PR19} .

VIII. GENERALIZATION

In this section the approach described in Secs. II–VI will be generalized and applied to a single-cell cavity with a symmetric cross section and half of its boundary defined as a Bézier curve of order three (cubic) with horizontal end

TABLE V. A comparison of a cavity on the Pareto front approximation, denoted FCC_{PR19}, with some storage ring single-cell cavities. The dimensions of CESR-B and HL-LHC cavities are scaled and tuned such that $f_{\rm FM}=400.79$ MHz. The cavities FCC_{HO18} and FCC_{IC18} are two other cavities, optimized with respect to a different set of objective functions, proposed for FCC by Zadeh and Kranjčević, respectively. The italic and bold entries indicate, respectively, a better and worse value than that of FCC_{PR19}. Note that in the calculation of $E_{\rm acc}$ (in $E_{\rm pk}/E_{\rm acc}$ and $B_{\rm pk}/E_{\rm acc}$), the actual cavity length (L) is considered.

Variable	FCC _{PR19}	CESR-B	HL-LHC	FCC _{HO18}	FCC _{IC18}
R_i [mm]	153.704	150.0	150.0	156.0	141.614
L [mm]	274.199	300.0	280.0	240.0	292.54
A [mm]	53.582	103.750	104.0	70.0	103.54
<i>B</i> [mm]	53.582	103.750	104.0	70.0	127.521
a [mm]	36.6831	25.0	25.0	25.0	41.921
b [mm]	36.6831	25.0	25.0	25.0	45.812
R_{eq} [mm]	363.346	341.856	338.512	350.574	339.166
QoI	FCC _{PR19}	CESR-B	HL-LHC	FCC _{HO18}	FCC _{IC18}
$f_{\mathrm{TE}_{111}}$ [MHz]	526.80	513.20	523.53	529.61	547.82
$f_{\mathrm{TM}_{110}}$ [MHz]	526.94	542.65	543.36	528.76	548.22
$R/Q[\Omega]$	78.2	89.5	90.6	79.0	94.9
$R/Q_{\perp,\mathrm{TE}_{111}}$ [Ω]	3.2	5.5	4.6	2.3	5.1
$R/Q_{\perp,\mathrm{TM}_{110}}$ [Ω]	26.8	24.1	26.7	27.8	31.2
α [degree]	109.2	104.9	99.0	102.8	91.7
$\frac{E_{\mathrm{pk}}}{E_{\mathrm{acc}}}$ [—]	1.8	2.0	2.0	1.9	1.9
$\frac{B_{ m pk}}{E_{ m acc}} \left[\frac{{ m mT}}{{ m MV/m}} \right]$	4.7	4.2	4.0	4.1	4.2
Objective	FCC _{PR19}	CESR-B	HL-LHC	FCC _{HO18}	FCC _{IC18}
$\overline{F_1 \text{ [MHz]}}$	-126.01	-112.42	-122.74	-127.97	-147.03
F_2 [MHz]	0.15	29.44	19.83	0.85	0.40
$F_3[\Omega]$	30.0	29.6	31.2	30.1	36.3
F_4 [k Ω^2]	-15.0	-21.8	-21.3	-15.5	-21.3
F_5 [mm]	-153.704	-150.0	-150.0	-156.0	-141.614
$F_6 \left[\frac{\text{MHz}}{\text{mm}} \right]$	1.21	1.32	1.36	1.31	1.39
$F_7 \left[\frac{\text{MHz}}{\text{mm}} \right]$	3.08	2.71	2.90	3.16	3.20
$F_8 \left[\frac{\text{MHz}}{\text{mm}} \right]$	1.68	1.98	1.99	1.79	1.92

slopes. Such a parametrization is shown in Fig. 12 and referred to as "single-cell Bézier cavity." Related work includes the study of superconducting rf cavities whose boundary is a Bézier curve [49] or a nonuniform rational B-spline [50]. A summary of the approach described in Secs. II–VI and its application to the single-cell Bézier cavity is given in Fig. 13.



FIG. 10. The electric field of the FM in half of the chosen elliptical cavity, denoted FCC_{PR19} in Table V.

The derivative of the contour of the elliptical cavity is not smooth at the intersection points of the ellipses and the tangential straight line. This is improved using Bézier curves. Four points, i.e., eight parameters in total, are required to model the geometry with a cubic Bézier curve. The cubic Bézier curve is given by

$$\mathbf{B}(t) = \sum_{i=0}^{3} \frac{3!}{i! \cdot (3-i)!} (1-t)^{3-i} \cdot t^{i} \cdot \mathbf{P}_{i}, \qquad 0 \le t \le 1,$$

where the four points P_0 , P_1 , P_2 , and P_3 are defined in the two-dimensional space. The requirement that the end slopes at P_0 and P_3 are zero is enforced by setting $y_0 = y_1 = R_{eq}$ and $y_2 = y_3 = R_i$. Furthermore, x_0 and x_3 can be set to 0 and L/2, respectively, so the cross section of the cell of such cavities can be parametrized by R_{eq} , R_i , L, x_1 , and x_2 , i.e.,

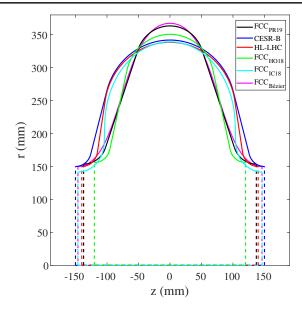


FIG. 11. A comparison of different cavity shapes from Table V. The details for the Bézier cavity are given in Sec. VIII.

$$\begin{bmatrix} \mathbf{P}_0 & \mathbf{P}_1 & \mathbf{P}_2 & \mathbf{P}_3 \end{bmatrix} = \begin{bmatrix} 0 & x_1 & x_2 & L/2 \\ R_{eq} & R_{eq} & R_i & R_i \end{bmatrix}.$$

Considering the intervals from Table VII, the main sensitivities are shown in Fig. 14, where it can be seen that the geometric parameter with the greatest influence on the frequency of the FM is R_{eq} . The main sensitivities in the case where, as in Sec. IV, R_{eq} is used to tune $f_{\rm FM}$ to 400.79 MHz are shown in Fig. 15. Note that the sensitivity of the Bézier cavity and the elliptical cavity cannot be compared as their shapes are parametrized differently. Based on the information from Figs. 14 and 15,

$$\frac{\partial f_{\rm FM}}{\partial R_{eq}}, \frac{\partial f_{\rm TE_{111}}}{\partial L}, \frac{\partial f_{\rm TE_{111}}}{\partial R_i}, \frac{\partial f_{\rm TM_{110}}}{\partial R_{eq}} \frac{\partial f_{\rm TM_{110}}}{\partial R_i} \quad \text{and} \quad \frac{\partial f_{\rm TM_{110}}}{\partial x_2}$$

need to be taken into account. As in Sec. V, due to the influence of x_1 on R_{eq} (Fig. 15) and their geometric connection (Fig. 12), in order to decrease the computation cost the local sensitivities with respect to x_1 are omitted. This leads to the constrained multiobjective optimization problem

TABLE VI. Some of the local sensitivities for the rf cavity from Fig. 10, denoted FCC_{PR19} in Table V. The magnitude of other sensitivities is below 0.7 MHz/mm.

Sensitivity Value [MHz/mm]	$\partial f_{\rm FM}/\partial R_{eq}$ -1.21	$\partial f_{\text{FM}}/\partial A$ -1.15	$\frac{\partial f_{\text{TE}_{111}}/\partial R_i}{-2.62}$
Sensitivity Value [MHz/mm]	$\partial f_{\mathrm{TM}_{110}} -1.3$		$\frac{\partial f_{\text{TM}_{110}}/\partial A}{-1.00}$

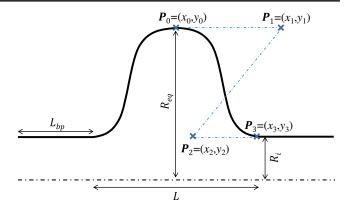


FIG. 12. Parametrization of a single-cell Bézier cavity. The cross section is symmetric, and half of the curved part of the boundary is a Bézier curve.

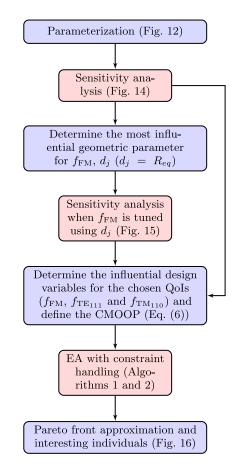


FIG. 13. A summary of Secs. II–VI, applied to a different parametrization. The methods are red and the conclusions blue. The corresponding information for the single-cell Bézier cavity is referenced in parentheses.

TABLE VII. Intervals for the single-cell Bézier cavity.

Variable [mm]	R_{eq}	R_i	L	x_1	x_2
Lower bound	325	145	240	50	30
Upper bound	375	160	380	120	100

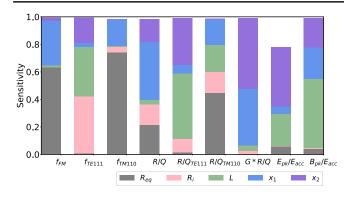


FIG. 14. Main sensitivities. The intervals for the geometric parameters R_{eq} , R_i , L, x_1 , and x_2 [cf. Fig. 12] are given in Table VII. The polynomial degree is p = 2, so the size of the sample is 84 [cf. Table I].

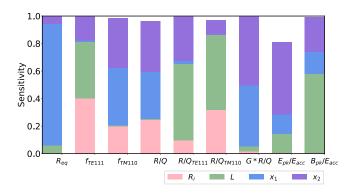


FIG. 15. Main sensitivities. For a point $(R_{eq}, R_i, L, x_1, x_2)$ [cf. Fig. 12] inside the intervals from Table VII, $f_{\rm FM}$ is tuned to 400.79 MHz using $R_{eq}[{\rm mm}] \in [325, 375]$. The polynomial degree is p=2, so the size of the sample is 45 [cf. Table I].

$$\min_{R_{i},L,x_{1},x_{2}} \left(f_{\text{FM}} - f_{1}, |f_{1} - f_{2}|, \frac{R}{Q_{\perp 1}} + \frac{R}{Q_{\perp 2}}, -G \cdot \frac{R}{Q}, \underbrace{-R_{i}}_{=F_{5}}, \underbrace{\left| \frac{\partial f_{\text{FM}}}{\partial R_{eq}} \right|}_{=F_{6}} \underbrace{\left| \frac{\partial f_{\text{TE}_{111}}}{\partial L} \right| + \left| \frac{\partial f_{\text{TE}_{111}}}{\partial R_{i}} \right|}_{=F_{7}}, \underbrace{\left| \frac{\partial f_{\text{TM}_{110}}}{\partial R_{eq}} \right| + \left| \frac{\partial f_{\text{TM}_{110}}}{\partial R_{i}} \right| + \left| \frac{\partial f_{\text{TM}_{110}}}{\partial x_{2}} \right|}_{F_{G,8}} \right)}_{F_{G,8}}$$
subject to $f_{\text{FM}} = 400.79 \text{ MHz}.$ (6)

It is implied that, for a point $I = (R_i, L, x_1, x_2)$, f_{FM} will be tuned to 400.79 MHz using R_{eq} .

The optimization problem is solved as described in Secs. VIB and VIIA. An approach similar to the one in Sec. VIIB is used to choose an interesting individual from the Pareto front approximation. The shape of the chosen Bézier cavity is shown in Fig. 16.



FIG. 16. The electric field of the FM in half of the chosen Bézier cavity.

The design variables, objective function values and QoIs for this cavity are given in Table VIII. The shape of the cavity (also shown in Fig. 11), as well as the values of the objective functions and QoIs are very close to the corresponding values for the elliptical cavity found in Sec. VII B (FCC_{PR19}). The smaller R_i value of the Bézier cavity increases R/Q and $G \cdot R/Q$ at the cost of a higher $R/Q_{\perp TE111}$. Having a different shape with rather similar rf properties provides flexibility to overcome problems such as multipactor effect that might exist in one of the cavity shapes.

TABLE VIII. Design variable and objective function values, as well as QoIs, for the Bézier cavity from Fig. 16. The frequency of the FM is $f_{\rm FM} = 400.79$ MHz.

$\frac{\text{dic I WI is } f_{\text{FM}} = 400.77 \text{ WHIZ.}}{}$	
Variable	Fig. 16
R_i [mm]	150.037
L [mm]	292.19
x_1 [mm]	78.6565
x_2 [mm]	75.4653
R_{eq} [mm]	367.265
QoI	Fig. 16
$f_{\mathrm{TE}_{111}}$ [MHz]	528.78
$f_{\mathrm{TM}_{110}}$ [MHz]	528.76
$R/Q[\Omega]$	79.9
$R/Q_{\perp { m TE}_{111}}$ [Ω]	4.1
$R/Q_{\perp \mathrm{TM}_{110}}$ [Ω]	26.7
$\frac{E_{\mathrm{pk}}}{E_{\mathrm{acc}}}$ [—]	1.6
$\frac{B_{ m pk}}{E_{ m acc}} \left[\frac{{ m mT}}{{ m MV/m}} \right]$	4.9
Objective	Fig. 16
F_1 [MHz]	-127.97
F_2 [MHz]	0.02
$\overline{F_3}$ [Ω]	30.8
F_4 [k Ω^2]	-15.6
F_5 [mm]	-150.037
$F_6 \left[\frac{\text{MHz}}{\text{mm}} \right]$	1.08
$F_7 \left[\frac{\text{MHz}}{\text{mm}} \right]$	2.79
$F_{G,8}$ [$\frac{\text{MHz}}{\text{mm}}$]	1.68

IX. CONCLUSIONS

In this paper an algorithm for solving constrained multiobjective rf cavity shape optimization problems was proposed and applied to the problem of optimizing the shape of the superconducting rf cavity for the FCC-ee-Z. The shape of the cavity was optimized with respect to both the properties of the fundamental mode and the first dipole band, focusing in particular on robustness against geometric perturbations.

In order to decrease the computation cost, the results of a global sensitivity analysis were used to reduce the search space and define the objective functions of interest. A good single-cell elliptical cavity was found, and the algorithm generalized and applied to a different type of cavity. The proposed algorithm and its implementation could also be applied to rf cavity shape optimization problems which take into account the properties of HOMs corresponding to arbitrary mode numbers.

ACKNOWLEDGMENTS

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