Robust suppression of noise propagation in Gottesman-Kitaev-Preskill error correction

C. Siegele and P. Campagne-Ibarcq [®]

Laboratoire de Physique de l'Ecole normale supérieure, Mines-Paristech, Inria, ENS-PSL, Université PSL, CNRS, Sorbonne Université, Paris, France

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Straightforward logical operations contrasting with complex state preparation are the hallmarks of the bosonic encoding proposed by Gottesman, Kitaev, and Preskill (GKP). The recently reported generation and error correction of GKP qubits in trapped ions and superconducting circuits thus holds great promise for the future of quantum computing architectures based on such encoded qubits. However, these experiments rely on error-syndrome detection via an auxiliary physical qubit, whose noise may propagate and corrupt the encoded GKP qubit. We propose a simple module composed of two oscillators and a physical qubit, operated with two experimentally accessible quantum gates and elementary feedback controls to implement an error-corrected GKP qubit protected from such propagating errors. In the idealized setting of periodic GKP states, we develop efficient numerical methods to optimize our protocol parameters and show that errors of the encoded qubit stemming from flips of the physical qubit and diffusion of the oscillators state in phase space may be exponentially suppressed as the noise strength over individual operations is decreased. Our approach circumvents the main roadblock towards fault-tolerant quantum computation with GKP qubits.

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I. INTRODUCTION

In their seminal paper [1,2], Gottesman, Kitaev, and Preskill (GKP) proposed to encode, within the vast Hilbert space of a harmonic oscillator, a qubit robustly against position and momentum shifts of the embedding oscillator. Clifford operations on encoded GKP qubits are straightforward to implement and do not amplify small shift errors. Therefore, concatenation of the GKP code into the surface code recently attracted interest [3-7] as, beyond the potentially enhanced coherence of GKP qubits compared to faulty physical qubits, analog information from the GKP error-correction layer may be decoded to improve the surface code threshold. Crucially, these desirable features rely on the assumption that noise-induced shifts of the embedding oscillators are short and can be detected before they accumulate. This hypothesis is not valid in current experimental implementations with superconducting circuits [8,9]. In order to comprehend this serious limitation, one needs to delve into the code structure and error-correction techniques employed in these experiments.

In reduced phase-space coordinates (q_a, p_a) ,¹ the basis states of the square GKP code are superpositions of periodically spaced position eigenstates

$$|+Z\rangle = \sum_{n\in\mathbb{Z}} |q_a = n\alpha\rangle, \quad |-Z\rangle = \mathbf{D}_{q_a}\left(\frac{\alpha}{2}\right)|+Z\rangle, \quad (1)$$

where $\alpha = 2\sqrt{\pi}$ and the operator $\mathbf{D}_{r_a}(\delta)$ displaces the oscillator state by δ along r_a , for $r_a = q_a$ or $r_a = p_a$. The logical

states $|\pm X\rangle$ are obtained by a $\pi/2$ rotation in phase space of $|\pm Z\rangle$. Note that infinitely delocalized states are unrealistic, but the essential properties and control techniques considered in our work apply to states normalized by a broad Gaussian envelope in phase space [1,8,10,11]. One may measure the GKP qubit in the $|\pm Z\rangle$ or $|\pm X\rangle$ basis by detecting the *modular logical operators* $\tilde{\mathbf{q}}_{a}^{L} = \mathbf{q}_{a} \mod \alpha$ and $\tilde{\mathbf{p}}_{a}^{L} = \mathbf{p}_{a} \mod \alpha$. Crucially, a code state $|\Psi\rangle$ shifted in position and momentum can still be correctly decoded as long as the shifts are shorter than $\alpha/4$. Moreover, these shifts can be detected without revealing the GKP qubit state by measuring the two commuting *modular stabilizers* $\tilde{\mathbf{q}}_{a}^{L} = \mathbf{q}_{a} \mod \alpha/2$ and $\tilde{\mathbf{p}}_{a}^{L} = \mathbf{p}_{a} \mod \alpha/2$.

Measuring the modular stabilizers without extracting logical information is the main challenge in GKP error correction [12–18]. It was only recently achieved experimentally with trapped ions [11,19,20] and superconducting circuits [8,9]. In these experiments, the target oscillator is coupled to an auxiliary qubit via a controllable Rabi-type interaction static in the interaction picture $-\chi \mathbf{r}_a \boldsymbol{\sigma}_z$ (where $\mathbf{r}_a = \mathbf{q}_a$ or $\mathbf{r}_a = \mathbf{p}_a$ [see Fig. 1(a)]), in order to implement a conditional displacement gate $\mathbf{U}_{r_a}^{\text{CD}} = e^{i\frac{\alpha}{2}\mathbf{r}_a\sigma_z}$ that rotates the qubit phase by $-\alpha \tilde{\mathbf{r}}_a^S$ around the z axis of its Bloch sphere. The gate is named after its backaction on the oscillator, which is displaced by $\pm \frac{\alpha}{2}$ along the $\pi/2$ -rotated quadrature r_a^{\perp} conditioned on the qubit state [see Fig. 1(a)]. This evolution deterministically shifts the logical operator $\tilde{\mathbf{r}}_a^{\perp L}$ by $\alpha/2$, accounted for in software, but otherwise leaves all modular operators unchanged. However, if a bit flip of the qubit occurs during the evolution, for instance due to energy relaxation as in Fig. 1, the displacement takes a value uniformly sampled in $\left[-\frac{\alpha}{2}, \frac{\alpha}{2}\right]$ depending on the unknown instant of the flip (see Appendix A). This randomizes the value of $\tilde{\mathbf{r}}_{a}^{\perp L}$ and the error propagates at the logical level with probability $\frac{1}{2}$. These propagating errors, which become more frequent as the error-correction clock

^{*}philippe.campagne-ibarcq@inria.fr

¹corresponding to the operators \mathbf{q}_a and \mathbf{p}_a with equal fluctuations and verifying ($[\mathbf{q}_a, \mathbf{p}_a] = i$).



FIG. 1. Error propagation during (a) qubit-based error correction. A Rabi-type interaction at rate χ is activated for a duration $T_{\rm CD} = \alpha/(2\chi)$ to map the value of the target modular stabilizer $\tilde{\mathbf{q}}_a^S$ onto the phase of an auxiliary qubit prepared in $|+x\rangle = (|g\rangle +$ $|e\rangle)/\sqrt{2}$ beforehand. As a backaction, the modular logical operator $\tilde{\mathbf{p}}_{a}^{L}$ gets shifted conditioned on the qubit state (conditional trajectories shown in color for an oscillator initially in an eigenstate of $\tilde{\mathbf{p}}_{a}^{L}$ with eigenvalue δ_{p_a}). At T_{CD} and in absence of error, $\tilde{\mathbf{p}}_a^L$ retrieves its initial value but for a deterministic shift by $\alpha/2$. Relaxation of the qubit to $|g\rangle$ at time $(1 - \tau_{err})T_{CD}$ (with $\tau_{err} \in [0, 1]$, gold lightning symbol) propagates as a long shift of $\tilde{\mathbf{p}}_{a}^{L}$ by $\alpha \tau_{\text{err}}$ (dashed trajectory). (b) Steane-type error correction. A quadrature interaction at rate χ' with an auxiliary oscillator b, initially in $|\phi\rangle = \sum_{n} |q_b = n\beta\rangle$, maps the value of $\tilde{\mathbf{q}}_a^S$ onto the modular stabilizer $\tilde{\mathbf{p}}_b$. As a backaction, $\tilde{\mathbf{p}}_a^L$ gets shifted conditioned on the position of the auxiliary oscillator (conditional trajectories shown in color for three position states). At $T_{\text{quad}} = \alpha / (\beta \chi')$ and in absence of error, $\tilde{\mathbf{p}}_a^L$ retrieves its initial value. A shift of $\tilde{\mathbf{q}}_b$ by $\delta_{q_b} \ll \beta$ occurring at $(1 - \tau_{\text{err}})T_{\text{quad}}$ (red lightning symbol) propagates as a short shift of $\tilde{\mathbf{p}}_{a}^{L}$ by $\frac{\delta_{q_{b}}}{\beta} \alpha \tau_{\text{err}}$ (dashed trajectories).

rate increases, are a serious bottleneck towards fault-tolerant quantum computation with GKP qubits. Various strategies were proposed [21–25] and experimentally tested [26,27] to mitigate this advert effect, but either provide only a first-order protection against auxiliary oscillator errors [21,23–25] or rely on a biased-noise auxiliary qubit [22] whose development is not yet mature enough [28,29] to unleash the full potential of GKP qubits.

This roadblock is not present in the so-called Steane-type error-correction scheme [1,30], where the target oscillator is probed via a quadrature interaction $-\chi' \mathbf{r}_a \mathbf{q}_b$ with an auxiliary oscillator *b* to implement a quadrature gate $\mathbf{U}_{r_a}^{\text{quad}} = e^{i\frac{\alpha}{\beta}\mathbf{r}_a \mathbf{q}_b}$. The auxiliary oscillator is itself prepared in a rectangular GKP state

$$|\phi\rangle = \sum_{n \in \mathbb{Z}} |q_b = n\beta\rangle \tag{2}$$

prior to the interaction. Since this state is employed as a displacement sensor [31] and does not encode logical information, we define only modular stabilizers $\tilde{\mathbf{q}}_b = \mathbf{q}_b \mod \beta$ and $\tilde{\mathbf{p}}_b = \mathbf{p}_b \mod 2\pi/\beta$, of whom $|\phi\rangle$ is the single joint eigenstate with eigenvalue 0 [1]. The quadrature gate displaces the auxiliary oscillator along p_b conditioned on the value of \mathbf{r}_a while, reciprocally, the target oscillator is shifted along r_a^{\perp} conditioned on the value of \mathbf{q}_b [see Fig. 1(b)]. We summarize

the gate effect on modular operators as²

$$\frac{\tilde{\mathbf{p}}_b}{2\pi/\beta} \to \frac{\tilde{\mathbf{p}}_b}{2\pi/\beta} + \frac{\tilde{\mathbf{r}}_a^S}{\alpha/2}, \quad \frac{\tilde{\mathbf{r}}_a^{\perp L}}{\alpha} \to \frac{\tilde{\mathbf{r}}_a^{\perp L}}{\alpha} + \frac{\tilde{\mathbf{q}}_b}{\beta}.$$
 (3)

The crucial difference with physical qubit-based error correction lies in the noise model, assumed to only generate short shifts of its state. A shift by δ_{q_b} along \mathbf{q}_b , occurring before or during the gate, propagates to the target oscillator as a shift shorter than $\frac{\delta_{q_b}}{\beta} \alpha$ [$\tau_{\text{err}} \in [0, 1]$ in Fig. 1(b)], correctable if $\delta_{q_b} \ll \beta$. However, if the auxiliary oscillator is prepared through a series of qubit-based measurements of its stabilizers, bit flips of the qubit may induce shifts along q_b covering the whole $\left[-\frac{\beta}{2}, \frac{\beta}{2}\right]$ interval, propagating as shifts of the target oscillator covering $\left[-\frac{\alpha}{2}, \frac{\alpha}{2}\right]$ irrespective of the value of β . In GKP surface code architectures, these structureless shifts cancel the benefits of GKP qubits with respect to physical qubits. Therefore, a central question for the viability of Steane-type error correction is as follows: How can we ensure a supply of auxiliary oscillator states $|\phi\rangle$ whose errors do not propagate as long shifts of the target oscillator?

II. ASYMMETRIC PREPARATION OF THE AUXILIARY OSCILLATOR

We consider the module depicted in Fig. 2(a) where the target oscillator interacts with an auxiliary oscillator, itself coupled to a physical qubit. The target oscillator is corrected by repeated Steane-type correction cycles denoted C_{r_a} , alternating $r_a = q_a$ and $r_a = p_a$ [Fig. 2(b)]. Each cycle starts with the auxiliary oscillator prepared in $|\phi\rangle$ [Fig. 2(c)], possibly shifted due to preparation errors. A quadrature gate $\mathbf{U}_{r_a}^{\text{quad}}$ maps the value of $\tilde{\mathbf{r}}_a^{S}$ onto the stabilizer $\tilde{\mathbf{p}}_b$. The auxiliary oscillator is then measured and reprepared through a sequence of preparation *rounds* labeled \mathcal{R}_{r_b} (for $r_b = q_b$ or $r_b = p_b$). Each round is built around a conditional displacement gate $\mathbf{U}_{r_b}^{\text{CD}}$ mapping the value of $\tilde{\mathbf{r}}_b$ onto the phase of the qubit, prepared beforehand in an eigenstate of σ_x and subsequently measured along σ_{v} [Fig. 2(b)]. Each qubit measurement controls a proportional feedback displacement by $\pm \epsilon$ along r_b . As detailed below, repeated \mathcal{R}_{r_b} rounds corral the auxiliary state toward $\tilde{r}_b = 0$ [8]. We further store the measurement record outputted by the \mathcal{R}_{p_b} rounds as it encodes the value of $\tilde{\mathbf{p}}_b$ following the quadrature gate, i.e., the target error syndrome (see Appendix B4). After straightforward decoding, a corrective feedback displacement is applied to the target oscillator, concluding the correction cycle.

Our proposal to suppress error propagation is based on two observations. First, if the oscillators only interact via the \mathbf{q}_b quadrature operator of the auxiliary one [see Fig. 2(a)], only shifts along this quadrature directly propagate to the target oscillator [second term in Eq. (3)]. As a consequence, the auxiliary state may be asymmetrically prepared, with a focus on avoiding long shifts along q_b , while allowing long shifts along p_b . Admittedly, shifts along p_b blur the extracted errorsyndromes (first term in Eq. (3)), but these errors are mitigated by cycle repetition. Second, during qubit-based preparation

²With the definition $\tilde{\mathbf{p}}_{a}^{\perp L} = -\tilde{\mathbf{q}}_{a}^{L}$.



FIG. 2. (a) In our proposed architecture, the target oscillator a couples to an auxiliary oscillator b via a controlled quadrature interaction. The auxiliary state is prepared and measured via a physical qubit. (b) Alternating C_{q_a} and C_{p_a} correction cycles protect the GKP qubit. (c) A C_{r_a} correction cycle ($r_a = q_a$ or $r_a = p_a$) starts with the auxiliary oscillator prepared in $|\phi\rangle$. The quadrature gate maps the value of the target stabilizer $\tilde{\mathbf{r}}_{a}^{S}$ to the stabilizer $\tilde{\mathbf{p}}_{b}$. The auxiliary state is then measured and prepared for the next cycle by a series of \mathcal{R}_{p_b} rounds *followed* by a series of \mathcal{R}_{q_b} rounds, robustly suppressing propagating errors. The measurement record from \mathcal{R}_{p_h} rounds is summed to estimate the value of $\tilde{\mathbf{p}}_b$ following the quadrature gate as detailed in Appendix B4 (double black lines represent classical communication channels). The result -m controls a displacement by -f(m) on the target oscillator (details on the feedback law f in Appendix D 1). (d) A \mathcal{R}_{r_b} round ($r_b = q_b$ or $r_b = p_b$) starts with the qubit prepared in $|+x\rangle$. A conditional displacement gate maps the value of $\tilde{\mathbf{r}}_b$ to the qubit phase. The final qubit measurement along $\boldsymbol{\sigma}_v$ controls a proportional feedback displacement by $\pm \epsilon$, a conditional flip of the qubit to reset it in $|+x\rangle$, and the outcome is stored for further processing.

of the auxiliary oscillator, flips of the qubit only trigger long shifts along q_b if they occur during \mathcal{R}_{p_b} rounds. Based on these two observations, we propose to prepare the auxiliary state with a large number N_p of \mathcal{R}_{p_b} rounds *followed* by a large number N_q of \mathcal{R}_{q_b} rounds [see Fig. 2(c)], allowing the latter to correct long shifts induced by the former.

The detailed analysis of this preparation sequence is facilitated by the periodicity of the auxiliary state along both quadratures, preserved by the applied controls and by our noise model. This model combines bit and phase flips of the physical qubit, with respective small probabilities p^{BF} and $p^{\rm PF}$ during each round, and quadrature noise of the oscillators at rate κ , equivalent to photon loss and gain at equal rate, inducing uniform state diffusion in phase space. Under these assumptions, we show in Appendix B that the density matrix of the auxiliary oscillator remains diagonal at all times in the Zak basis [32], whose base vectors are GKP states displaced in $\left[-\frac{\beta}{2}, \frac{\beta}{2}\right]$ along q_b and in $\left[-\frac{\pi}{\beta}, \frac{\pi}{\beta}\right]$ along p_b . Thus, its state is encoded by a two-dimensional (2D) wrapped probability distribution, and may be viewed as a classical particle living on a torus with coordinates \tilde{q}_b and \tilde{p}_b . Furthermore, we show that this 2D distribution is separable into two one-dimensional (1D) distributions, respectively defined along \tilde{q}_b and denoted Q_b , and along \tilde{p}_b and denoted *P_b*. In this picture, repeated \mathcal{R}_{r_b} rounds ($\tilde{r}_b = \tilde{q}_b$ or $\tilde{r}_b = \tilde{p}_b$) induce a classical random walk of the particle along \tilde{r}_b , whose steps by $\pm \epsilon$ are biased toward $\tilde{r}_b = 0$. In the limit of short steps, the corresponding R_b distribution evolves with a position-dependent drift velocity $v(\tilde{r}_b) = -\frac{\epsilon p^{NF}}{T_{round}} \sin(2\pi \frac{\tilde{r}_b}{r_0})$ and a uniform diffusion constant $D = \frac{\epsilon^2}{T_{round}} + \kappa$, where T_{round} is the round duration, $p^{NF} = 1 - p^{BF} - 2p^{PF}$ is close to 1 and $r_0 = \beta$ when $\tilde{r}_b = \tilde{q}_b$ and $r_0 = 2\pi/\beta$ when $\tilde{r}_b = \tilde{p}_b$.

The steady-state of this dynamic approaches a wrapped normal distribution whose variance depends on ϵ and reaches a minimum $V_{min} = (\kappa T_{round})^{1/2} r_0 / (2\pi p^{NF})$ for $\epsilon_{min} =$ $(\kappa T_{\text{round}})^{1/2}$. However, the vanishing drift velocity in the vicinity of $\tilde{r}_b = r_0/2$ and small diffusion constant for $\epsilon = \epsilon_{min}$ $(\kappa T_{\text{round}} < 10^{-4} \text{ considered in this work})$ result in a long convergence time and persisting tails of the R_b distribution at this optimal value. We mitigate this advert effect by varying the feedback displacement length ϵ_i as a function of the round index j, starting with $\epsilon_j \sim r_0/2$ to suppress the tails of R_b and ending with $\epsilon_i \sim \epsilon_{\min}$ to limit its central peak width. We exactly compute the evolution of the distributions throughout this preparation, compactly encoded in the form of $(2n_F +$ 1) Fourier coefficient vectors ($n_F \sim 30-60$ throughout this work). In Fig. 3 (top panel), we represent the Q_b distribution obtained after a given number N_q of \mathcal{R}_{q_b} rounds. Its tails are exponentially suppressed as N_q increases while its central peak has a constant variance V_{\min} , ensuring robust suppression of error propagation to the target oscillator.

As the R_b distribution is being sculpted by repeated \mathcal{R}_{r_b} rounds, long shifts triggered by bit flips of the qubit uniformize the distribution along the conjugate quadrature, and quadrature noise deflates its central peak. In our asymmetric preparation scheme, the Q_b distribution is sculpted last and its final value is not impacted by these errors. On the other hand, they have a dramatic effect on P_b which becomes near uniform for large values of N_q (Fig. 3, bottom panel) as the probability $(1 - p_{\rm BF})^{N_q}$ that no bit-flip occurred during the \mathcal{R}_{q_b} rounds approaches 0. Thus, N_q cannot be arbitrarily large for the auxiliary state to be a resource for Steane-type error correction, even in the limit of weak intrinsic noise of the oscillators (see Appendix D 2).

III. TARGET MODE OSCILLATOR CORRECTION

We now consider the evolution of the target oscillator over alternating C_{q_a} and C_{p_a} error-correction cycles. As for the auxiliary oscillator during preparation, the target oscillator state remains periodic (see Appendix B 3). In order to estimate the decay rate of the z component of the GKP qubit Bloch vector κ_{\log} (the x component decays at the same rate and the y component twice faster in the square code) we consider the evolution of the wrapped distribution of the logical operator \tilde{q}_a^L only, denoted Q_a . We compactly represent it as an $(2n_F +$ 1) Fourier coefficient vector and encode the system evolution over a pair of C_{q_a} and C_{p_a} cycles in an $(2n_F + 1) \times (2n_F + 1)$ evolution matrix, which accounts for realistic auxiliary state preparation and $\tilde{\mathbf{p}}_b$ detection (see Fig. 2 and Appendix C for details). The only approximation made in this formalism is to model noise as effective quantum channels applied in-between perfect gates, with negligible impact on the estimate of error-



FIG. 3. Wrapped distributions Q_b and P_b of the stabilizers $\tilde{\mathbf{q}}_b$ and $\tilde{\mathbf{p}}_{b}$ (for a square lattice corresponding to $\beta = \sqrt{2\pi}$) prepared from a uniform distribution by $N_p + N_q$ preparation rounds ($N_p = 50$, varying N_a encoded in color), in presence of quadrature noise at rate $\kappa = (10^5 T_{\text{round}})^{-1}$ and physical qubit flips with probabilities $p^{\text{BF}} =$ $2p^{\rm PF} = 0.002$ per round. Arrows above the top panel schematically represent the drift velocity (single-headed blue arrows) and diffusion constant (double-headed brown arrows) of each distribution R_b $(R_b = Q_b \text{ or } R_b = P_b)$ during corresponding \mathcal{R}_{r_b} rounds. Vanishing drift velocity in the neighborhood of $\tilde{r}_b = \sqrt{\pi}$ results in persisting tails of the distributions. The length of the displacements ϵ_i is varied throughout the preparation to mitigate this effect while maintaining a minimal variance for the central peak (see text, the black dashed line is a Gaussian with variance V_{\min}), ensuring robust suppression of error propagation to the target oscillator. As N_a increases, bit flips of the qubit entail more frequent shifts along p_b elevating the tails of P_b , and the central peak of P_b deflates due to quadrature noise.

correction performances (see Appendix A). Choosing, as an initial guess, a simple *sine* function for the feedback law *f* controlling displacements applied to the target oscillator [see Fig. 2(c)], we observe that the Q_a distribution converges over a few cycles from an arbitrary initial state to a metastable state with two peaks centered at $\tilde{q}_a^L = 0$ and $\tilde{q}_a^L = \alpha/2$ (shown in Fig. 6), as expected from a state close to the GKP code manifold. A slow dynamic then comes into play, following which the respective amplitudes of the two peaks equilibrate as the GKP qubit relaxes to the fully mixed logical state.

For given numbers of preparation rounds N_p and N_q and noise values p^{BF} , p^{PF} , κ , we efficiently extract κ_{log} by spectral analysis of the evolution matrix (see Appendix C 3). Moreover, we adjust the cycle feedback parameters (auxiliary state displacements ϵ_j and Fourier coefficients f_k of a general feedback law f on the target) by gradient ascent in order to minimize κ_{log} . Finally, we select the preparation round number yielding the smallest error rate, assuming a quadrature gate time $T_{quad} = 5T_{round}$: a longer gate time does not impact significantly the performances as long as it does not dominate the overall cycle duration. In Fig. 4, we report the rate κ_{log} obtained after this optimization. Strikingly, κ_{log} decreases exponentially as the system noise strength, or equivalently the



FIG. 4. Decay rate of the *z* component of the GKP qubit Bloch vector κ_{\log} as a function of the oscillator's quadrature noise rate κ and of the probability for flips of the physical qubit during each round $p^{BF} = 2p^{PF}$, linearly varied. For each noise value, the number of preparation rounds N_q and N_p are swept together (allowing different values did not significantly improve error-correction performances) and the cycle feedback parameters optimized by gradient ascent. We report the minimum error rate with the corresponding round number, encoded in color. The red circle marks state-of-the-art hardware parameters for $T_{round} = 1.5 \,\mu$ s (see text).

gate duration, decreases. This is in stark contrast with the linear scaling found for simple qubit-based error correction, even considering multimode GKP codes [33], and, following the argument given at the end of Sec. I, for Steane-type error correction with a symmetrically prepared auxiliary state. Assuming that the protocol presented in this work may be adapted to protect finite-energy GKP states against photon loss with similar performances, we find that for a state-of-the-art system (30 ms photon lifetime in the oscillators [34], $T_1 \sim T_2 \sim 500 \,\mu s$ for the physical qubit [35,36] and a preparation round duration of 1.5 μs [8,37]), the coherence time of the GKP qubit could surpass that of the embedding hardware by two orders of magnitude (red circle in Fig. 4).

IV. CONCLUSION AND OUTLOOK

In this paper, we proposed a simple architecture controlled with two elementary gates to robustly protect an encoded GKP qubit. The conditional displacement gate is now routinely employed in superconducting circuit experiments. As for the quadrature gate, it may be decomposed into a sequence of single-mode squeezing gates and a beam-splitter gate [16,30,38] or enabled by activating simultaneously a beamsplitter Hamiltonian and a two-mode squeezing Hamiltonian [39]. Recent progress toward the implementation of these operations in parametrically driven superconducting circuits [40,41] gives reason to hope that a quadrature gate suited for continuous-variable quantum computing will be experimentally demonstrated in the near future. This gate will anyhow be needed to perform operations on encoded GKP qubits. In that sense, the module we consider does not unnecessarily increase the complexity of a GKP qubit-based computing platform.

Numerical simulations assuming a simplified noise model indicate that the lifetime of the GKP qubit protected by our protocol is exponentially enhanced as the noise strength during each gate decreases. Extending this result to normalized GKP code states and more realistic noise models will be the subject of future work. In particular, we have not considered the impact of photon loss nor of imperfect quadrature gates in this work, but do not expect these errors to qualitatively impact our results if they can be mapped to short displacements of the target oscillator state during each correction cycle. While entering the regime of strong suppression of logical errors requires both oscillators and the qubit to be at the state of the art, a substantial margin for improvement exists by refining the feedback law beyond the short memory model considered in this work, allowing longer conditional displacements whose lengths are multiples of the GKP lattice period [12,15], and considering more extensive hardware with multiple auxiliary oscillators and physical qubits in order to multiplex error-syndrome detection. Given that Clifford operations in the GKP code rely on the same controls considered in this letter, our work opens a clear path toward fault-tolerant quantum computation with GKP qubits.

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APPENDIX A: EFFECTIVE NOISE CHANNELS

In this work, we consider a simplified noise model in which instantaneous and perfect conditional displacement and quadrature gates are followed by the application of an effective noise channel accounting for the errors having occurred during the gates. This effective model considerably reduces numerical simulation complexity, but leads to approximations. In this section, we describe the effective noise channel applied after each gate, and argue that these approximations should not impact significantly the error-correction performances estimated for our protocol.

1. Bit flips of the physical qubit

We consider bit flips induced by qubit relaxation and excitation at respective rates Γ_+ and Γ_- . Their effect on the system density matrix is modeled by Lindblad dissipators $\sqrt{\Gamma_+}\mathcal{D}[\sigma_+]$ and $\sqrt{\Gamma_-}\mathcal{D}[\sigma_-]$, where σ_+ and σ_- are, respectively, the raising and lowering operators of the qubit. Each dissipator $\mathcal{D}[\mathbf{L}]$ yields, over an infinitesimal time step dt, an evolution of the density matrix of the system

$$d\boldsymbol{\rho} = dt \,\mathcal{D}[\mathbf{L}](\boldsymbol{\rho}) = dt \left(\mathbf{L}\boldsymbol{\rho}\mathbf{L}^{\dagger} - \frac{1}{2}(\mathbf{L}^{\dagger}\mathbf{L}\boldsymbol{\rho} + \boldsymbol{\rho}\mathbf{L}^{\dagger}\mathbf{L})\right). \quad (A1)$$

We focus on the case $\Gamma_+ = \Gamma_- = \Gamma_1/2$ and briefly describe the most general case $\Gamma_+ \neq \Gamma_-$ at the end of this section. Note that this particular case of equal rates of qubit excitation and relaxation applies to current experiments with superconducting circuits, as argued below.

We consider the effect of bit flips during the application of a conditional displacement gate $\mathbf{U}_{q_b}^{\text{CD}} = e^{i\frac{\pi}{\beta}\mathbf{q}_b\sigma_z}$ along the p_b

quadrature of an oscillator: the calculation is directly adaptable to the case of a conditional displacement along q_b . We assume this gate to be performed by the application of a Rabi-type Hamiltonian, also known as longitudinal coupling Hamiltonian

$$\mathbf{H}_{q_b}^{\text{CD}} = -\chi \mathbf{q}_b \boldsymbol{\sigma}_z \tag{A2}$$

with constant rate $\chi = \frac{\pi}{\beta T_{\rm CD}}$ over the gate duration $T_{\rm CD}$ (the coupling Hamiltonian is then turned off until the following gate). We unravel the effect of bit flips as stochastic collapses onto the ground state $|g\rangle$ or the excited state $|e\rangle$ [42]. Intuitively, the trajectories so unraveled are obtained by detecting, with perfect efficiency, photon emission into the environment, and photon absorption from the environment. At each infinitesimal time step dt, the Kraus operators of this evolution are $\mathbf{P}^- = \sqrt{\frac{\Gamma_1 dt}{2}} |g\rangle \langle e|$, $\mathbf{P}^+ = \sqrt{\frac{\Gamma_1 dt}{2}} |e\rangle \langle g|$, respectively, modeling a jump to $|g\rangle$ or $|e\rangle$ and the no-jump operator $\mathbf{1} - i\mathbf{H}^{\rm NJ} dt$ where

$$\mathbf{H}^{\mathrm{NJ}}dt = \mathbf{H}_{q_b}^{\mathrm{CD}}dt - \frac{i}{2}(\mathbf{P}^{-\dagger}\mathbf{P}^- + \mathbf{P}^{+\dagger}\mathbf{P}^+)$$
$$= \mathbf{H}_{q_b}^{\mathrm{CD}}dt - \frac{i\Gamma_1dt}{4}\mathbf{1}$$
(A3)

may be viewed as a non-Hermitian Hamiltonian.

We place ourselves in the weak noise limit ($\Gamma_1 T_{CD} \ll 1$) and neglect the possibility of trajectories with two jumps. The evolution through the gate in absence of jumps is given by the operator

$$\mathbf{O}^{\mathrm{NJ}} = e^{-\frac{\Gamma_1 T_{\mathrm{CD}}}{4}} \mathbf{U}_{q_b}^{\mathrm{CD}}.$$
 (A4)

and, from a state encoded by the density matrix ρ , the system evolves to the non-normalized density matrix

$$\boldsymbol{\rho}^{\mathrm{NJ}} = \mathbf{O}^{\mathrm{NJ}} \boldsymbol{\rho} \, \mathbf{O}^{\mathrm{NJ}\dagger} = e^{-\frac{\Gamma_{1} T_{\mathrm{CD}}}{2}} \mathbf{U}_{q_{b}}^{\mathrm{CD}} \boldsymbol{\rho} \mathbf{U}_{q_{b}}^{\mathrm{CD}\dagger}$$
(A5)

We find the probability of no bit-flip having occurred to be $1 - p^{BF} = e^{-\frac{\Gamma_1 T_{CD}}{2}} \simeq 1 - \frac{\Gamma_1 T_{CD}}{2}$, and the normalized density matrix conditioned on no jump having occurred is the same as for an evolution through a perfect gate. We detail the evolution of the auxiliary state through the remaining steps of the preparation round in Appendix B 2 in this no-jump case. Note that the value of p^{BF} is simply understood as the probability of a qubit excitation and relaxation over a small time step dt being $n_e \Gamma_- dt + (1 - n_e) \Gamma_+ dt = \Gamma_1 dt/2$, where n_e is the expectation value of $|e\rangle \langle e|$.

We now focus on the trajectories during which the qubit has flipped. The evolution through the gate when the qubit excites or relaxes in a time interval of duration dt around t(with $0 < t < T_{CD}$) reads as

$$\mathbf{O}_{t}^{\pm} = e^{-\frac{\Gamma_{1}(T_{\text{CD}}-t)}{4}} e^{i\chi(T_{\text{CD}}-t)\mathbf{q}_{b}\sigma_{z}} \mathbf{P}^{\pm} e^{-\frac{\Gamma_{1}t}{4}} e^{i\chi t \mathbf{q}_{b}\sigma_{z}}$$
$$= e^{-\frac{\Gamma_{1}T_{\text{CD}}}{4}} e^{\mp i\frac{\pi}{\beta}(1-\frac{2t}{T_{\text{CD}}})\mathbf{q}_{b}} \mathbf{P}^{\pm}$$
$$= e^{-\frac{\Gamma_{1}T_{\text{CD}}}{4}} \mathbf{D}_{t}^{\mp} \mathbf{P}^{\pm}, \tag{A6}$$

where we have defined $\mathbf{D}_{l}^{\mp} = \mathbf{D}_{p_{b}}[\mp \frac{\pi}{\beta}(1 - \frac{2t}{T_{CD}})]$. \mathbf{O}_{l}^{\pm} thus collapses the qubit onto $|g\rangle$ or $|e\rangle$ and displaces the oscillator state along p_{b} by $\pm \frac{\pi}{\beta}(1 - \frac{2t}{T_{CD}}) \in I = [-\frac{\pi}{\beta}, \frac{\pi}{\beta}]$. In the protocol described in this work, conditional displacement gates

employed for auxiliary state preparation are immediately followed by a measurement of the σ_y Pauli operator of the qubit whose outcome controls a feedback displacement by $\pm \epsilon$ along q_b and a qubit rotation resetting it in $|+x\rangle$ [see Fig. 2(d)]. Recombining all trajectories during which the qubit has flipped, the system state then reads as

$$\boldsymbol{\rho}^{J,\mathrm{FB}} = e^{-\frac{\Gamma_{1}T_{\mathrm{CD}}}{2}} \sum_{\substack{s=+,-\\u=+,-}} \int_{t=0}^{T_{\mathrm{CD}}} \mathbf{D}_{q_{b}}(s\epsilon) |+x\rangle \langle sy|\mathbf{O}_{t}^{u}\boldsymbol{\rho}\mathbf{O}_{t}^{u^{\dagger}}|sy\rangle \langle +x|\mathbf{D}_{q_{b}}(s\epsilon)^{\dagger}$$

$$= e^{-\frac{\Gamma_{1}T_{\mathrm{CD}}}{2}} \sum_{s=+,-} \int_{t=0}^{T_{\mathrm{CD}}} \frac{\Gamma_{1}dt}{2} \mathbf{D}_{q_{b}}(s\epsilon) |+x\rangle \langle sy|(\mathbf{D}_{t}^{-}|e\rangle \langle g|\boldsymbol{\rho}|g\rangle \langle e|\mathbf{D}_{t}^{-\dagger} + \mathbf{D}_{t}^{+}|g\rangle \langle e|\boldsymbol{\rho}|e\rangle \langle g|\mathbf{D}_{t}^{+\dagger}\rangle |sy\rangle \langle +x|\mathbf{D}_{q_{b}}(s\epsilon)^{\dagger}$$

$$= e^{-\frac{\Gamma_{1}T_{\mathrm{CD}}}{2}} \sum_{s=+,-} \int_{t=0}^{T_{\mathrm{CD}}} \frac{\Gamma_{1}dt}{2} \mathbf{D}_{q_{b}}(s\epsilon) |+x\rangle \langle sy|\mathbf{D}_{t}^{-}(|e\rangle \langle g|\boldsymbol{\rho}|g\rangle \langle e| + |g\rangle \langle e|\boldsymbol{\rho}|e\rangle \langle g|\mathbf{D}_{t}^{-\dagger}|sy\rangle \langle +x|\mathbf{D}_{q_{b}}(s\epsilon)^{\dagger}$$

$$= e^{-\frac{\Gamma_{1}T_{\mathrm{CD}}}{2}} \sum_{s=+,-} \int_{t=0}^{T_{\mathrm{CD}}} \frac{\Gamma_{1}dt}{4} \mathbf{D}_{q_{b}}(s\epsilon) \mathbf{D}_{t}^{-} |+x\rangle (\langle g|\boldsymbol{\rho}|g\rangle + \langle e|\boldsymbol{\rho}|e\rangle) \langle +x|\mathbf{D}_{t}^{-\dagger}\mathbf{D}_{q_{b}}(s\epsilon)^{\dagger}$$

$$= e^{-\frac{\Gamma_{1}T_{\mathrm{CD}}}{2}} \sum_{s=+,-} \int_{t=0}^{T_{\mathrm{CD}}} \frac{\Gamma_{1}dt}{4} \mathbf{D}_{q_{b}}(s\epsilon) \mathbf{D}_{t}^{-} |\mathbf{\rho}_{b}\mathbf{D}_{t}^{-\dagger}\mathbf{D}_{q_{b}}(s\epsilon)^{\dagger} \otimes |+x\rangle \langle +x|, \qquad (A7)$$

where we have used that $\mathbf{D}_{T_{CD}-t}^+ = \mathbf{D}_t^-$ from the second to the third line, and $\boldsymbol{\rho}_b$ denotes the density matrix of the auxiliary state found after tracing out the qubit mode on the last line. Thus, when a qubit flip occurs, which happens with probability $e^{-\frac{\Gamma_1 T_{CD}}{2}} \frac{\Gamma_1 T_{CD}}{2} \simeq p^{\text{BF}}$, the auxiliary state is randomly shifted in the whole interval *I* along p_b , and randomly shifted by $\pm \epsilon$ along q_b .

We now argue that the *echoed conditional displacement* gate employed in current superconducting circuit experiments [8,9], performed on a qubit affected by relaxation only ($\Gamma_{-} = \Gamma_{1}, \Gamma_{+} = 0$), is equivalent to a simple conditional displacement gate performed on a qubit affected by excitation and relaxation at equal rates ($\Gamma_{+} = \Gamma_{-} = \Gamma_{1}/2$). The echoed conditional displacement includes a π rotation of the qubit around σ_{y} together with a change of sign for the oscillator-qubit interaction at $T_{CD}/2$:

$$\mathbf{H}_{q_b}^{\text{ECD}} = \begin{cases} \mathbf{H}_{q_b}^{\text{CD}} & \text{if } t < \frac{T_{\text{CD}}}{2}, \\ -\mathbf{H}_{q_b}^{\text{CD}} & \text{if } t > \frac{T_{\text{CD}}}{2}. \end{cases}$$
(A8)

Thus, for a noiseless qubit (perfect gate), the evolution reads as

$$\mathbf{U}_{q_b}^{\text{ECD}} = e^{-i\frac{\pi}{2\beta}\mathbf{q}_b\sigma_z} e^{-i\frac{\pi}{2}\sigma_y} e^{i\frac{\pi}{2\beta}\mathbf{q}_b\sigma_z}$$

$$= e^{-i\frac{\pi}{2}\sigma_y} \mathbf{U}_{q_b}^{\text{CD}}.$$
(A9)

Formally, the final π rotation around σ_y does not impact the subsequent qubit measurement along σ_y so that, after applying a feedback displacement and tracing out the qubit, this evolution is equivalent to that obtained after a simple conditional displacement gate.

Here again, we unravel the effect of qubit relaxation at rate $\Gamma_{-} = \Gamma_{1}$ as stochastic jumps, assuming that the jump probability is small ($\Gamma_{1}T_{CD} \ll 1$). In particular, this implies that double-jump trajectories may be neglected. At each infinitesimal time step dt, the Kraus operators are $\sqrt{\Gamma_{1}dt}|g\rangle\langle e| = \sqrt{2}\mathbf{P}^{-}$ and the corresponding no-jump operator $\mathbf{1} - i\mathbf{\tilde{H}}^{NJ}dt$ with

$$\tilde{\mathbf{H}}^{\text{NJ}}dt = \mathbf{H}_{q_b}^{\text{ECD}}dt - i\mathbf{P}^{-\dagger}\mathbf{P}^{-}$$

$$= \mathbf{H}_{q_b}^{\text{ECD}}dt - \frac{i\Gamma_1}{4}(\mathbf{1} - \boldsymbol{\sigma}_z)dt.$$
(A10)

First focusing the no-jump evolution, it reads as

$$\mathbf{O}^{\text{ECD,NJ}} = e^{-\frac{\Gamma_1 T_{\text{CD}}}{8}(1-\sigma_z)} e^{-i\frac{\pi}{2\beta} \mathbf{q}_b \sigma_z} e^{-i\frac{\pi}{2} \sigma_y} e^{-\frac{\Gamma_1 T_{\text{CD}}}{8}(1-\sigma_z)} e^{i\frac{\pi}{2\beta} \mathbf{q}_b \sigma_z}$$
$$= e^{-i\frac{\pi}{2} \sigma_y} e^{-\frac{\Gamma_1 T_{\text{CD}}}{4}} \mathbf{U}_{q_b}^{\text{CD}},$$
(A11)

where the various terms have simply been ordered using that the two terms in the non-Hermitian Hamiltonian (A10) commute. Thus, we recover the same no-jump evolution as for the simple conditional displacement gate. Now focusing on the case where a jump occurred at time t during the gate, the evolution operator reads as

$$\tilde{\mathbf{O}}_{t} = \begin{cases} \tilde{\mathbf{O}}_{t}^{<} = \sqrt{2} e^{-i\frac{\pi}{2\beta} \mathbf{q}_{b} \sigma_{z}} e^{-\frac{\Gamma_{1} T_{\text{CD}}}{8} (1-\sigma_{z})} e^{-i\frac{\pi}{2} \sigma_{y}} e^{i\frac{\pi}{2\beta} (1-\frac{2t}{T_{\text{CD}}}) \mathbf{q}_{b} \sigma_{z}} e^{-\frac{\Gamma_{1} (T_{\text{CD}}-2t)}{8} (1-\sigma_{z})} \mathbf{P}^{-} e^{i\frac{\pi}{2\beta} \frac{2t}{T_{\text{CD}}} \mathbf{q}_{b} \sigma_{z}} e^{-\frac{\Gamma_{1} 2t}{8} (1-\sigma_{z})} & \text{if } t < \frac{T_{\text{CD}}}{2} , \\ \tilde{\mathbf{O}}_{t}^{>} = \sqrt{2} e^{-i\frac{\pi}{2\beta} (2-\frac{2t}{T_{\text{CD}}}) \mathbf{q}_{b} \sigma_{z}} e^{-\frac{\Gamma_{1} (2T_{\text{CD}}-2t)}{8} (1-\sigma_{z})} \mathbf{P}^{-} e^{-i\frac{\pi}{2\beta} (\frac{2t}{T_{\text{CD}}} -1) \mathbf{q}_{b} \sigma_{z}} e^{-\frac{\Gamma_{1} (2t-T_{\text{CD}})}{8} (1-\sigma_{z})} e^{-i\frac{\pi}{2} \sigma_{y}} e^{i\frac{\pi}{2\beta} \mathbf{q}_{b} \sigma_{z}} e^{-\frac{\Gamma_{1} T_{\text{CD}}}{8} (1-\sigma_{z})} & \text{if } t > \frac{T_{\text{CD}}}{2} . \end{cases}$$
(A12)

After commuting the projectors and the qubit rotation operator through the conditional displacements, we find that

$$\tilde{\mathbf{O}}_{t} = \begin{cases} \tilde{\mathbf{O}}_{t}^{<} = \sqrt{2}e^{-\frac{\Gamma_{1}t}{2}}e^{-i\frac{\pi}{2}\sigma_{y}}\mathbf{O}_{t}^{-} & \text{if } t < \frac{T_{\text{CD}}}{2}, \\ \tilde{\mathbf{O}}_{t}^{>} = -\sqrt{2}e^{-\frac{\Gamma_{1}(t-T)}{2}}e^{-i\frac{\pi}{2}\sigma_{y}}\mathbf{O}_{t}^{+} & \text{if } t > \frac{T_{\text{CD}}}{2}, \end{cases}$$
(A13)

where \mathbf{O}^{\pm} is defined as in Eq. (A6). $\tilde{\mathbf{O}}_t$ thus collapses the qubit onto $|e\rangle$ and displaces the oscillator state along p_b by $\frac{\pi}{\beta}(1-2\frac{t}{T_{CD}}) \in I_{>} = [0, \frac{\pi}{\beta}]$ if $t < \frac{T_{CD}}{2}$, and collapses the qubit onto $|g\rangle$ and displaces the oscillator state along p_b by $-\frac{\pi}{\beta}(1-2\frac{t}{T_{CD}}) \in I_{>}$ if $t > \frac{T_{CD}}{2}$. Notice that for a periodic state with period $\frac{2\pi}{\beta}$, the former displacement is equivalent to a displacement by $-\frac{\pi}{\beta}(1+2\frac{t}{T_{CD}}) \in I_{<}$. In both cases, measuring the σ_y Pauli operator of the qubit following the evolution yields a random outcome, and thus a random feedback displacement by $\pm \epsilon$ along q_b . The integrated probability for the trajectories during which a jump occurred is, at first order in $\Gamma_1 T_{CD} - \Gamma_1 T_{CD}/2 = p^{BF}$, and the non-normalized density matrix conditioned on a jump having occurred is the same as the one found in Eq. (A7), at first order in $\Gamma_1 T_{CD}$.

The most notable difference with the case of a simple conditional displacement in presence of transmon excitation and relaxation at equal rates lies in the distribution of displacements entailed by a jump. After the qubit measurement and feedback, the non-normalized system state conditioned on a jump having occurred differs from Eq. (A7) at second order

$$\boldsymbol{\rho}^{\text{ECD},J,\text{FB}} = e^{-\frac{\Gamma_{1}T_{\text{CD}}}{2}} |+x\rangle \langle +x| \sum_{s=+,-} \int_{t=0}^{\frac{T_{\text{CD}}}{2}} \frac{\Gamma_{1}dt}{2} \\ \times \cosh(\Gamma_{1}t) \mathbf{D}_{q_{b}}(s\epsilon) \mathbf{D}_{t}^{+} \boldsymbol{\rho}_{b} \mathbf{D}_{t}^{+\dagger} \mathbf{D}_{q_{b}}(s\epsilon)^{\dagger}, \quad (A14)$$

where the $\cosh(\Gamma_1 t)$ slightly favors shortly displaced states over states displaced by a long distance (the oscillator modular position is no longer uniformly sampled in *I*). We insist that this correction is of second order in $\Gamma_1 T_{CD}$, and is given here to show that such second-order correction should not impact significantly the performance of our protocol.

To conclude this section, we mention that in the case where a simple conditional displacement is applied to a qubit in presence of excitation and relaxation at different rates, one finds a modified no-jump evolution by which the qubit partially collapses onto $|g\rangle$. This partial collapse slightly unbalances the relative amplitudes of probability for the two conditionally displaced copies of ρ , thereby reducing the contrast of the subsequent qubit measurement. The impact on the performances of our protocol is expected to be similar to that of the qubit phase flips, which is described in the next section.

2. Phase flips of the physical qubit

By comparison with bit flips, phase flips of the qubit are simpler to model. Indeed, in the quantum trajectory approach described above, they correspond to σ_z gates randomly applied to the qubit over any time interval of duration dt with probability $\frac{\Gamma_{\phi}}{2}dt$, where Γ_{ϕ} is the qubit pure dephasing time. Since σ_z commutes with the interaction Hamiltonian, phase flips are equivalently modeled as a σ_z gate applied after the gate with probability $p^{\text{PF}} = \frac{\Gamma_{\phi}}{2}T_{\text{CD}}$ (in the weak noise limit). By flipping the sign of the subsequently measured σ_y Pauli operator, this error results in an erroneously applied feedback displacement. We set $p^{\text{PF}} = p^{\text{BF}}/2$, typical of superconducting circuit experiments, in all simulations performed in this work.

Note that qubit readout errors have an impact similar to phase flips, but may cause more damage when the qubit is actively reset based on the measurement outcome, yielding a qubit erroneously prepared in $|-x\rangle$ for the subsequent auxiliary state preparation round [see Fig. 2(d)]. Experimentally, such reset errors may be mitigated by repeating the reset procedure in $|g\rangle$, assuming the measurement to be quantum nondemolition for the $|g\rangle$ state [43,44]. Readout errors are not modeled in this work.

3. Quadrature noise

Quadrature noise at rate κ is modeled by two Lindblad dissipators $\sqrt{\kappa}\mathcal{D}[\mathbf{q}]$ and $\sqrt{\kappa}\mathcal{D}[\mathbf{p}]$, inducing uniform diffusion of an oscillator state in phase space. Its effect can equivalently be modeled by the application of stochastic evolution operators

$$\mathbf{U}_{dt}^{p} = e^{i\sqrt{\kappa}dW_{q}\mathbf{q}},
\mathbf{U}_{dt}^{p} = e^{i\sqrt{\kappa}dW_{p}\mathbf{p}},$$
(A15)

where dW_q and dW_p are independent Wiener processes characterized by $\overline{dW_q} = \overline{dW_p} = 0$ and $dW_q^2 = dW_p^2 = dt$ [42].

a. Effective noise channel after a conditional displacement gate

We here consider a conditional displacement gate applied on the q quadrature of the auxiliary oscillator reading

$$\mathbf{U}_{a}^{\mathrm{CD}} = e^{i\frac{\pi}{\beta}\mathbf{q}\boldsymbol{\sigma}_{z}},\tag{A16}$$

where we dropped the subscript b to designate the auxiliary oscillator quadrature. It is straightforward to adapt the following calculation to the case of a conditional displacement along the p quadrature.

When the gate is applied in finite time T_{CD} and in presence of quadrature noise, we use Trotter decomposition over $N = \frac{T_{\text{CD}}}{dt} \gg 1$ steps to write the stochastic evolution over a single trajectory

$$\tilde{\mathbf{U}}_{q_a}^{\text{CD}} = \prod_{j=1}^{N} \left(e^{i\frac{\pi}{N\beta} \mathbf{q} \boldsymbol{\sigma}_z} e^{i\sqrt{\kappa} dW_q^j \mathbf{q}} e^{i\sqrt{\kappa} dW_p^j \mathbf{p}} \right), \tag{A17}$$

where all Wiener processes dW_q^j , dW_p^j are independent. Using Baker-Campbell-Hausdorff formula, we reorder this product to put the noise terms in front:

$$\begin{split} \tilde{\mathbf{U}}_{q_a}^{\text{CD}} &= \prod_{j=1}^{N} \left(e^{-i\sqrt{\kappa}dW_p^j \frac{j\pi}{N\beta}\boldsymbol{\sigma}_z} \right) \\ &\times \prod_{j=1}^{N} \left(e^{i\sqrt{\kappa}dW_q^j \mathbf{q}} e^{i\sqrt{\kappa}dW_p^j \mathbf{p}} \right) \prod_{j=1}^{N} \left(e^{i\frac{\theta}{N}\mathbf{q}\boldsymbol{\sigma}_z} \right). \end{split}$$
(A18)

The center and rightmost products correspond to a quadrature noise channel applied for a duration T_{CD} after an error-free conditional displacement gate \mathbf{U}_q^{CD} , corresponding to our effective noise model. We thus neglect the leftmost term, which rotates the qubit Bloch vector around its σ_z axis conditioned on the value of dW_p^J , i.e., the particular value of the q shifts induced by noise during the interaction. Its physical interpretation is clear: q shifts of the oscillator that occur at the beginning of the gate $(j \rightarrow N)$ leave an imprint on the qubit phase similarly to shifts having occurred before the gate, while shifts that occur toward the end of the gate $(i \rightarrow 1)$ impact negligibly the qubit phase. Thus, if we were to exactly model the system evolution with all the terms in Eq. (A18), the recentering feedback displacement following the gate and qubit readout would partially correct for q-quadrature noise during the gate. By neglecting the leftmost term in Eq. (A18), we carry over this noise to the next round, resulting in a slightly broadened *q*-probability distribution for the auxiliary state. Similar arguments can be made for the p distribution. For the noise figures considered in this work ($\kappa T_{CD} < \kappa T_{round} <$ 10^{-4}), we expect this approximation to have a marginal impact on the estimated performances of our error-correction protocol, which are only slightly underestimated.

b. Effective noise channel after a quadrature gate

We follow a similar reasoning for the quadrature gate

$$\mathbf{U}_{q_a}^{\text{quad}} = e^{i\theta \mathbf{q}_a \mathbf{q}_b},\tag{A19}$$

where $\theta = \alpha/\beta$. Here, for simplicity, we consider q_a and q_b quadrature noise only, corresponding to stochastic terms of the form $e^{i\sqrt{\kappa}dW_{p_a}\mathbf{p}_a}$ and $e^{i\sqrt{\kappa}dW_{p_b}\mathbf{p}_b}$: the terms inducing shifts along p_a and p_b commute trivially through the gate. We decompose the noisy gate over $N = \frac{T_{\text{quad}}}{dt} \gg 1$ steps as

$$\tilde{\mathbf{U}}_{q}^{\text{quad}} = \prod_{j=1}^{N} \left(e^{i\frac{\theta}{N} \mathbf{q}_{a} \mathbf{q}_{b}} e^{i\sqrt{\kappa} dW_{p_{a}}^{j} \mathbf{p}_{a}} e^{i\sqrt{\kappa} dW_{p_{b}}^{j} \mathbf{p}_{b}} \right), \tag{A20}$$

where all Wiener processes $dW_{p_a}^j$, $dW_{p_b}^j$ are independent. Using Baker-Campbell-Hausdorff formula, we reorder this product to place the noise terms in front:

$$\begin{split} \tilde{\mathbf{U}}_{q}^{\text{quad}} &= e^{i\phi} \prod_{j=1}^{N} \left(e^{i\sqrt{\kappa}dW_{p_{a}}^{j}\mathbf{p}_{a}} e^{-i\sqrt{\kappa}dW_{p_{b}}^{j}\frac{j\theta}{N}\mathbf{q}_{a}} \right) \\ &\times \prod_{j=1}^{N} \left(e^{i\sqrt{\kappa}dW_{p_{b}}^{j}\mathbf{p}_{b}} e^{-i\sqrt{\kappa}dW_{p_{a}}^{j}\frac{j\theta}{N}\mathbf{q}_{b}} \right) \prod_{j=1}^{N} \left(e^{i\frac{\theta}{N}\mathbf{q}_{a}\mathbf{q}_{b}} \right) \end{split}$$

$$= e^{i\phi'} \prod_{j=1}^{N} e^{-i\sqrt{\kappa}dW_{p_b}^j \frac{j\theta}{N}} \mathbf{q}_a \prod_{j=1}^{N} e^{-i\sqrt{\kappa}dW_{p_a}^j \frac{j\theta}{N}} \mathbf{q}_b$$
$$\times \prod_{j=1}^{N} e^{i\sqrt{\kappa}dW_{p_a}^j} \mathbf{p}_a \prod_{j=1}^{N} e^{i\sqrt{\kappa}dW_{p_b}^j} \mathbf{p}_b \prod_{j=1}^{N} e^{i\frac{\theta}{N}} \mathbf{q}_a \mathbf{q}_b, \quad (A21)$$

where ϕ and ϕ' are irrelevant global phases that can be omitted. Our simplified model includes the fifth product (equal to the noiseless evolution $\mathbf{U}_{q_a}^{ ext{quad}}$) and the third and fourth products (corresponding to quadrature noise channels applied on idling oscillators). The first product corresponds to an extra p_a quadrature noise term, correlated to the auxiliary oscillator q_b noise during the gate. Since, in our protocol, we discard the result of measurements of the modular stabilizer $\tilde{\mathbf{q}}_{b}$ and reset the auxiliary state at the end of each cycle, this correlated noise boils down to random displacements of the target state along p_a , with zero mean value and variance $\overline{(\frac{\sqrt{k}\theta}{N}\sum_j jdW_{p_b}^j)^2} \rightarrow$ $\kappa \frac{\theta^2}{3} T_{\text{quad}}$, where the last limit is taken for $N \to \infty$. We account for this term by renormalizing the p_a quadrature noise rate during the gate following $\kappa \to \kappa (1 + \frac{\theta^2}{3})$. Note that its effect could be partially mitigated by decoding the information yielded by the $\tilde{\mathbf{q}}_b$ measurements at the end of the cycle. Similarly, the second product describes shifts along p_b correlated to the target oscillator q_a noise during the gate. In analogy to the case of the conditional displacement gate detailed in the previous section, it is interpreted as partial mapping of the target oscillator shift errors occurring during the gate onto the auxiliary oscillator (shifts occurring at the beginning of the evolution leave a stronger imprint than those occurring toward the end). By neglecting this term in our simplified model, we carry over to the following cycle errors that would have been partly corrected in a more accurate model, and thereby expect to slightly underestimate the performances of our protocol.

APPENDIX B: OSCILLATOR DYNAMICS IN THE ZAK BASIS

1. The Zak basis

The dynamics of our system is conveniently described in the Zak basis [32] of the oscillators, which is the basis formed by displaced GKP states within one unit cell of the GKP lattice. Equivalently, the Zak basis we will consider for the target oscillator can be seen as the joint eigenbasis of the modular logical operator $\tilde{\mathbf{q}}_a^L$ and of the modular stabilizer $\tilde{\mathbf{p}}_a^S$, and the Zak basis for the auxiliary oscillator as the joint eigenbasis of the modular stabilizers $\tilde{\mathbf{q}}_b$ and $\tilde{\mathbf{p}}_b$. Formally, the Zak states are defined as

$$|u, v\rangle_{a} = e^{-iu\mathbf{p}_{a}+iv\mathbf{q}_{a}}|+Z\rangle = e^{\frac{i}{2}uv}\sum_{n\in\mathbb{Z}}e^{inv\alpha}|n\alpha+u\rangle_{q_{a}},$$
$$|u', v'\rangle_{b} = e^{-iu'\mathbf{p}_{b}+iv'\mathbf{q}_{b}}|\phi\rangle = e^{\frac{i}{2}u'v'}\sum_{m\in\mathbb{Z}}e^{imv'\beta}|m\beta+u'\rangle_{q_{b}},$$
(B1)

where we use the convention $u \in [-\frac{\alpha}{2}, \frac{\alpha}{2}], v \in [-\frac{2\pi}{\alpha}, \frac{2\pi}{\alpha}], u' \in [-\frac{\beta}{2}, \frac{\beta}{2}], and v' \in [-\frac{2\pi}{\beta}, \frac{2\pi}{\beta}]$ and denote by $|r_0\rangle_r$ an eigenstate of the operator **r** with eigenvalue r_0 .

We will later use the following properties.

$$|u, v\rangle_{a} = e^{-\frac{i}{2}uv} \sum_{n \in \mathbb{Z}} e^{-inu\frac{2\pi}{\alpha}} \left| n\frac{2\pi}{\alpha} + v \right\rangle_{p_{a}},$$
$$|u', v'\rangle_{b} = e^{-\frac{i}{2}u'v'} \sum_{m \in \mathbb{Z}} e^{-imu'\frac{2\pi}{\beta}} \left| m\frac{2\pi}{\beta} + v' \right\rangle_{p_{b}}.$$
(B2)

Displacements (for Zak states of either oscillator):

$$e^{-iw\mathbf{p}}|u, v\rangle = e^{-\frac{i}{2}wv}|u+w, v\rangle,$$

$$e^{+iw\mathbf{q}}|u, v\rangle = e^{\frac{i}{2}wu}|u, v+w\rangle,$$
(B3)

where u + w and v + w are to be considered as modular coordinates (respectively modulo α or β and modulo $\frac{2\pi}{\alpha}$ or $\frac{2\pi}{\beta}$).

We will now show that, if the target mode is initialized in the $|\pm Z\rangle$ logical basis, the states of both the auxiliary and target oscillators are described by diagonal density matrices in their respective Zak bases throughout auxiliary state preparation and Steane-type error correction. Therefore, they can be represented by wrapped probability distributions Q_a and P_a for the target oscillator, Q_b and P_b for the auxiliary oscillator. Moreover, these distributions are separable between the two parameters of each Zak basis:

$$\boldsymbol{\rho}_{a} = \int_{u} \int_{v} Q_{a}(u) P_{a}(v) |u, v\rangle \langle u, v|_{a},$$
$$\boldsymbol{\rho}_{b} = \int_{u'} \int_{v'} Q_{b}(u') P_{b}(v') |u', v'\rangle \langle u', v'|_{b}.$$
(B4)

We also give evolution rules for these distributions throughout correction rounds and cycles, on which numerical simulations used in this work are based.

2. Auxiliary state preparation

We here describe the evolution of the auxiliary state through a \mathcal{R}_{q_b} preparation round. The results can be directly adapted to the case of \mathcal{R}_{p_b} rounds. Moreover, we drop the subscript *b* to simplify notations.

A \mathcal{R}_q round labeled j ($N_p + 1 \leq j \leq N_p + N_q$) starts with a qubit initialization in the +1 eigenstate of its Pauli operator σ_x , followed by a conditional displacement gate $\mathbf{U}_q^{\text{CD}} = e^{i\theta q\sigma_z}$ where $\theta = \frac{\pi}{\beta}$. The qubit is then measured along σ_y , and a feedback displacement by $\pm \epsilon_j$ is applied along q depending on the outcome. The Kraus operators corresponding to the two possible outcomes are [8]

$$\mathbf{M}_{+} = e^{-i\epsilon_{j}\mathbf{p}}\cos\left(\theta \mathbf{q} + \frac{\pi}{4}\right),$$
$$\mathbf{M}_{-} = e^{+i\epsilon_{j}\mathbf{p}}\cos\left(\theta \mathbf{q} - \frac{\pi}{4}\right).$$
(B5)

If, before the round, the auxiliary state is of the form (B4) with probability distributions

$$Q_{i-1}(u), P_{i-1}(v),$$
 (B6)

the conditional states after the qubit readout and feedback displacement are of the same form. In detail, if no qubit flip occurred during the gate, which happens with probability $1 - p^{\text{PF}} - p^{\text{BF}}$ (in the limit of small flip probability), the nonnormalized conditional probability distributions read as

$$Q_{j-1}^{\pm,\mathrm{NF}}(u) = (1 - p^{\mathrm{PF}} - p^{\mathrm{BF}}) \left[\frac{1}{2} \pm \frac{1}{2} \sin\left(\frac{2\pi}{\beta}(u \pm \epsilon_j)\right) \right]$$
$$\times Q_{j-1}(u \pm \epsilon_j),$$
$$P_{j-1}^{\pm,\mathrm{NF}}(v) = (1 - p^{\mathrm{PF}} - p^{\mathrm{BF}}) P_{j-1}(v).$$
(B7)

As detailed in Appendix A, phase flips of the qubit during the gate, occurring with probability p^{PF} , lead to an erroneously applied feedback displacement, yielding non-normalized conditional probability distributions

$$Q_{j-1}^{\pm, \text{PF}}(u) = p^{\text{PF}} \left[\frac{1}{2} \pm \frac{1}{2} \sin\left(\frac{2\pi}{\beta}(u \mp \epsilon_j)\right) \right] Q_j(u \mp \epsilon_j),$$

$$P_{j-1}^{\pm, \text{PF}}(v) = p^{\text{PF}} P_{j-1}(v)$$
(B8)

while bit flips of the qubit during the gate, occurring with probability p^{BF} , result in a randomly applied feedback displacement by $\pm \epsilon_j$ along q_b and a long displacement along p_b uniformly sampled in $\left[-\frac{\pi}{\beta}, \frac{\pi}{\beta}\right]$. The corresponding non-normalized conditional probability distributions read as

$$Q_{j-1}^{\rm BF}(u) = \frac{p^{\rm BF}}{2} (Q_{j-1}(u+\epsilon_j) + Q_{j-1}(u-\epsilon_j)),$$

$$P_{j-1}^{\rm PF}(v) = p^{\rm BF} \frac{\beta}{2\pi}.$$
 (B9)

After recombining all conditional probability distributions to model the proportional (memoryless) feedback strategy, the summed distributions read as

$$Q_{j-1}^{\text{FB}}(u) = \left[\frac{1}{2} + \frac{p^{\text{NF}}}{2}\sin\left(\frac{2\pi}{\beta}(u+\epsilon_{j})\right)\right]Q_{j-1}(u+\epsilon_{j}) \\ + \left[\frac{1}{2} - \frac{p^{\text{NF}}}{2}\sin\left(\frac{2\pi}{\beta}(u-\epsilon_{j})\right)\right]Q_{j-1}(u-\epsilon_{j}), \\ P_{j-1}^{\text{FB}}(v) = (1-p^{\text{BF}})P_{j-1}(v) + p^{\text{BF}}\frac{\beta}{2\pi}, \tag{B10}$$

where we defined $p^{\text{NF}} = 1 - 2p^{\text{PF}} - p^{\text{BF}}$. At the end of the round, we apply an effective quadrature noise channel, which convolves the probability distributions with wrapped normal distributions G_q and G_p , respectively, defined on $\left[-\frac{\beta}{2}, \frac{\beta}{2}\right]$ and $\left[-\frac{\pi}{\beta}, \frac{\pi}{\beta}\right]$, both with variance $\sigma^2 = \kappa T_{\text{round}}$. We thus get at the beginning of the following round a state of the form (B4) with probability distributions

$$Q_{j}(u) = Q_{j-1}^{\text{FB}} * G(u),$$

$$P_{j}(v) = P_{j-1}^{\text{FB}} * G(v).$$
(B11)

The evolution of the auxiliary state through a \mathcal{R}_p round is simply obtained by the exchange $q \rightarrow p$ in the above formulas. In our simulations, we initialize the auxiliary state of the form (B4) with uniform Q_0 and P_0 distributions before preparation (see Appendix B 4 for a justification of this hypothesis). It follows from the above analysis that the auxiliary state remains of this form throughout preparation. Note that the Kraus map defined by the operators (B5), as well as quadrature noise, suppress both off-diagonal terms of the Zak basis density matrix and classical correlations between the values of the modular operators $\tilde{\mathbf{q}}$ and $\tilde{\mathbf{p}}$. Therefore, a state which is not initially of the form (B4) becomes so after a long sequence of preparation rounds.

3. Target oscillator error correction

We here describe the evolution of the target oscillator state through a C_{q_a} correction cycle [see Fig. 2(c)]. A cycle starts with the auxiliary state prepared as described in the previous section. A quadrature gate $\mathbf{U}_{q_a}^{\text{quad}} = e^{i\theta \mathbf{q}_a \mathbf{q}_b}$ with $\theta = \frac{\alpha}{\beta}$ is applied to the oscillators, followed by detection of the $\mathbf{\tilde{p}}_b$ stabilizer with outcome $m \in [-\frac{\pi}{\beta}, \frac{\pi}{\beta}]$. In this section, we assume this detection to be perfect, and detail how to model its finite accuracy in Appendix B 4. Finally, the target oscillator is displaced by -f(m) along q_a , and an effective noise channel is applied to the target oscillator state to account for quadrature noise throughout the cycle.

We suppose the target state to be of the form (B4) when the *j*th cycle begins. After the auxiliary state preparation, which yields a state of the form (B4) with probability distributions Q_{b,N_p+N_q} and P_{b,N_p+N_q} , abbreviated to Q_b and P_b for simplicity, the joint state of the system reads as

$$\rho_{j-1}^{0} = \int_{u} \int_{v} \int_{u'} \int_{v'} Q_{a_{j-1}}(u) P_{a_{j-1}}(v) Q_{b}(u')$$

$$\times P_{b}(v') |u, v\rangle \langle u, v|_{a} |u', v'\rangle$$

$$\times \langle u', v'|_{b} du dv du' dv'.$$
(B12)

After the quadrature gate, the state reads as

$$\boldsymbol{\rho}_{j-1}^{1} = \int_{u} \int_{v} \int_{u'} \int_{v'} Q_{a_{j-1}}(u) P_{a_{j-1}}(v) Q_{b}(u')$$

$$\times P_{b}(v') | u, v + \theta u' \rangle \langle u, v + \theta u' |_{a} | u', v' + \theta u \rangle$$

$$\times \langle u', v' + \theta u |_{b} du dv du' dv'. \tag{B13}$$

Detection of the $\tilde{\mathbf{p}}_b$ stabilizer yielding an outcome *m* is modeled by the application of the Kraus operator $\mathbf{M}_m = \int_{u'} |u', m\rangle \langle u', m|_b$. After tracing out the auxiliary state, the non-normalized target oscillator density matrix conditioned on the outcome *m* reads as

$$\rho_{a_{j-1}}^{m} = \int_{u} \int_{v} \int_{u'} Q_{a_{j-1}}(u) P_{a_{j-1}}(v) Q_{b}(u') P_{b}(m-\theta u)$$
$$\times |u, v + \theta u'\rangle \langle u, v + \theta u'|_{a} du dv dv'.$$
(B14)

After a feedback displacement by -f(m) and summing over m to model our memoryless feedback strategy (in the sense that the measurement records are not carried to the following cycle), we get

$$\boldsymbol{\rho}_{a_{j-1}}^{\text{FB}} = \int_{u} \int_{v} \int_{u'} \int_{m} Q_{a_{j-1}}(u) P_{a_{j-1}}(v) Q_{b}(u')$$

$$\times P_{b}(m - \theta u) |u - f(m), v + \theta u'\rangle$$

$$\times \langle u - f(m), v + \theta u'|_{a} \, du \, dv \, dv' dm. \quad (B15)$$

Given that the probability distributions are wrapped functions and that the integrals are defined over their whole domains, we find that this state is of the form (B4) with probability distributions

$$Q_{a_{j-1}}^{\text{FB}}(u) = \int_{m} Q_{a_{j-1}}[u+f(m)]P_b(m-\theta[u+f(m)])dm,$$
$$P_{a_{j-1}}^{\text{FB}}(v) = \int_{u'} P_{a_{j-1}}(v-\theta u')Q_b(u')du'.$$
(B16)

Finally, we apply the effective noise channel accounting for quadrature noise during the gate and the following auxiliary state preparation rounds. It convolves the probability distribution Q_a with a wrapped normal distribution G_a of variance $\sigma^2 = \kappa [T_{quad} + (N_q + N_p)T_{round}]$, and the probability distribution P_a with a wrapped normal distribution \tilde{G}_a with slightly larger variance to account for the renormalized quadrature noise $\kappa \rightarrow \tilde{\kappa}$ during the quadrature gate (Appendix A 3 b). We thus get the target state at the beginning of the following cycle, also of the form (B4), with probability distributions

$$Q_{a_{j}}(u) = Q_{a_{j-1}}^{\text{FB}}(u) * G_{a}(u),$$

$$P_{a_{j}}(v) = P_{a_{j-1}}^{\text{FB}}(v) * \tilde{G}_{a}(v).$$
(B17)

The evolution of the target oscillator state during a C_{p_a} correction cycle is derived with similar calculations, inverting the role of Q_{a_j} and P_{a_j} . It also transforms a state of the form (B4) into a state of the same form. Therefore, if the target is initialized in a state of this form, e.g., when prepared in $|+Z\rangle$, it remains so indefinitely. In order to extract the decay rate of the *z* component of the GKP qubit Bloch vector under a particular set of error-correction parameters, one only needs to compute the evolution of Q_a through successive C_{q_a} and C_{p_a} cycles. After some number of cycles N_c , the GKP qubit is decoded and its *z* Bloch sphere coordinates read as

$$z(N_c) = \int_u Q_{a_{N_c}}(u) \Theta(u) du, \qquad (B18)$$

where Θ is a step function with value 1 on $\left[-\frac{\alpha}{4}, \frac{\alpha}{4}\right]$ and -1 elsewhere. By fitting the decay of $z(N_c)$ with an exponential function, one extracts the decay rate of the *z* component of the GKP qubit Bloch vector κ_{\log} . In Appendix C we present a more efficient method to extract this same rate.

Note that with the Zak basis we chose, constructed from the logical $|+Z\rangle$ state, we cannot directly simulate the decay of other logical Pauli operators. One could do so by considering alternative Zak basis definitions. However, the square GKP code symmetry properties ensure that the three components of the logical Bloch vector decay with respective rates $\kappa_z = \kappa_x = \kappa_y/2 = \kappa_{log}$.

4. Detection of the modular stabilizer

In the previous section, we considered the detection of the $\tilde{\mathbf{p}}_b$ stabilizer as perfect and instantaneous. Since this measurement can be destructive for the auxiliary oscillator state, homodyne detection of p_b is typically considered in the literature. However, letting the field leak out of the auxiliary resonator to be detected requires to wait at least a timescale of a few $1/\kappa$. This is not a viable option for error correction, which requires $\kappa T_{cycle} \ll 1$. One could partly circumvent the issue by mapping the value of p_b to a supplementary, low-Q resonator via a quadrature gate, but we found that, for a quadrature-quadrature interaction strength of the same order

as that activated between the target and auxiliary oscillators, the operation would here again dominate the error-correction cycle duration. Moreover, combined photon collection and homodyne detection efficiencies are in practice limited to $\eta \lesssim \frac{1}{2}$ in all experimental platforms, which would result in a too low detection accuracy. Alternatively, we consider detecting the modular operator $\tilde{\mathbf{p}}_b$ through repeated physical qubit-based measurement rounds. Indeed, the outcome of the \mathcal{R}_{p_b} rounds preparing the auxiliary state for the following cycle can be straightforwardly decoded to estimate the value of $\tilde{\mathbf{p}}_b$ prior to repreparation, with sufficient accuracy for error correction. This method belongs to the class of phase-estimation protocols [12–18]. Indeed, measuring the value of the modular operator $\tilde{\mathbf{p}}_b$ is equivalent to estimating the phase of $e^{i\beta\mathbf{p}_b}$. Note, however, that here the measurement is not quantum nondemolition (QND) in the sense that the phase of $e^{i\beta \mathbf{p}_b}$ is modified during each \mathcal{R}_{p_b} round by the applied feedback displacements. As detailed below, the memory of the initial phase is fully erased after a few tens of rounds.

To justify this approach and estimate the \tilde{p}_b detection accuracy, we suppose that the auxiliary oscillator is in a Zak-diagonal state of the form (B4) with a P_b probability distribution Dirac peaked in p_0 , whose value we want to estimate. Over a number N_p of \mathcal{R}_{p_b} preparation rounds, this distribution is on average shifted and broadened by the feedback displacements $\{\pm \epsilon_j\}_{1 \leq j \leq N_p}$ applied at the end of each round. Denoting $S = \{s_j\}_{1 \leq j \leq N_p}$ a particular measurement record (with $s_j = \pm 1$ for each round j) and $-m(S) = \sum_{j=1}^{N_p} s_j \epsilon_j$ the total applied displacement, the auxiliary state p distribution after re-preparation reads as

$$P_{p_0}^{\mathcal{S}}(p) = \delta_{\frac{2\pi}{\beta}}[p - p_0 + m(\mathcal{S})],$$
(B19)

where $\delta_{\frac{2\pi}{\beta}}$ is a Dirac comb of period $\frac{2\pi}{\beta}$. Averaging over all possible measurement outcomes, the auxiliary state distribution after repreparation reads as

$$P_{p_0}(p) = \sum_{\mathcal{S}} \mathcal{P}_{p_0}(\mathcal{S}) \delta_{\frac{2\pi}{\beta}}[p - p_0 + m(\mathcal{S})], \qquad (B20)$$

where $\mathcal{P}_{p_0}(S)$ is the probability of the measurement record S. P_{p_0} becomes smooth for N_p sufficiently large.

We simply propose to estimate p_0 with m(S) for a given measurement record S. The accuracy of the $\tilde{\mathbf{p}}_b$ detection so performed is characterized by the distribution \mathcal{E}_{p_0} of the error $e(S) = m(S) - p_0$. It reads as

$$\mathcal{E}_{p_0}(e) = \sum_{\mathcal{S}} \mathcal{P}_{p_0}(\mathcal{S}) \delta_{\frac{2\pi}{\beta}}[e - m(\mathcal{S}) - p_0] = P_{p_0}(-e).$$
(B21)

The last equality simply means that the detection accuracy is as good as the auxiliary state repreparation.

Crucially, we observe in Fig. 5 that for all the repreparation sequences used in this work, the auxiliary state distribution P_{p_0} after the \mathcal{R}_{p_b} rounds, and thus the error distribution \mathcal{E}_{p_0} , depends negligibly on p_0 . This is simply understood as the long feedback kicks ϵ_j applied during the first few \mathcal{R}_{p_b} rounds quickly erase the memory of its prior state. This justifies *a posteriori* the hypothesis made in Appendix B 2 that the auxiliary state is of the form (B4) with uniform distributions prior to



FIG. 5. Measurement and repreparation of the auxiliary state. We compute the P_{p_0} distribution of the auxiliary oscillator (plain lines) prepared by a number $N_p = 20$ of \mathcal{R}_{p_b} rounds from a narrow Gaussian distribution (dashed lines, standard deviation $\sigma \sim 0.1$) centered at p_0 (encoded in color). The feedback displacements applied after each round and the rectangularity parameter of the auxiliary oscillator are the ones returned by gradient ascent to minimize the logical error rate for $p^{\text{BF}} = 2p^{\text{PF}} = 0.005$ and $\kappa T_{\text{round}} = 2.10^{-5}$. We pick these example parameters, in particular the small number of preparation rounds, as the *a priori* less favorable situation for the prepared auxiliary state not to depend on the initial condition p_0 . We observe that the final distributions corresponding to different initial states do not differ significantly, justifying our approach to modular operator detection as detailed in Appendix B 4.

repreparation: any initial state would yield the same prepared state. As for the finite accuracy of the $\tilde{\mathbf{p}}_b$ detection this method yields, it can be modeled by an ideal detection preceded by a convolution of the P_b probability distribution with the error function $\mathcal{E} = P_{N_p}$, where P_{N_p} is the distribution describing the auxiliary state prepared, from an arbitrary state, by a number N_p of \mathcal{R}_{p_b} rounds as detailed in Appendix B 2 (we used that P_{N_p} is an even distribution). Note that in Fig. 5, the auxiliary state distribution prior to repreparation is a Gaussian centered in p_0 and with width $\sigma = 0.1$ in rescaled coordinates. Taking $\sigma \rightarrow 0$ as in the above calculation led to numerical aberrations attributed to the encoding of the wrapped distributions in the form of Fourier coefficient vectors of length $2n_F + 1$ with $n_F = 60$ (see Appendix C 1).

In our reasoning, we have omitted shifts of the auxiliary state distribution entailed by flips of the qubit and intrinsic quadrature noise of the auxiliary oscillator during the \mathcal{R}_{p_b} rounds. The former only entails shifts of the Q_b probability distribution during \mathcal{R}_{p_b} rounds, and has no impact on P_b . The effect of the latter is to broaden the P_b distribution as it is being measured and reprepared. We model it by including quadrature noise in the numerical computation of P_{N_p} , by which we expect to slightly underestimate the $\tilde{\mathbf{p}}_b$ detection accuracy. Indeed, by supposing that P_{N_p} is solely broadened by the stochastic nature of the applied feedback displacements we overestimate the spread of \mathcal{E} .

APPENDIX C: EFFICIENT NUMERICAL ESTIMATE OF THE GKP QUBIT DECOHERENCE RATE

Computing the evolution of the auxiliary and target state under the form of classical probability distributions Q_a , P_a , Q_b , and P_b as detailed in the previous section greatly reduces the cost of numerical simulation compared to a full description in terms of density matrices. Typically, one keeps track of the distributions as two vectors of length 1000. In this section, we further reduce simulation costs by representing the probability distributions in Fourier domain, as vectors of $2n_F + 1$ Fourier coefficients with $30 \le n_F \le 60$. After translating the calculations of Appendixes B 2 and B 3 to the Fourier domain in Appendixes C1 and C2, we encode the evolution of the target oscillator over a pair of C_{q_a}/C_{p_a} cycle in an evolution matrix and show how to extract the decay rate of the GKP qubit by spectral analysis of this matrix in Appendix C3. We also show how to compute the gradient of this rate with respect to the continuous parameters of the protocol (length of the feedback displacements ϵ_i applied after each round, Fourier coefficients of the target oscillator feedback law f, rectangularity $R = \sqrt{2\pi/\beta}$ of the auxiliary GKP state lattice), which greatly facilitates their optimization.

In order to simplify calculations, we consider in the following rescaled wrapped distributions Π_{q_a} , Π_{q_b} , and Π_{p_b} , and rescaled feedback shifts e_j at the end of each round, $\tilde{\mathbf{p}}_b$ detection outcome ψ at the end of each cycle, and feedback law *F* governing the feedback displacement applied to the target oscillator, all defined over $[-\pi, \pi]$:

$$\Pi_{q_a}(\phi) = \frac{1}{\sqrt{\pi}} Q_a \left(\frac{\phi}{\sqrt{\pi}}\right),$$

$$\Pi_{q_b}(\phi) = \frac{\beta}{2\pi} Q_b \left(\frac{\beta\phi}{2\pi}\right),$$

$$\Pi_{p_b}(\phi) = \frac{1}{\beta} P_b \left(\frac{\phi}{\beta}\right),$$

$$e_j = \epsilon_j \beta \quad \text{for } 1 \leq j \leq N_p,$$

$$e_j = \epsilon_j \frac{2\pi}{\beta} \quad \text{for } N_p + 1 \leq j \leq N_p + N_q,$$

$$\psi = m\beta,$$

$$F(\psi) = \sqrt{\pi} f\left(\frac{\psi}{\beta}\right).$$
(C1)

We also define the Fourier coefficients of a 2π -periodic function g as $g^{(k)} = \frac{1}{2\pi} \int_{-\pi}^{\pi} g(\phi) e^{-ik\phi} d\phi$. Note that the evolution of the P_a probability distribution is not considered here as the decay rate κ_{\log} of the GKP qubit. The *z*-Bloch vector component is computed from the evolution of Q_a only.

1. Auxiliary state preparation in Fourier domain

We revisit the auxiliary state preparation described in Appendix B 2 to translate it in the Fourier domain. The distributions are uniform before preparation, with Fourier coefficients $\Pi_{p_{b},0}^{(k)} = \Pi_{q_{b},0}^{(k)} = \delta_{k}/(2\pi)$, where δ is the Kronecker symbol.

During the *j*th \mathcal{R}_{p_b} round, the Π_{p_b} distribution evolves after the physical qubit readout and application of a feedback displacement following Eq. (B10), which reads as in rescaled

coordinates

$$\Pi_{p_{b},j-1}^{\text{FB}}(\phi) = \left[\frac{1}{2} + \frac{p^{\text{NF}}}{2}\sin(\Phi + e_{j})\right]\Pi_{p_{b},j-1}(\phi + e_{j}) + \left[\frac{1}{2} - \frac{p^{\text{NF}}}{2}\sin(\phi - e_{j})\right]\Pi_{p_{b},j-1}(\phi - e_{j}),$$
(C2)

where we used the shorthand notation $p^{\text{NF}} = 1 - p^{\text{BF}} - 2p^{\text{PF}}$. Expanding this expression in powers of e_j , we get

$$\Pi_{p_{b},j-1}^{\text{FB}}(\phi) \simeq \frac{1}{2} \sum_{n=0}^{n_{T}} \frac{e_{j}^{n}}{n!} \left(\frac{\partial^{n} \Pi_{p_{b},j-1}(\phi)}{\partial \phi^{n}} [1 + (-1)^{n}] + p^{NF} \frac{\partial^{n} [\Pi_{p_{b},j-1}(\phi) \sin(\phi)]}{\partial \phi^{n}} (1 - (-1)^{n}) \right).$$
(C3)

Note that the term n = 1 corresponds to a drift velocity $e_j p^{\text{NF}} \sin(\phi)/T_{\text{round}}$ and the term n = 2 to a diffusion constant $e_j^2/(2T_{\text{round}})$, quoted in Sec. II in nonrescaled coordinates. Neglecting following terms, one obtains a Fokker-Planck equation, which is only valid for $e_j \rightarrow 0$. For the numerical simulations performed in this work, we truncate the expansion at $n_T = 30$.

In Fourier domain, this expression translates to

$$\Pi_{p_{b},j-1}^{\text{FB}(k)} = \sum_{n \text{ even}} \frac{(ike_{j})^{n}}{n!} \Pi_{p_{b},j-1}^{(k)} + \sum_{n \text{ odd}} p^{\text{NF}} \frac{(ike_{j})^{n}}{n!} \frac{1}{2i} (\Pi_{p_{b},j-1}^{(k-1)} - \Pi_{p_{b},j-1}^{(k+1)}). \quad (C4)$$

The distribution is then convolved with a Gaussian kernel modeling the effect of quadrature noise [see Eq. (B11)]. In Fourier domain, it reads as

$$\Pi_{p_{b},j}^{(k)} = \Pi_{p_{b},j-1}^{\text{FB}\,(k)} e^{-\frac{1}{2}\kappa_{p}T_{\text{round}}k^{2}},\tag{C5}$$

where $\kappa_p = 2\pi\kappa/R^2$ is the rescaled quadrature noise rate. After N_p rounds, the error function \mathcal{E} for the $\tilde{\mathbf{p}}_b$ detection is inferred from the distribution Π_{p_b,N_p} (see Appendix B 2). The Π_{q_b,N_p} distribution is still uniform at this stage as we assume the distributions $\Pi_{q_b,0}$ and $\Pi_{p_b,0}$ prior to preparation to be uniform (see Appendix B 4).

Through the sequence of \mathcal{R}_{q_b} rounds, the Π_{p_b} distribution evolves due to quadrature noise and random displacements induced by bit flips of the qubit as

$$\Pi_{p_b,N_p+N_q}^{(k)} = \left(1 - p_{\text{tot}}^{\text{BF}}\right) \Pi_{p_b,N_p}^{(k)} e^{-\pi N_q \kappa_p k^2} + p_{\text{tot}}^{\text{BF}} \frac{\delta_k}{2\pi}, \quad (C6)$$

where $p_{\text{tot}}^{\text{BF}} = 1 - (1 - p^{\text{BF}})^{N_q}$ is the probability for at least one bit flip to have occurred. As for the Π_{q_b} distribution, it evolves through \mathcal{R}_{q_b} rounds following the same rules as Π_{p_b} through \mathcal{R}_{p_b} rounds [Eqs. (C4) and (C5)], albeit with a rescaled quadrature noise rate $\kappa_q = 2\pi\kappa R^2$ for the Gaussian kernel convolution.

Overall, we thus compute the prepared auxiliary state under the form of two $(2n_F + 1)$ vectors of Fourier coefficients $(-n_F \leq k \leq n_F)$, and obtain the error function \mathcal{E} for the $\tilde{\mathbf{p}}_b$ detection under the same form. Moreover, it is straightforward

to compute the gradient of each vector with respect to each feedback displacement length e_i , as well as with respect to the grid rectangularity parameter R, by taking the derivative of the formulas given above and applying chain rules.

2. Target oscillator dynamics in Fourier domain

We revisit the target oscillator evolution over a pair of C_{q_a}/C_{p_a} cycles, labeled j and j + 1, described in Appendix B 3 to translate it in Fourier domain. The auxiliary state distributions before the quadrature gate are Π_{q_b,N_p+N_q} (abbreviated to Π_{q_b}) and Π_{p_b,N_p+N_q} , as computed in the previous section. As detailed in Appendix B 4, we model the inaccuracy of the $\tilde{\mathbf{p}}_b$ detection by convolving Π_{p_b,N_p+N_q} with an error distribution $\mathcal{E} = \prod_{p_b, N_b}$, which is a simple vector multiplication in Fourier domain, and denote the resulting distribution by Π_{p_b} .

During the C_{q_a} cycle, the initial target oscillator distribution $\Pi_{q_a, j-1}$ is first evolved with the left expression in (B16) modeling the quadrature gate followed by a measurement of $\tilde{\mathbf{p}}_b$ whose outcome controls a feedback displacement applied to the target oscillator. In rescaled coordinates, this evolution reads as

$$\Pi_{q_{a},j-1}^{\text{FB}}(\phi) = \int_{-\pi}^{\pi} \Pi_{q_{a},j-1}[\phi + F(\psi)] \\ \times \Pi_{p_{b}}\{\psi - 2[\phi + F(\psi)]\}d\psi.$$
(C7)

We now expand this expression in powers of the rescaled feedback displacement $F(\psi)$ applied to the target oscillator, and truncate the series at n_T ($n_T = 30$ for all simulations performed in this work). We then get

$$\Pi_{q_{a},j-1}^{\text{FB}}(\phi) \simeq \int_{-\pi}^{\pi} \sum_{n=0}^{n_{T}} \frac{F^{n}(\psi)}{n!} \frac{\partial^{n}}{\partial \phi^{n}} \\ \times \left(\Pi_{q_{a},j-1}(\phi)\Pi_{p_{b}}(\psi-2\phi)\right) d\psi \\ = \sum_{n=0}^{n_{T}} \frac{1}{n!} \frac{\partial^{n}}{\partial \phi^{n}} \left(D_{n}(\phi)\Pi_{q_{a},j-1}(\phi)\right), \quad (C8)$$

where we defined the generalized Fokker-Planck coefficient functions D_n :

$$D_n(\phi) = \int_{-\pi}^{\pi} F^n(\psi) \Pi_{p_b}(\psi - 2\phi)$$
$$= (F^n * \Pi_{p_b})(2\phi)$$
(C9)

(we use that Π_{p_b} is even in the last equality). In Fourier domain, this translates to

$$\Pi_{q_a,j-1}^{\text{FB}(k)} = \sum_{n=0}^{n_T} \frac{(ik)^n}{n!} \left(\sum_{l=-N}^N D_n^{(k-l)} \Pi_{q_a,j-1}^{(l)} \right)$$
(C10)

and the Fourier coefficients of D_n are computed with

$$D_n^{(k)} = \begin{cases} (F^n * \Pi_{p_b})^{(\frac{k}{2})} = 2\pi (F^{\frac{3}{4}^n})^{(\frac{k}{2})} \Pi_{p_b}^{(\frac{k}{2})} & \text{if } k \text{ even,} \\ 0 & \text{if } k \text{ odd,} \end{cases}$$
(C11)

where $\tilde{*}^n$ denotes the *n*-fold discrete convolution product defined as $(u \tilde{*} v)^{(k)} = \sum_{l=-N}^{N} u^{(k-l)} v^{(l)}$. In simulations, we truncate this sum in order to maintain a $2n_F + 1$ structure for the Fourier coefficient vectors.

The distribution is then convolved with a Gaussian kernel G_a modeling the effect of quadrature noise during the C_{q_a} cycle [left equation in (B17)], then convolved with the Π_{q_b} distribution to model the backaction of the quadrature gate in the following C_{p_a} cycle [right equation in (B16), replacing $P_a \rightarrow Q_a$], and again convolved with a Gaussian kernel \tilde{G}_a modeling the effect of quadrature noise during the C_{p_a} cycle [right equation in (B17) replacing $P_a \rightarrow Q_a$]. In Fourier domain, it reads as

$$\Pi_{q_a,j+1}^{(k)} = 2\pi \,\Pi_{q_b}^{(k)} e^{-\frac{k^2 \sigma_{\text{tot}}^2}{2}} \Pi_{q_a,j-1}^{\text{FB}\,(k)} \tag{C12}$$

with $\sigma_{tot}^2 = 2\kappa T_{cycle} + \frac{\theta^2}{3}\kappa T_{quad}$. Combining Eqs. (C10) and (C12), the evolution through the two cycles can be expressed under a matrix form

$$\Pi_{q_a,j+1}^{(k)} = \sum_{l=-N}^{N} M_{kl} \Pi_{q_a,j-1}^{(l)}.$$
 (C13)

Note that M is real when F is odd, which is the case in the following.

3. GKP qubit decoherence rate and convergence rate to the code manifold by spectral analysis of the evolution matrix

The evolution matrix M is the Fourier transform of a stochastic matrix. As such, it shares the same eigenspectrum $\{\lambda_i\}$ where we arrange the eigenvalues in decreasing magnitude order. In particular $\lambda_0 = 1$, and $|\lambda_i| \leq 1$ for $i \geq 1$.

In the regime where the logical flip probability per cycle is small, we find that the spectrum is gapped with $|\lambda_j| \ll |\lambda_1|$ for j > 1. Qualitatively, this gap indicates a fast convergence of the system to a 2D manifold of probability vectors (distributions), at a rate

$$\Gamma_{\rm conv} = -\ln(|\lambda_2|)/(2T_{\rm cycle}) \tag{C14}$$

(note that the evolution matrix corresponds to two errorcorrection cycles). We interpret this fast dynamics as a convergence of the target oscillator state to a metastable state in the vicinity the GKP code manifold. It is followed by a slow relaxation, within this manifold, to the system steady state (the probability distribution Π_0 obtained by inverse Fourier transform of the eigenvector attached to λ_0) at a rate

$$\kappa_{\log} = -\ln(\lambda_1)/(2T_{\text{cycle}}). \tag{C15}$$

In this expression, we have used that, since M is real and λ_1 does not have a conjugate eigenvalue, λ_1 is real. We interpret this slow dynamics as the relaxation of the GKP qubit towards the mixed logical state.

We confirm this intuition by representing the probability distributions Π_0 and Π_1 corresponding to λ_0 and λ_1 in Fig. 6, for cycle parameters allowing a robust protection of the GKP qubit. Π_0 displays two peaks of equal height centered in $\Phi = 0$ and $\phi = \pi$, as expected from a state close to the code manifold and decoded as the fully mixed logical state. With the proper normalization, $\Pi_0 + \Pi_1$ displays a single peak centered in $\phi = 0$, as expected from a state close to the code manifold and decoded as $|+Z\rangle$.



FIG. 6. Eigenvectors of the evolution matrix M. For $\kappa T_{\text{round}} =$ 2×10^{-5} , $p^{\text{BF}} = 1 \times 10^{-5}$, $N_p = N_q = 60$, and all other parameters optimized by gradient ascent, we represent the inverse Fourier transform of the eigenvectors of M with largest eigenvalues $\lambda_0 = 1$ and $\lambda_1 = 1 - 1.3 \times 10^{-5}$, respectively labeled Π_0 and Π_1 . Rescaled to a unit L_1 norm, Π_0 is the probability distribution of the target oscillator steady state under error correction ($\Pi_{q_a,j}$ with $j \to \infty$ in Appendix C 2). This state is close to the code manifold, with narrow peaks centered at $\phi = 0 \mod \pi$ and is decoded as the fully mixed state of the GKP qubit. Π_1 has a null L_1 norm, and is here rescaled to the same L_{∞} norm as Π_0 . Given that λ_1 is close to 1 and that a gap exists with the next largest eigenvalue ($\lambda_2 = 0.55$), a general state converges in a few correction cycles to a probability distribution $\Pi_0 + \zeta \Pi_1$, where ζ is an excellent approximation of the z component of the GKP qubit Bloch vector when the peaks of Π_0 and Π_1 are sufficiently narrow.

This spectral analysis in Fourier domain is a powerful tool to estimate the decay rate of the z component of the GKP qubit Bloch vector. We compared its results to brute-force computation of the evolution of the target oscillator state, encoded as a probability vector, over a large number of errorcorrection cycles (see Appendix B 3) before fitting the decay of the decoded z component of the GKP qubit Bloch vector. Both methods agree quantitatively when the oscillators state are encoded in a sufficiently long Fourier vector of length $2n_F + 1$, and when the Taylor expansion in Eqs. (C3) and (C8) is truncated at a sufficiently high order n_T (not shown). In practice, we found that $n_F = n_T = 30$ was sufficient for all numerical simulations presented in this paper, except to estimate the smallest decay rates of Fig. 8 and to obtain the real-domain distributions with no visible ripples presented in Fig. 3, for which $n_F = 60$ was used. Given the small matrix size involved, spectral analysis in Fourier domain is significantly faster than brute-force simulation in real domain. It also allows us to estimate the convergence rate to the code manifold Γ_{conv} , as represented in Fig. 8.

Furthermore, for a given feedback parameter set, the method allows us to compute the gradient of λ_1 with respect to the cycle continuous parameters (length e_j of the feedback displacements on the auxiliary oscillator, Fourier coefficients $F^{(k)}$ of the feedback function, and rectangularity R of the auxiliary state). To this end, we first take the derivative of the evolution rules for the target and the auxiliary state probability distributions (see Appendixes C 1 and C 2), respectively,

through a cycle and through a round, and apply chain rules to obtain the derivative of the evolution matrix M with respect to a given parameter x. Each component of the gradient is then given by

$$\frac{\partial \lambda_1}{\partial x} = \frac{P_1^L \cdot M \cdot P_1^R}{P_1^L \cdot P_1^R} \tag{C16}$$

where the center dot denotes the matrix product and P_1^L and P_1^R are, respectively, left and right eigenvectors of M for the eigenvalue λ_1 .

APPENDIX D: OPTIMIZATION OF ERROR-CORRECTION PARAMETERS

1. Optimizing continuous parameters by gradient ascent

For a given set of noise values κT_{round} , p^{BF} , and p^{PF} ($p^{\text{PF}} = p^{\text{BF}}/2$ throughout this work) and given preparation round numbers N_q and N_p , we optimize the remaining parameters of the error-correction cycle by gradient ascent to maximize the value of λ_1 . In detail, we consider the gradients

$$A = \left\{ \frac{\partial \lambda_1}{\partial e_j} \right\}_{\{1 \le j \le N_p\}},$$

$$B = \left\{ \frac{\partial \lambda_1}{\partial e_j} \right\}_{\{N_p + 1 \le j \le N_p + N_q\}},$$

$$C = \left\{ \frac{\partial \lambda_1}{\partial F_s^{(k)}} \right\}_{\{1 \le k \le k_{\max}\}},$$

$$D = \left\{ \frac{\partial \lambda_1}{\partial R} \right\},$$

(D1)

where we defined $F_s^{(k)} = (F^{(k)} - F^{(-k)})/(2i)$. This choice constrains the feedback function *F* to the odd sector, ensuring that the target probability distribution remains symmetric at all time (real evolution matrix *M*). We choose to limit the number of free Fourier coefficients of *F* to $n'_F = 10 < n_F$ to limit aberrations entailed by Fourier series truncation during the convolution step (C11). Pushing n'_F to larger values, and increasing n_F accordingly to avoid aberrations, did not lead to a significant improvement in error-correction performances.

At each step l of the gradient ascent, for a total number of steps L = 100, we update the parameter values following

$$\{e_{j}\}_{\{1 \leq j \leq N_{p}\}}^{l+1} = \{e_{j}\}_{\{1 \leq j \leq N_{p}\}}^{l} + a \frac{A}{|A|_{\infty}} \Delta_{l},$$

$$\{e_{j}\}_{\{N_{p}+1 \leq j \leq N_{p}+N_{q}\}}^{l+1} = \{e_{j}\}_{\{N_{p}+1 \leq j \leq N_{p}+N_{q}\}}^{l} + b \frac{B}{|B|_{\infty}} \Delta_{l},$$

$$\{F_{s}^{(k)}\}_{\{1 \leq k \leq k_{\max}\}}^{l+1} = \{F_{s}^{(k)}\}_{\{1 \leq k \leq k_{\max}\}}^{l} + c \frac{C}{|C|_{\infty}} \Delta_{l},$$

$$\{R\}^{l+1} = \{R\}^{l} + d \frac{D}{|D|_{\infty}} \Delta_{l},$$

$$(D2)$$

where the step Δ_l decreases linearly from 1 to 0.05 when l varies from 1 to L, the parameters a = b = 0.005, c = 0.02, and d = 0.04 were adjusted empirically such that the vectors A, B, C, and D would converge at comparable speeds toward their final values.



FIG. 7. Optimization of feedback parameters by gradient ascent. The continuous control parameters of our protocol are optimized by gradient ascent to minimize the decay rate of the *z* component of the GKP qubit Bloch vector. The top two panels represent this decay rate κ_{log}^{l} as a function of the ascent step number *l*. On the left, we vary the physical qubit flip probability $p^{BF} = 2p^{PF}$ (encoded in color) for a fixed value of the oscillators' noise strength $\kappa T_{round} = 10^{-5}$ (left). On the right, we vary κT_{round} (encoded in color) for a fixed value of $p^{BF} = 2p^{PF} = 5 \times 10^{-4}$. In the three next lines of panels, we represent, for the same noise figures and at the end of the gradient ascent (*l* = 100), the feedback displacements applied after each \mathcal{R}_{p_b} round as a function of the round index $j \leq N_p$ (second line), the feedback displacements applied to the target oscillator as a function of $N_p < j \leq N_p + N_q$ (third line), and the feedback law *F* controlling the displacements applied to the target oscillator as a function of the normalized outcome ψ of the $\tilde{\mathbf{p}}_b$ estimation (fourth line). For each parameter, the initial guess (before gradient ascent, l = 0) is represented by a black dashed line. The total number of rounds is fixed to $N_q = N_p = 60$.

As initial guess parameters, we set

$$e_{j}^{0} = \frac{e_{i}e_{f}}{e_{f} + (e_{i} - e_{f})\frac{j}{N_{p}}} \quad \text{for } j \leq N_{p},$$

$$e_{j}^{0} = \frac{e_{i}e_{f}}{e_{f} + (e_{i} - e_{f})\frac{j - N_{p}}{N_{q}}} \quad \text{for } j > N_{p},$$

$$F_{s}^{(k)0} = f_{1}\delta_{k-1}, \quad R^{0} = 1.$$
(D3)

In these expressions, the initial guess for the feedback displacements on the auxiliary oscillator (first two expressions) is a truncated 1/j function with large initial value $e_i = \pi/2$ in order to suppress the tails of the Π_{p_b} and Π_{q_b} distributions, and small final value $e_f = 0.05 \ll 2\pi$ to limit the distributions' central peak width (see Sec. II). The 1/j power law was chosen to maximize the reduction rate of the distributions central peak width, while ensuring that this width reaches 0 when N_q , $N_p \to \infty$, in absence of intrinsic noise of the oscillator.



FIG. 8. Convergence rate towards the GKP code manifold (top panel) and decay rate of the *z* component of the GKP qubit Bloch vector (bottom panel) as a function of the number N_q of \mathcal{R}_{q_b} preparation rounds in each cycle, in absence of intrinsic oscillator noise ($\kappa = 0$) and for various physical qubit flip probabilities per round (encoded in color, with $p^{\text{PF}} = p^{\text{BF}}/2$). For each value of N_q , remaining control parameters are optimized by gradient ascent. The convergence rate towards the code manifold decreases with N_q as the probability of a single qubit flip to have occurred during the \mathcal{R}_{q_b} preparation rounds approaches 1, blurring error syndromes extracted during the following error-correction cycle. As a consequence, there exists a finite value of N_q that minimizes the GKP qubit decay rate.

The initial guess for the rescaled feedback law F is a simple sine function of amplitude $f_1 = 0.2$.

We observed that the final value of λ_1 (and hence the decay rate κ_{log}) and the correction parameters returned by the gradient ascent algorithm depends slightly on the initial guess, indicating the existence of multiple local minima of λ_1 (not shown). The rugged aspect of $\{e_j\}_{\{N_p+1 \le j \le N_p+N_q\}}^{l+1}$ observed after gradient ascent for some noise values (see Fig. 7, third line) tends to confirm this complex structure. More refined gradient ascent techniques may avoid these issues, but were not attempted in this work.

2. Optimizing the number of preparation rounds

In this section, we analyze the impact of the number of auxiliary state preparation rounds, N_q and N_p , on the performances of our error-correction protocol, and comment on the existence of an optimal, finite value for N_q and N_p , irrespective of the system noise strength.

We first consider the case of noiseless oscillators ($\kappa = 0$). Since the cycle duration is irrelevant in that limit, we allow the auxiliary state to be prepared with a large number $N_p \rightarrow \infty$ of \mathcal{R}_{p_b} rounds. Letting $e_j = e \rightarrow 0$, the evolution of the Π_{p_b} distribution over a \mathcal{R}_{p_b} round is given by Eq. (C3), which can be truncated at $n_T = 2$. We can approximate this discrete time evolution with a continuous time evolution governed by a Fokker-Planck equation

$$\frac{\partial \Pi_{p_b}}{\partial t} = -\frac{\partial \left(v(\phi) \Pi_{p_b} \right)}{\partial \phi} + \frac{1}{2} \frac{\partial^2 \left(D \Pi_{p_b} \right)}{\partial \phi^2}, \qquad (D4)$$

where $v(\phi) = -e(1 - p^{\text{BF}} - 2p^{\text{PF}})/T_{\text{round}}$ and $D = e^2/T_{\text{round}}$. After an infinite number of \mathcal{R}_{p_b} rounds, Π_{p_b} reaches the steady state of this equation, which approaches a wrapped normal distribution with variance $e/[2(1 - p^{\text{BF}} - 2p^{\text{PF}})]$ for $e \to 0$. In other words, Π_b is a Dirac distribution, and it follows from Apppendix B 4 that the detection of $\tilde{\mathbf{p}}_b$ following the quadrature gate is perfect.

We now consider the decay of the GKP qubit in this configuration. In Fig. 8, we represent the decay rate κ_{log} of the z component of the GKP qubit Bloch vector and the convergence rate towards the code manifold Γ_{conv} , in units of T_{cycle} , as a function of N_q . For each value of N_q , the remaining feedback parameters (feedback displacements following \mathcal{R}_{q_b} rounds, feedback law F, and rectangularity R) are optimized by gradient ascent as detailed in the previous section. Surprisingly, even in this limit case of noiseless oscillators, we find that an optimal number of \mathcal{R}_{q_b} rounds exists, which can be understood with the following arguments. When $N_q \rightarrow 0$, the Q_b distribution becomes widely spread (see Fig. 3, bottom panel) and long shifts propagate through the quadrature gate, increasing κ_{\log} . In the opposite limit $N_q \to \infty$, P_b is a near-uniform distribution as the probability of at least one bit flip during \mathcal{R}_{q_b} rounds approaches 1, blurring the error syndromes extracted from the target oscillator during the following correction cycle (see Fig. 3, bottom panel). As a result, the convergence rate to the code manifold drops to 0 (see Fig. 8, top panel), and small shifts propagating through the quadrature gate are sufficient to trigger logical errors. Admittedly, we still expect κ_{\log} to vanish for $N_q \rightarrow \infty$ (Dirac-peaked Q_b distribution) and F = 0(no feedback displacement applied to the target oscillator) as the target oscillator dynamics cancels, but this regime is reached for round numbers far beyond the range considered here.

Now considering the case of noisy oscillators ($\kappa > 0$), the optimal value of N_q is lower than in the noiseless case [see Figs. 4 and 9(a)]. Indeed, quadrature noise on the auxiliary oscillator during \mathcal{R}_{q_b} rounds causes the Π_{p_b} distribution to diffuse and homogenize, so that the convergence rate to the code manifold decreases faster with N_q than in the noiseless case. It also impacts the accuracy of the $\tilde{\mathbf{p}}_b$ detection, with a similar effect (see Appendix B4). In Fig. 9(a), we represent the optimal number of \mathcal{R}_{q_b} rounds N^{\min} found when sweeping together $N_q = N_p$, for the same range of system noise strength considered in Fig. 4. Note that here, N^{\min} is the value found to minimize $\kappa_{\log} T_{round}$ (and not $\kappa_{\log} T_{cycle}$ as in Fig. 8), which tends to favor smaller numbers of preparation round. For completeness, we represent in Figs. 9(b)-9(e) the value of the other cycle parameters, found by gradient ascent at $N_q = N_p = N^{\min}$, which yield the GKP qubit decay rates presented in Fig. 4. Finally, in order to confirm the exponential suppression of logical errors as a function of both the qubit error probability per round p^{BF} and the oscillator noise strength κT_{round} , we plot in Fig. 9(f) the same data as in Fig. 4 (decay rate of the z component of the GKP qubit Bloch vector κ_{\log}) but here as a function of p^{BF} , with κT_{round} encoded in color.



FIG. 9. Error-correction parameters used for the simulations whose results are presented in Fig. 4. For each considered noise figure (oscillator noise strength κT_{round} varied along the *x* axis of each panel, physical qubit flip probability per round $p^{BF} = 2p^{PF}$ encoded color) we sweep the number of preparation rounds $N_q = N_p$ and perform gradient ascent on the continuous correction parameters to minimize the logical qubit decay rate for each value of $N_q = N_p$. (a) Round number N^{min} yielding the minimum logical error rate after gradient ascent. For this round number, (b) represents the auxiliary lattice rectangularity R^{min} found by gradient ascent, (c) the dominant Fourier coefficient $F^{(1)min}$ of the feedback law controlling the feedback displacement applied to the target oscillator, (d) the average value of the feedback displacements $\overline{e_j}^{min}$ $(j \leq N_p)$ applied to the auxiliary oscillator during \mathcal{R}_{p_b} rounds, and (e) the average value of the feedback displacements $\overline{e_j}^{min}$ ($j > N_p$) applied to the auxiliary oscillator during \mathcal{R}_{q_b} rounds. (f) Decay rate κ_{log} of the *z* component of the GKP qubit Bloch vector as a function of p^{BF} , with κT_{round} encoded in color (different colorscale from other panels). These are the same data as presented in Fig. 4, albeit plotted against different axes.

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Correction: A change in wording has been made to the third sentence of the paragraph following Eq. (3).