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Fast adiabatic qubit gates using only σ_z control

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A controlled-phase gate was demonstrated in superconducting Xmon transmon qubits with fidelity reaching 99.4%, relying on the adiabatic interaction between the $|11\rangle$ and $|02\rangle$ states. Here we explain the theoretical concepts behind this protocol, which achieves fast gate times with only σ_z control of the Hamiltonian, based on a theory of nonlinear mapping of state errors to a power spectral density and use of optimal window functions. With a solution given in the Fourier basis, optimization is shown to be straightforward for practical cases of an arbitrary state change and finite bandwidth of control signals. We find that errors below 10^{-4} are readily achievable for realistic control wave forms.

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I. INTRODUCTION

High-fidelity controlled-phase (CZ) gates were recently demonstrated for adjacent qubits in a five-qubit quantum processor [1]. For a two-qubit gate, CZ fidelity as high as 99.4% was obtained using the phase shift from near crossing of the |11\rangle and |02\rangle states, similar to previous work on superconducting qubits [2–5], but following a "fast adiabatic" protocol that gives both fast gate times and low error rates. Here we explain the theoretical concepts behind this protocol, showing how wave forms may be generated and parametrized in a manner that is readily optimized, both numerically and experimentally.

The general problem of achieving fast adiabatic performance is of great interest to the physics community [6–9], with applications in coherent manipulation, precision measurements, and quantum computing. Finding an improved control methodology therefore has potential for applications in a variety of quantum systems.

Although adiabatic quantum control has a long history, we are not aware of a theory that could address in detail our particular experimental problem: What is the optimal control trajectory for the desired CZ operation, having both short gate time and small nonadiabatic error, described in a way that is readily optimized with numerical and experimental methods? For example, although a previous experiment used adiabaticity for a CZ gate [3], the characterization and optimization of the adiabatic control wave form was not developed. The goal here is to describe a simple theory that can be used to intuitively understand optimal wave forms for fast and accurate operation.

Since we are concerned here with transition errors between the states $|11\rangle$ and $|02\rangle$ as their frequencies are varied, we may map this to a qubit problem for states $|0\rangle$ and $|1\rangle$ that has a fixed energy in σ_x from the avoided level crossing and adjustable qubit frequency in σ_z . We note that this optimization problem differs from conventional quantum control, where the σ_z part of the Hamiltonian is fixed and both the σ_x and σ_y parts can be varied. Here, the single-component control in σ_z makes optimal control much more difficult since a second operator does not exist that could be used to restore the desired adiabatic character [10], disallowing standard pulse shaping techniques [10,11] and "superadiabatic" theory [12–14].

The set of possible adiabatic solutions is also influenced by other constraints with the superconducting qubit system. First, the nonlinearity of the qubits is small [1], so other unwanted transitions are not necessarily far off resonance. For example, although the wave form is optimized for adiabaticity of the near-resonance |11\rangle and |02\rangle states, fast changes in the control wave form may produce transitions between the states |01\rangle and |10\rangle, even though off resonance by \gtrsim 200 MHz. A natural solution to this constraint is to use smooth wave forms, which also seems to be a good strategy in terms of controlling many qubits, achieving low cross talk, and simplifying the requirements of control electronics and their system calibration. We thus rule out pulse and refocusing types of protocols common in nuclear magnetic resonance (NMR) [15,16], but welcome any theory making this a practical solution.

Second, we are looking for control wave forms with fast gate times, of order 40 ns for our qubits, that efficiently use the moderate coupling $g/2\pi \simeq 30$ MHz between qubits [17]; note that $2\pi/g = 33$ ns, making this problem nontrivial for adiabaticity. Fast gates are important for coherence, especially since errors from 1/f flux noise and off-resonant coupling to other qubits, described as frequency shifts, produce phase errors proportional to characteristic gate times: If limited by these processes, the error probability thus scales as the *square* of gate time. A full system optimization of a qubit processor, for example describing the optimal choice of on-resonance qubit-qubit coupling g and its smaller off-resonant coupling, is beyond the scope of this article. But clearly even a small improvement of gate time can be quite important, motivating here a detailed understanding of both fast and accurate control.

Third, the CZ gate uses adiabatic control that is nonstandard, as it partially moves into the avoided level crossing regime and then back out again, with the amplitude and time of the wave form adjusted to give a π phase shift to the $|11\rangle$ state. Typical solutions to an adiabatic problem consider moving through an avoided level crossing [18–20], so a theory is needed to calculate optimal wave forms for an arbitrary control problem.

Since one varies only a single control parameter in σ_z , this theory may be described in a simple manner with familiar concepts that give insight into a physical understanding of

adiabatic behavior. We first show how state change from σ_z control in the qubit Hamiltonian can be intuitively described in an instantaneous coordinate system. Here, the evolution of the quantum system can be understood in the small-error limit using a geometrical construction. Second, this description allows errors to be calculated using a generalized Fourier transform. Third, to obtain a minimum-time solution for small changes in angle, the theory of window functions is used to find optimal control wave forms. We then show how to efficiently parametrize the optimum wave form using a few-term Fourier series. Lastly, a nonlinear mapping is used to find the optimal control wave form for an arbitrary change in the control. We also examine a second problem, microwave control of a qubit, and show that conventional two-component optimal control using σ_x and σ_y gives better results, as expected, but not by much.

These results present a simple yet effective way to optimize for adiabatic fidelity when changing control only in σ_z . Perhaps surprisingly, nonadiabatic errors can be small even if the interaction time is only slightly larger than the time scale of the transverse coupling, as proposed earlier for a nonadiabatic gate [21]. A recent experiment by Kelly and co-workers has also demonstrated that these theoretical ideas can be used to optimize CZ performance [22].

II. GEOMETRICAL SOLUTION

We start with a geometrical solution to nonadiabatic error, a problem previously described using conventional methods [14,23]. The description of a qubit given by the two-state Hamiltonian is

$$H = H_x \sigma_x + H_z \sigma_z = \begin{pmatrix} H_z & H_x \\ H_x & -H_z \end{pmatrix}, \tag{1}$$

where H_x is constant and the qubit frequency is controlled by a time-dependent $H_z(t)$, for example the magnetic field in the z direction for an electron spin. We consider the qubit to be controlled by a Hamiltonian vector $(H_x, 0, H_z)$, having both length and direction. As shown in Fig. 1(a), we define

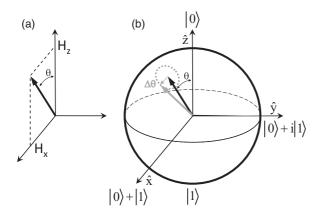


FIG. 1. Geometric picture of nonadiabatic errors. (a) Control field of constant H_x and variable $H_z(t)$ gives control angle θ from the z axis. (b) Bloch sphere representation gives a ground eigenstate parallel to the control vector (solid arrow). For a small step in θ , the quantum state (gray arrow) points slightly away, with change $\Delta\theta$ (small gray arrow). This state rotates around the eigenstate at frequency ω (gray dotted lines).

the control parameter θ as the angle between the Hamiltonian vector and the z axis,

$$\theta = \arctan(H_x/H_z). \tag{2}$$

The qubit state is described by a Bloch vector that can point in any direction on the Bloch sphere. Here, the usefulness of this geometric construction is the idea that the lowest-energy eigenstate has a Bloch vector that points in the direction of the control vector: The eigenstate vector has angle θ from vertical and lies in the plane of the \hat{x} and \hat{z} axes. The ground state eigenvalue $E_- = -\sqrt{H_x^2 + H_z^2}$ is proportional to the length of the control vector, whereas the excited eigenstate points in the opposite direction with eigenenergy $E_+ = -E_-$. If the Bloch state vector is not an eigenstate, then its time dynamics is to rotate (precess) around the eigenstate vector at a frequency proportional to the difference in eigenstate energies,

$$\omega = (E_{+} - E_{-})/\hbar = 2\sqrt{H_x^2 + H_z^2}/\hbar.$$
 (3)

Nonadiabatic errors are typically understood by considering the Landau-Zener trajectory $H_z = \dot{H}_z t$. Initially at large negative times t, where H_z is large and negative, the Hamiltonian control vector points in the down direction, with the state given by the Bloch vector also in the down direction, having $\theta = \pi$. As the time goes to zero, the control vector $(H_x, 0, H_z = 0)$ points along the equator, with the Bloch vector $\theta = \pi/2$. At large positive times the control and Bloch vectors point in the up direction $\theta = 0$.

When the field is slowly varied $(\dot{H}_z \rightarrow 0)$, the state adiabatically changes from $\theta = \pi$ to $\theta = 0$. A finite ramp rate gives errors, with the probability of making a transition to the excited final eigenstate as

$$P_e = \exp(-\pi H_x^2/\hbar \dot{H}_z). \tag{4}$$

A small final error $P_e = 10^{-4}$ implies the condition $\dot{H}_z = 0.341 H_x^2/\hbar$. If we assume a change in H_z from $-10H_x$ to $10H_x$, the total time for the control pulse is $t_c = 18.6(h/2H_x)$, where $h/2H_x$ is the oscillation time for the transverse Hamiltonian.

The factor 18.6 implies a long control time for the Landau-Zener trajectory. We will show that with optimal control this time factor can be reduced significantly, to order unity.

For adiabatic evolution of the quantum state, the state vector stays aligned with the direction of the control Hamiltonian. To better characterize nonadiabatic errors, we change the reference frame of the Bloch sphere to coincide with that of the control Hamiltonian. In this moving frame $\theta_m=0$ represents being in the ground state of the instantaneous Hamiltonian, with no nonadiabatic error.

To understand nonadiabatic deviations, we consider the time dynamics of state change. We first consider an infinitesimal time step Δt , during which the the control vector initially changes its direction by $\Delta \theta$, as shown in Fig. 1(b). A transformation to the new moving frame produces a small change in the angle of the Bloch vector of $-\Delta \theta$. During the remainder of the time Δt the Bloch vector rotates around the $\theta_m=0$ axis, at the frequency ω . Starting with $\theta_m=0$ before this time step, the angles θ_x and θ_y in the x and y directions

can be written compactly as

$$\theta_m \equiv \theta_x + i\theta_y = -\Delta\theta \, e^{-i\omega\Delta t} \tag{5}$$

after time Δt , assuming small angles as appropriate for small nonadiabatic errors. This finite angle implies that the quantum state has not changed adiabatically (as $\theta_m \neq 0$), with the error probability given by $P_e = (\sin |\theta_m|/2)^2 \approx |\theta_m|^2/4$.

We next understand the error coming from a sequence of time steps Δt_i each with a small angle change $\Delta \theta_i$. A useful simplification comes from changing to a new frame that rotates with the precession of the Bloch vector, which implies that the above θ_m vector does not rotate. Instead, the change $\Delta \theta_i$ rotates in time by an angle ϕ_i , which is computed by summing over all the previous phase changes in the individual steps. The net change in the complex angle of the Bloch vector in this moving and rotating frame is

$$\theta_{mr} = -\sum_{i} \Delta \theta_i \, e^{-i\phi_i} \tag{6}$$

$$= -\int dt \, (d\theta/dt) \exp\left[-i \int_{-t}^{t} \omega(t')dt'\right], \qquad (7)$$

$$P_e = |\theta_{mr}|^2 / 4, \tag{8}$$

where P_e is the total probability error. These equations are the general solution for nonadiabatic error in the small-error limit, derived here using a geometric description.

The amplitude error θ_{mr} is proportional to the rate of change in the eigenstates, as given by $d\theta/dt$. This quantity is mostly averaged to zero because of the rapidly changing phase from the ω integral, with averaged amplitude decreasing with increasing qubit frequency ω .

Note that we have assumed here that all changes add linearly. This is a good assumption as long as the net change in angle is very small; for example, it is known that qubits with small change in state can be approximated by a harmonic oscillator, a linear system. This assumption is generally acceptable since we are interested only in qubit control that gives small errors.

An exact calculation of the error, starting from the Schrödinger equation, is derived in the Appendix. It gives results identical to this geometrical solution in the small-error limit. There are two changes for the exact solution: the driving amplitude $d\theta/dt$ should also be multiplied by the factor $\cos |\theta_{mr}|$, and the probability is $P_e = \{\sin[\arcsin(\theta_{mr})/2]\}^2$.

To understand how to construct an optimal wave form, we first note that there is a trade-off between the control time and the magnitude of nonadiabatic error. Zero error is not possible—we are instead looking for acceptably small errors, say below 10^{-4} , with as short of a control time as possible. We also want stable control wave forms, so that the error does not increase rapidly for small changes in the wave form.

We first note that nonadiabatic errors are proportional to a change in the control variable θ . Errors will be relatively small when $|H_z| \gg H_x$, due to the slow dependence of θ on H_z/H_x and the large oscillation frequency ω , as described in Eqs. (2) and (3).

III. OPTIMAL SOLUTION: SMALL CHANGE IN θ

For the case where H_z changes by only a small amount, the qubit frequency can be approximated as being constant, $\omega = \omega_0$. Then Eq. (7) is just the Fourier transform of $d\theta/dt$ at frequency ω_0 , which gives a probability proportional to the power spectral density of the signal at the oscillation frequency ω_0 ,

$$P_e = (1/4)S_{d\theta/dt}(\omega_0). \tag{9}$$

This makes sense physically since it is power at the transition frequency, given by $|\theta_{mr}|^2$, that drives the qubit transition to produce errors.

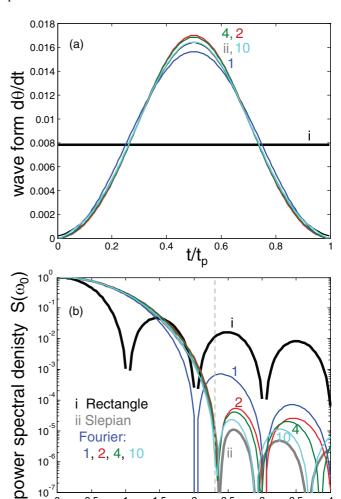


FIG. 2. (Color online) Wave form and power spectral density $S(\omega_0)$, for a small change in control θ such that the qubit frequency is approximately constant, $\omega = \omega_0$. Here, the spectral density is dimensionless and proportional to the qubit error $P_e \simeq (1/4)S(\omega_0)$, valid in the limit of large t_p . (a) Plot of wave form $d\theta/dt$ versus t/t_p for a rectangular pulse (i, black), Hanning window (1, blue), Slepian (ii, gray), and optimized Fourier coefficients with two (2, red), four (4, green), and ten (10, cyan) terms. Note the similarity of the Slepian curves to the optimized Fourier functions, except near $t/t_p = 0, 1$. (b) Plot of power spectral density obtained from a Fourier transform of the wave forms in (a). All wave forms are normalized to have unit area. The vertical dashed line shows the spectral cutoff for minimization of integrated noise.

1.5

2.5

 $t_p \omega_0/2\pi$

3

3.5

0.5

For the simple case of a linear change in the control θ from θ_i to θ_f , the error is proportional to the (constant) derivative $d\theta/dt = (\theta_f - \theta_i)/t_p$, shown as the black line in Fig. 2(a). The error versus the pulse time t_p is the power spectral density of a rectangular pulse,

$$P_e^{\text{sq}} = (1/4) \left| (d\theta/dt) \int_0^{t_p} dt \exp(-i\omega_0 t) \right|^2$$
 (10)

$$= (\theta_f - \theta_i)^2 \sin^2(\omega_0 t_p/2)/\omega_0^2 t_p^2, \tag{11}$$

which corresponds to the square of the sinc function. Although the error is zero when $t_p=2\pi/\omega_0$, small errors occur only around a small range of this time, so we consider this solution not to be stable or practically useful. For large times, the general falloff of the error is quite slow, scaling as $1/t_p^2$, so this control wave form does not present a good solution to adiabatic control.

The functional dependence of the error versus pulse time can be changed by using different control wave forms [24]. Another example is the function $[1 - \cos(2\pi t/t_p)]/2$, known as the Hanning window, as shown in blue in Fig. 2. In this case, the error is

$$P_e^{\rm H} = P_e^{\rm sq}/|1 - (\omega_0 t_p/2\pi)^2|^2,$$
 (12)

where the first zero is at twice the time of the square window $t_p = 4\pi/\omega_0$, but the error falls off more quickly as $1/t_p^6$ at large times. Although this window function is superior to the square window, the errors are still somewhat large right after the first zero, so other window functions should be considered.

As well known from NMR [25], optimizing nonadiabatic error versus total control time t_p can be understood as optimizing windowing functions: the optimization criteria depend on the requirements for a particular application.

Since we are interested in low errors for a range of pulse times, we define the optimal wave form as one that gives low error for any time larger than some chosen time: this definition is essentially equivalent to adiabatic behavior. We can use the theory of optimal window functions in signal processing to find this control wave form, known as a Slepian [26]. This function is defined by finding the optimal wave form that minimizes the integrated spectral density above a chosen frequency. For example, the Slepian wave form optimized for frequencies above $\omega_0 t_p/2\pi > 2.3$ is shown in gray in Fig. 2. Although the first minimum is at slightly longer times than for the Hanning window, the error is always below 10^{-5} above this value. A different Slepian cutoff parameter can be used to trade off the maximum error for the time of the first minimum.

A disadvantage of the Slepian function is that it does not go to zero at the beginning and end of the pulse, which creates a sharp edge as for a rectangular pulse. This makes the function harder to synthesize, since real wave form generators have finite response times. A practical way around this issue is to find a near-optimum solution using Fourier basis functions [25] given by

$$\frac{d\theta}{dt} = \sum_{n=1,2,\dots,n_m} \lambda_n [1 - \cos(2\pi nt/t_p)],\tag{13}$$

which goes smoothly to zero at the beginning and end of the pulse. The coefficients are constrained by the height of the

TABLE I. Table of λ_n coefficients obtained from numerical minimization of noise for frequencies $\omega_0 t_p/2\pi > 2.3$, using the constraint $\sum \lambda_n = 1$.

$\overline{n_m}$	λ_1	λ_2	λ_3	λ_4	λ_5	λ_6	λ ₇
2	1.0866	-0.0866					
4	1.0751	-0.0811	0.0017	0.0044			
10	1.0280	-0.0606	0.0052	0.0055	0.0047	0.0046	0.0035

pulse,

$$\theta_f - \theta_i = t_p \sum_n \lambda_n. \tag{14}$$

Using standard numerical minimization (the "fminsearch" function in MATLAB), the coefficients $\{\lambda_n\}$ can be found that optimizes for the minimum integrated spectral density, as for the Slepian. The wave form and spectral densities for

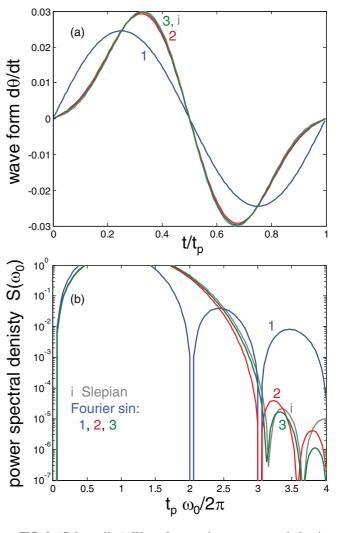


FIG. 3. (Color online) Wave form and power spectral density $[S(\omega_0) \simeq 4P_e]$ as for Fig. 2, but with a trajectory from θ_i to θ_f and back. (a) Plot of wave form $d\theta/dt$ versus t/t_p for optimized Fourier coefficients with one (1, blue), two (2, red), and three (3, green) terms, as well as the Slepian (i, gray). (b) Plot of power spectral density from Fourier transforms of the wave forms in (a). Wave forms normalized using $2\sum \lambda'_n = 1$.

 $n_m = 2$, 4, and 10 are plotted in Fig. 2, which shows performance close to the Slepian. Coefficients are given in Table I, showing that the first two are dominant. It is particularly interesting to note that the wave form for $n_m = 2$ (red), with only two Fourier coefficients, is reasonably close in shape to the Slepian and has acceptably small errors $<10^{-4}$.

The Fourier basis functions can also include the terms $\lambda_n^s \sin(\pi n t/t_p)$ for n odd, which go to zero at $t=0,t_p$ but have nonzero derivatives there. Upon using this larger basis set, we find that the numerically optimized solutions are not much better than found above using only cosine terms. We choose not to use use these sine basis functions because we prefer to also have a zero derivative at the beginning and end of the pulse.

We are also interested in control wave forms that take an initial θ_i to θ_f and then back again to θ_i . In this case the wave form for θ , not $d\theta/dt$, is given by the Fourier basis

$$\theta - \theta_i = \sum_{n=1,2,...,n_m} \lambda'_n [1 - \cos(2\pi nt/t_p)],$$
 (15)

with constraints on the coefficients

$$\theta_f - \theta_i = 2 \sum_{n \text{ odd}} \lambda_n'. \tag{16}$$

The derivative $d\theta/dt$ is now a sum of sine functions, which have nonzero derivatives at the beginning and end of the pulse. Figure 3 shows optimized results for one, two, and three Fourier coefficients, as well as for the first-harmonic solution of the Slepian. As for the previous case, the optimized Fourier solution is close to that of the Slepian: only two coefficients (red curve) are required for close approximation to the ideal solution.

IV. OPTIMUM SOLUTION: ARBITRARY θ

The optimal solution for an arbitrary change in θ will have to account for the qubit frequency ω changing during the control pulse. The solution is found by first making a coordinate transformation to an accelerating frame in time where the oscillation frequency is constant, and then using optimal wave forms as found previously.

The basic idea is to define a new frame with time τ such that the rate of change of the phase is constant. Choosing this frequency as the constant $\omega_x = 2H_x/\hbar$, the times are related by

$$\omega_x d\tau = \omega(t) dt. \tag{17}$$

In this frame, the error angle from Eq. (7) becomes

$$\theta_{mr} = -\int d\tau \, (d\theta/d\tau) \, \exp[-i\omega_x \tau], \tag{18}$$

so that the optimal shape of $d\theta(\tau)/d\tau$ can be computed for constant frequency ω_x , as done in the last section. Integration of $d\theta/d\tau$ is next used to compute the quantity $\theta(\tau)$. The relationship between the real time t and τ is finally solved

by integrating Eq. (17),

$$t(\tau) = \int_{-\tau}^{\tau} d\tau' \omega_x / \omega(\tau')$$
 (19)

$$= \int_{-\tau}^{\tau} d\tau' \sin \theta(\tau'), \tag{20}$$

where in the last equation we have used Eqs. (2) and (3) to find the relation $\omega = \omega_x / \sin \theta$. The function $t(\tau)$ can be inverted numerically using interpolation. For example, in MATLAB code the dependence $\theta(t)$ would be given through the interpolation of the vectors $t(\tau)$ and $\theta(\tau)$ using the command $\theta(t) = \text{interp}1(t(\tau), \theta(\tau), t)$.

To illustrate this solution for an arbitrary control problem, we first consider the avoided level crossing from $H_z = -10H_x$ to $H_z = 10H_x$, as discussed earlier for the Landau-Zener trajectory. The wave form is shown as solid lines in Fig. 4(a) for the two Fourier coefficients $\{\lambda_1 = 1.086, \lambda_2 = -0.086\}$ found earlier. Here we find that the control wave form is nearly linear

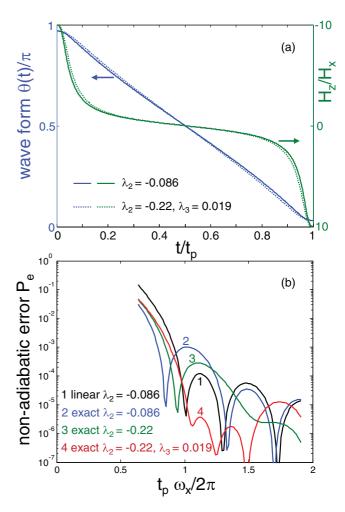


FIG. 4. (Color online) Wave form and error for H_z changing from $-10H_x$ to $10H_x$. (a) Blue lines are plot of control $\theta(t)$ versus t/t_p for two (solid) and three (dashed) Fourier coefficients. The right scale shows the wave form $H_z(t)$ as green lines. (b) Plot of nonadiabatic error versus pulse time t_p for linearized and exact solutions, with both two and three optimized coefficients. Low errors can be achieved for times greater than $t_p = 2\pi/\omega_x$, corresponding to one oscillation at the avoided level crossing. We define $\omega_x = 2H_x/\hbar$.

in θ , not in H_z as assumed for the standard Landau-Zener problem. The nonadiabatic error P_e is plotted in Fig. 4(b), for both the linearized (black) and the exact (blue) solutions of the qubit dynamics. The linearized solution has similar shape to that found earlier, as expected, whereas the exact solution shows significantly higher errors, demonstrating that an exact solution for this control problem is important.

To further reduce error, we next optimize the λ_n coefficients for better performance. Shown in green is the best performance obtained for varying λ_2 , whereas the red curve shows optimal performance for a choice of λ_2 and λ_3 . It is impressive that nonadiabatic errors in the 10^{-5} range can be obtained with only a small number of coefficients. Also note that low errors occur at a pulse time t_p equal to one oscillation period $h/2H_x$ of the avoided level crossing, about 20 times faster than in the simple Landau-Zener trajectory.

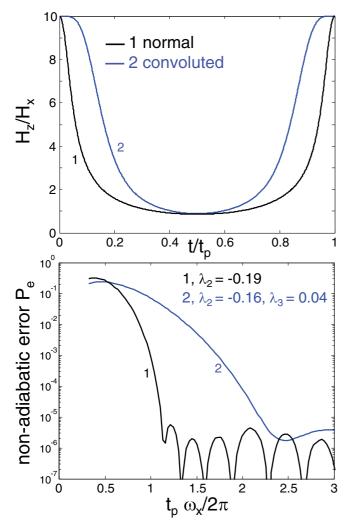


FIG. 5. (Color online) Wave form and error for H_z changing from $10H_x$ to $1.17H_x$, and then back to $10H_x$. (a) Black line (1) is for optimal wave form, whereas blue line (2) includes the effect of wave form rounding by control electronics, here accounted for by a convolution. (b) Plot of nonadiabatic error versus pulse time t_p for the two cases, using exact solutions. Optimal coefficients are $\{\lambda_2',\lambda_3'\}=\{-0.19,0\}$ for black (1) and $\{-0.16,0.04\}$ for blue (2) lines.

Another important wave form is for a controlled pulse in θ , as shown earlier in Fig. 3. Here we consider the control angle changing from $\theta_i = 0.1$ to $\theta_f = 0.55 \, \pi/2$, and then back again to θ_i ; these parameters are chosen to give the CZ gate $|11\rangle \rightarrow \exp(i\pi)|11\rangle$, where this additional π phase shift comes from the avoided level crossing between the states $|11\rangle$ and $|02\rangle$. In this case $\theta(\tau)$ is given by Eqs. (15) and (16), and $\theta(t)$ is found as for the last example. We find a near-optimal solution for only one additional coefficient $\lambda_2' = -0.19$, as shown in black in Fig. 5. We obtain a very low error for a total control time about one oscillation of $\omega_x/2\pi$.

This wave form requires fast changes in H_x at the beginning and end of the pulse. Because of the finite bandwidth of the control electronics, it may not be possible to synthesize such a shape, so we next consider the optimal control including rounding of the wave form, here accounted for by convolution of the output wave form with a Gaussian. With convolution, the computed nonadiabatic error increases significantly with the original λ_2 coefficient. We found that adjusting λ_2 could not improve the error much, but by including the next coefficient λ_3 the error could be effectively minimized. This optimal solution is shown as the blue line in Fig. 5, where low error is now found at about twice the oscillation period of $\omega_x/2\pi$. Note that about 1/2 of the increase in time of the control wave form is accounted for by the extra time added at the beginning and end of the pulse from convolution.

V. OPTIMUM SOLUTION FOR TWO-STATE ERRORS

The Fourier basis may be used to optimize wave forms for other control problems, so it is natural to ask whether these ideas can be used to minimize the excitation to the second excited state when driving weakly anharmonic qubits. This corresponds to the standard problem for optimal control, where shaping occurs in the off-diagonal σ_x and σ_y parts of the Hamiltonian. A solution that works extremely well is wave form shaping by derivative removal by adiabatic gate (DRAG) [27,28]. It is interesting to consider whether simple shaping using the Fourier basis wave forms can improve upon DRAG.

Errors from the $|2\rangle$ state come in two forms: (I) direct transitions to $|2\rangle$ producing 2-state errors, and (II) virtual excitations to $|2\rangle$, producing an ac Stark shift between the qubit states $|0\rangle$ and $|1\rangle$ that give changes in rotation amplitude and direction. The transition frequencies between the states $|0\rangle \leftrightarrow |1\rangle$ and $|1\rangle \leftrightarrow |2\rangle$ are ω_0 and $\omega_0 + \Delta$, which defines a nonlinearity Δ that is negative for transmons.

(I) Transitions to the 2 state come from the drive wave form having spectral density at the $|1\rangle \leftrightarrow |2\rangle$ frequency. As the drive needs to be resonant with the $|0\rangle \leftrightarrow |1\rangle$ transition, it can be described in general with a two-quadrature form $x(t)\cos(\omega_0 t) + y(t)\sin(\omega_0 t)$, which can be written compactly in complex notation as $[x(t)-iy(t)]\exp(i\omega_0 t)$. The drive wave form for DRAG has the quadrature component proportional to the time derivative $\dot{x}(t)$,

$$W(t) = [x(t) - iD\dot{x}(t)/\Delta]e^{i\omega_0 t}, \qquad (21)$$

with its magnitude adjusted by a dimensionless derivative parameter D. The transition rate is expected to scale as the spectral density of W, which is the squared magnitude of the

Fourier transform,

$$W(\omega) = \int dt \, e^{-i\omega t} W(t) \tag{22}$$

$$= [1 + D(\omega - \omega_0)/\Delta] \int dt \, e^{-i\omega t} x(t) e^{i\omega_0 t}, \qquad (23)$$

where integration by parts is used [29] to shift the derivative of x to the term $\exp[i(\omega_0 - \omega)t]$. Driving the $|0\rangle \leftrightarrow |1\rangle$ transition at ω_0 , the spectral density is zero at the $|1\rangle \leftrightarrow |2\rangle$ transition frequency $(\omega - \omega_0 = \Delta)$ when D = -1. As this is a qualitative argument, we use numerical simulation to minimize the population error to the 2 state; below we show that our optimal DRAG coefficient is D = -1.20, close to the computed value.

(II) The origin of the ac Stark shift can be understood as coming from level repulsion between the $|1\rangle$ and $|2\rangle$ states when driven off resonance, which changes the frequency of the $|1\rangle$ state proportionally to $-x^2/2\Delta$. The derivative term also Stark shifts the qubit, giving a net shift [30] proportional to $-(1+2D)x^2/2\Delta$. As the x^2 term gives a shift that is time dependent, a practical solution is to choose D=-1/2 for no Stark shift, allowing the qubit frequency to be constant [30]. In this case transition errors to the 2 state are not optimized, but the solution is still useful since errors are lowered by $(1+D)^2=1/4$.

As we wish to use D=-1 for minimum 2-state errors, we note here there exists another practical solution: constant detuning of the microwave frequency. This works because the dynamic Stark shift, which produces a small change in the rotation direction of the Bloch vector in the qubit subspace, can always be compensated for by a static change in drive frequency and phase. Equivalently stated, the three parameters of the drive frequency, phase, and amplitude move the Bloch vector by any desired rotation direction and angle. Finding zero error thus becomes the same procedure as for normal tuning up of qubit gates, only here the frequency and phase must be separately tuned for $\pi/2$ and π rotations. As this optimization can always be performed, we need only be concerned with 2-state errors.

These ideas were evaluated using numerical simulations of a three3-level system with Δ nonlinearity, integrating the Hamiltonian matrix over the wave form pulse to find the net unitary transformation. We find good performance using the Fourier basis, shown initially for a drive $x(t) \propto$ $1 - \cos(2\pi t/t_p)$ with amplitude adjusted to give a π pulse. As we find that the transition probabilities for $|0\rangle \rightarrow |2\rangle$ and $|1\rangle \rightarrow |2\rangle$ are quite close in magnitude, we plot in Fig. 6 the average of these 2-state errors versus pulse time t_p normalized with the qubit anharmonicity $|\Delta|/2\pi$. Here, we optimize amplitude and detuning parameters for qubit subspace errors below 10^{-7} at $t_p |\Delta|/2\pi \simeq 2.5$; we find the 2-state errors change little when slightly varying these parameters to zero errors at different values of t_p . The error without a derivative, D = 0, is plotted in black. The wave form with the half derivative D = -0.48, plotted in red, shows a reduction in error by about a factor of 4 which is consistent with the above argument. As expected, we find a much better solution using the DRAG coefficient near unity, D = -1.20, as plotted in green.

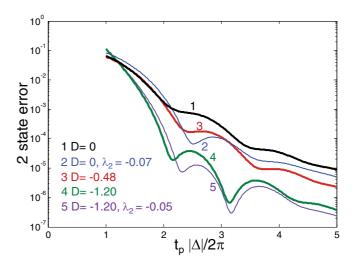


FIG. 6. (Color online) Average probability of 2-state population versus pulse time t_p for a qubit π pulse. The time is normalized with the qubit nonlinearity $|\Delta|/2\pi$. Parameters are chosen to minimize qubit errors near $t_p|\Delta|/2\pi=2.5$. The DRAG terms D=0, -0.48, and -1.20 are shown in black (1), red (3), and green (4), respectively. Optimizations of the wave form x(t) with Fourier coefficient λ_2 are shown in blue (2) and purple (5).

We also tested for wave form optimization of x(t) by adjusting the Fourier coefficient λ_2 . For the D=0 case we find better performance for $\lambda_2=-0.07$, plotted in blue, with an average magnitude similar to that of the D=-0.48 solution. For D=-1.20 and $\lambda_2=-0.05$, plotted in purple, we found that adjusting λ_2 shifted the first minimum; we thus propose that this parameter can be used to tune for a somewhat smaller 2-state error.

Comparing to plots in Ref. [28] using a truncated Gaussian wave form, we find that the D=-0.48 solution has errors comparable to those of first-order DRAG (purple curve of their Fig. 3). Our D=-1.20 solution is about one order of magnitude below the second-order DRAG solution (green curve of their Fig. 4) for $t_p|\Delta|/2\pi\lesssim 3.1$. We conjecture that a combination of second-order DRAG with our optimized Fourier-based wave forms would perform even better.

Wave forms corrected by a derivative term with D=-1.20 produce a good optimum solution. However, one-component optimal control is not that much slower, acceptable for problems where there is access only to σ_z control.

VI. CONCLUSION

We have shown that an optimal control wave form for adiabatic behavior may be obtained by minimizing the spectral weight integrated above a cutoff frequency. Although the best solution is a Slepian, optimization of a few-term Fourier series shows comparably low errors, with acceptable performance achieved with only two or three coefficients. This solution can be used for arbitrary control, even when the transition frequency is changing in time, by first solving the problem with a fixed frequency and then transforming to the proper time frame. We find that effects of finite bandwidth in the control wave form can be readily accounted for by optimizing the Fourier coefficients. This method of solution can be used

for other control problems, such as minimizing the 2-state occupation for a single qubit π pulse.

With this protocol, we find that gates with intrinsic error below 10^{-4} should be possible for the two-qubit controlled-phase operation, even for relatively fast gate times. This theoretical understanding may help open up a route to ultrahigh-fidelity quantum operations and fault-tolerant quantum computation.

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APPENDIX: EXACT SOLUTION

The exact quantum solution to the nonadiabatic problem may be expressed in the form found geometrically. We start by considering the effect of a rotation around the H_x axis. If the wave function in a fixed basis is given by $|\Psi\rangle = a|0\rangle + b|1\rangle$, then it can be rewritten in a basis rotated by the angle θ as $|\Psi\rangle = \alpha|-\rangle + \beta|+\rangle$, reexpressed in eigenstates of the new basis with amplitudes given by

$$\alpha = a\cos(\theta/2) - b\sin(\theta/2),\tag{A1}$$

$$\beta = b\cos(\theta/2) + a\sin(\theta/2). \tag{A2}$$

In the θ -rotated basis we define the energy eigenvalues as $\pm \omega/2$, which introduces phase factors that have differential form $\dot{\alpha} = -i(\omega/2)\alpha$ and $\dot{\beta} = i(\omega/2)\beta$. We are interested in changes in the control vector, so if θ changes with time one finds

$$\dot{\alpha} = -i(\omega/2)\alpha - a\sin(\theta/2)\dot{\theta}/2 - b\cos(\theta/2)\dot{\theta}/2 \quad (A3)$$

$$= (-i\omega\alpha - \dot{\theta}\beta)/2,\tag{A4}$$

Note the symmetry of the rotations, with factors $i\omega$ for the time dependence of H_z rotations and $\dot{\theta}$ for the H_x rotation.

The time dependence can be described simply if one considers the quantity $\alpha^*\beta$, where the asterisk represents the complex conjugate. Its derivative is

$$\frac{d}{dt}(\alpha^*\beta) = (i\omega\alpha^* - \beta^*\dot{\theta})\beta/2 + \alpha^*(i\omega\beta + \alpha\dot{\theta})/2 \quad (A6)$$

$$= i\omega(\alpha^*\beta) + (|\alpha|^2 - |\beta|^2)\dot{\theta}/2 \tag{A7}$$

$$= i\omega(\alpha^*\beta) + \sqrt{1 - 4|\alpha^*\beta|^2} \dot{\theta}/2, \tag{A8}$$

where in the last equation we have used the relation

$$(|\alpha|^2 - |\beta|^2)^2 = |\alpha|^4 + 2|\alpha|^2|\beta|^2 + |\beta|^4 - 4|\alpha^*\beta|^2$$
(A9)

$$= (|\alpha|^2 + |\beta|^2)^2 - 4|\alpha^*\beta|^2$$
 (A10)

$$=1-4|\alpha^*\beta|^2. \tag{A11}$$

Note that the square root factor in Eq. (A8) changes sign when $|\alpha|^2 - |\beta|^2$ is negative.

For Bloch vector angles Θ and φ the amplitudes are $\alpha = \cos(\Theta/2)$ and $\beta = \sin(\Theta/2)e^{i\varphi}$, which gives

$$\alpha^* \beta = \cos(\Theta/2) \sin(\Theta/2) e^{i\varphi} \tag{A12}$$

$$= \sin\Theta \, e^{i\varphi}/2,\tag{A13}$$

so that its magnitude is the error probability $P_e = [\sin(\Theta/2)]^2 \approx |\alpha^*\beta|$ only for small angles Θ . Inserting this result for $\alpha^*\beta$ into Eq. (A8), one finds

$$\frac{d}{dt}(\sin\Theta e^{i\varphi}) = i\omega(\sin\Theta e^{i\varphi}) + \cos\Theta \dot{\theta}. \tag{A14}$$

Changing phase variables $\varphi = \varphi' + \phi$, where $\phi = \int \omega dt$ and noting that $d\phi/dt = \omega$, we can remove the $i\omega$ phase term,

$$\frac{d}{dt}(\sin\Theta e^{i\varphi'}) = \cos\Theta \dot{\theta} e^{-i\phi}, \tag{A15}$$

which is equivalent to the previous geometrical result Eq. (7) in the low-amplitude limit $\Theta \to 0$. Here, the exact solution has the amplitude of the drive term $d\theta/dt$ rescaled by $\cos \Theta$ to account for the geometry of the Bloch sphere.

 $[\]dot{\beta} = (+i\omega\beta + \dot{\theta}\alpha)/2. \tag{A5}$

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