Efficacy of noisy dynamical decoupling

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Dynamical decoupling (DD) refers to a well-established family of methods for error mitigation, comprising pulse sequences aimed at averaging away slowly evolving noise in quantum systems. Here we revisit the question of its efficacy in the presence of noisy pulses in scenarios important for quantum devices today: pulses with gate control errors, and the computational setting where DD is used to reduce noise in every computational gate. We focus on the well-known schemes of periodic (or universal) DD and its extension, concatenated DD, for scaling up its power. The qualitative conclusions from our analysis of these two schemes nevertheless apply to other DD approaches. In the presence of noisy pulses, DD does not always mitigate errors. It does so only when the added noise from the imperfect DD pulses does not outweigh the increased ability in averaging away the original background noise. We present break-even conditions that delineate when DD is useful, and further find that there is a limit in the performance of concatenated DD, specifically in how far one can concatenate the DD pulse sequences before the added noise no longer offers any further benefit in error mitigation.

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I. INTRODUCTION

The adverse effects of noise pose some of the biggest challenges in realizing useful quantum technologies. The very quantum effects that give quantum technologies their edge over classical devices are also the obstacles to success: They are extremely fragile and easily destroyed by the presence of unwanted interactions with the environment noise. Much of the current research and technological push in the community are centered around exploring ways to reduce and remove the effects of noise in quantum devices [1,2].

An effective noise-suppression technique is dynamical decoupling (DD), which requires the application of fast control pulse sequences on individual qubits to average away the effects of noise processes [3–11]. Advancing well beyond its root in control techniques for nuclear magnetic resonance, DD has been used in many different types of experiments as a viable way to combat decoherence in quantum information processing systems [12–18], alongside other applications such as noise spectroscopy [19] and quantum metrology [20]. Compared with quantum error correction (QEC) (see, for example, Ref. [21]), a more widely studied noise-removal approach, DD is much more economical as it requires no encoding of logical qubits using multiple physical qubits, nor real-time close-loop control through periodic syndrome measurement and recovery. All that is needed are regular single-qubit fast pulses that are usually easy to implement—they employ the same gates used for quantum computational tasks that are typically part of the capabilities of the quantum device. DD can be used by itself, or as the first layer of defense against noise within a standard QEC scheme as a hybrid noise-reduction approach [22-24]. The use of DD does not, however, come at no cost. The multiple pulses that have to be applied can be imperfect. Imperfect DD pulses can add, rather than remove, errors in the system. When the pulses are too noisy, those added errors can happen often enough to eliminate the benefit of having DD in the first place [23].

Like the concatenated codes in QEC, it is also possible to construct pulse schemes in a recursive manner to "scale up" the power of decoupling and form what is known as concatenated DD (CDD) [5]. In an ideal world, one could in principle construct arbitrarily accurate DD-protected gates through concatenation [25]. However, under realistic constraints such as control errors and finite pulse rate, both theoretical papers and experimental papers find that increasing concatenation may not always be beneficial [25–31].

In this work, we study the efficacy of DD as a noiseremoval technique when the DD pulses themselves are noisy by asking similar fault-tolerance questions usually asked of QEC procedures. Our investigation can be split into two lines of inquiry. First, for a given DD sequence, what is the maximum amount of noise in the DD pulses that can be tolerated, before DD stops offering any benefit? We refer to this maximum amount of tolerated noise as the break-even point (it is also sometimes referred to as the pseudothreshold in OEC and fault-tolerance literature). Second, we ask for the accuracy threshold—again borrowing terminology of QEC—specifying the level of noise permissible in the DD

pulses below which a given prescription for scaling up the DD scheme can remove more and more noise, and hence attain better and better computational accuracy in the quantum device. As we will see, we find that, in typical situations, there is no nonzero threshold. Instead, there is a maximal scale for DD, beyond which no additional benefit can be derived.

Past research involves some aspects of our queries. In particular, the performance of DD with imperfect pulses has been extensively studied with various models accounting for the effects of finite pulse width and systematic unitary rotation errors [26,32–35]. A generic description of noisy DD has also been attempted using a stochastic process model [35]. Experimental evidence suggesting the adverse impacts of noisy controls are also available in many reports [27,36,37]. Among these studies on nonideal DD, however, a comprehensive cost-benefit analysis from the fault-tolerance perspective is yet available to the best of our knowledge. Furthermore, for a fully fault-tolerant discussion relevant to experiments today, more general noises for the pulses have to be incorporated. Experimentally, gate-control noise, which can arise every time a DD pulse is applied, is often of a very different nature than the background noise—assumed in the case of finite pulse widths—in a quantum device, and can be highly dependent on the specific gate being applied. Such noise has to be treated separately in a realistic study of the efficacy of DD.

Here we focus our discussion on the most commonly used scheme of periodic DD (PDD) based on what is known as the universal decoupling sequence [3]. The same analysis, however, extends to other DD schemes. For scaling up the DD protection, to be able to remove more noise, we make use of CDD [5], organizing the DD pulse sequence in a recursive manner. We discuss the break-even point in two operational settings: computation and memory. In the computational setting, we consider using DD to reduce noise in computational gates in the course of carrying out a quantum circuit. The break-even comparison is thus about the noise per gate, with or without DD. In the memory setting, we instead compare the noise over a fixed time interval, during which DD can be carried out or not. Most of our discussion will be in the computational setting, the one most relevant to the current interest in quantum devices, though we mention the memory setting in specific cases. Following past papers, we quantify the performance of DD using the error phase, namely, the strength of the effective noise Hamiltonian—with and without DD—acting on the qubits in the quantum device. As we will see, the error phase permits easy analytical treatment. We numerically check for consistency with another natural figure of merit, the system state infidelity, in a specific physical setting.

We begin in Sec. II with a few basic concepts needed for our analysis. Section III discusses the break-even point for PDD, starting with the ideal-pulse situation before moving to the realistic noisy case. We examine the example of unitary errors in detail, numerically exploring the infidelity measure in addition to the analytical treatment of the error phase. Section IV extends our discussion to the case of CDD and explores the existence of an accuracy threshold for DD. We conclude and summarize our findings in Sec. V.

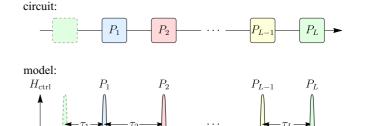


FIG. 1. Illustration of one cycle of a generic DD sequence comprising L pulses P_i , for i = 1, 2, ..., L. Pulse P_i is applied at time t_i ; pulses are separated by time τ_i . The dashed pulse at time t_0 is the final pulse of the previous DD cycle. H_{ctrl} is the control Hamiltonian that implements the DD pulses.

II. PRELIMINARIES

We begin our discussion with the introduction of a few basic concepts necessary for understanding the rest of the paper.

A. Basics of DD

DD involves the repeated application of a fixed sequence of short pulses (or fast gates) to individual quantum registers that average away the effect of any noise with a time scale slow compared to the sequence time. For a given DD scheme, let L be the length, i.e., the number of pulses, of the sequence. We denote the ith pulse of the sequence as P_i , and write the sequence as $P_L \cdots P_2 P_1$, proceeding from right to left in time. Pulse P_i is applied at time $t_i = t_{i-1} + \tau_i$, for $i = 1, 2, \ldots, L$, with τ_i the time between pulses P_{i-1} and P_i , and t_0 is the time of the last pulse of the previous sequence, coincident with the start time of the current sequence; see Fig. 1.

All DD schemes share some fundamental similarities. First, all DD pulses are chosen from a specific transformation group. Second, all DD sequences must satisfy the constraint that in the absence of errors, $P_L P_{L-1} \cdots P_1$ is the identity map (up to a phase factor). This constraint ensures that there is no net transformation on the quantum register at the completion of the sequence. Different strategies could differ in the following respects: (1) the transformation group—for a qubit register, a common choice is the Pauli group generated by the Pauli operators X and Z. Simpler schemes such as spin echo [38] and the CPMG sequence [39] use only the subgroup $\{I, Z\}$. (2) The specific sequence of pulses, which can be deterministic as well as randomized [6]. and (3) The pulse times—one can have regular-interval pulses, with $\tau_i \equiv \tau \ \forall i$, as is the case in PDD [3] and CDD [5] schemes. One could, however, have variable-interval schemes, such as the Uhrig DD sequence and its variants [7,11,40]. It is also possible to apply continuous control that follows a Eulerian cycle instead of using the "bang-bang"-style control [41]. In this work, we focus on the regular-interval schemes of PDD and CDD, differing in sequence length and the specific pulse sequence, but which employ pulses drawn from the Pauli group. The physical model we are considering is one where gates,

including the DD pulses, are applied at some operating frequency, with time τ in between consecutive gates. Such nonzero τ can simply be the finite switch time between gates, or there may be other practical reasons for a synchronized clock cycle time.

Generally speaking, the errors can be attributed to two major sources: (1) the system (quantum register) interacting with its environment (local quantum bath and field fluctuations, etc.), which constitutes the background noise on the system presenting even in the absence of control and (2) the control imperfections. In much past research, imperfect DD pulses were modeled as finite-width (or finite-duration) pulses during which the always-on system-bath interaction acts and leads to errors in the pulses. Our approach, explained in Sec. III B 1, allows additionally incorporating general control errors in the pulses as well. This second source of errors is perhaps more dominant and relevant in practical situations. In the followings, we refer to the DD sequences with perfect, instantaneous pulses as ideal DD, and to the case with imperfect pulses as noisy DD.

In the absence of any DD pulses, the system and bath evolve jointly according to the Hamiltonian H, assumed to be time-independent as is appropriate for standard DD analysis [42]. H here can be written as $H = H_{\rm B} + H_{\rm SB}$, with $H_{\rm B}$ as the bath-only Hamiltonian, and $H_{\rm SB}$ the interaction. No system-only term appears in H as we assume no nontrivial dynamics, other than that arising from $H_{\rm SB}$, occur in the system during the course of a DD sequence. Computational gates, if any, can be done only between complete DD sequences for the noise averaging to work. With this, we can write the evolution operator for a single complete ideal DD sequence as

$$U_{\rm DD} = P_L e^{-i\tau H} P_{L-1} \cdots P_2 e^{-i\tau H} P_1 e^{-i\tau H} \equiv e^{-i\Omega_{\rm DD}}. \tag{1}$$

Here we have defined $\Omega_{\rm DD} \equiv T H_{\rm eff}$ as the dimensionless effective Hamiltonian appropriate for describing the evolution for the time $T \equiv L\tau$ of the DD sequence. $\Omega_{\rm DD}$ can be written formally using the Magnus expansion,

$$\Omega_{\mathsf{DD}} = \sum_{m=1}^{\infty} \Omega_{\mathsf{DD}}^{(m)},\tag{2}$$

where the *m*th term consists of products of *m* copies of τH , and hence is of order $\|\tau H\|^m$. We refer the reader to past analyses of DD (see, for example, Ref. [23]) for a detailed derivation of the Magnus expansion. Here we provide only the basic expressions needed for our discussion below. In particular, we will need the expressions for the three lowest-order Magnus terms for piecewise-constant Hamiltonian evolution [43]. For $U = e^{-iO_K}e^{-iO_{K-1}}\cdots e^{-iO_1}$, for O_i s a sequence of time-independent dimensionless Hamiltonians (e.g., $O_1 = \tau H$), we have $U \equiv e^{-i\Omega}$ with $\Omega = \Omega^{(1)} + \Omega^{(2)} + \Omega^{(3)} + \cdots$, where

$$\Omega^{(1)} = \sum_{i=1}^{K} O_i,$$

$$\Omega^{(2)} = -\frac{i}{2} \sum_{i,j=1}^{K} [O_i, O_j],$$

and
$$\Omega^{(3)} = -\frac{1}{6} \sum_{\substack{i,j,k=1\\i\geqslant j\geqslant k}}^{K} \frac{([O_i,[O_j,O_k]] + [O_k,[O_i,O_j]])}{\operatorname{sym}(i,j,k)}.$$
 (3)

Here $\operatorname{sym}(i, j, k)$ is the symmetry factor (equaling to 1 when i, j, k are all different, 2 when any two indices are equal, and 6 when all indices are equal); $[\cdot, \cdot]$ denotes the commutator. Within the radius of absolute convergence of the Magnus series [44],

$$L\|\tau H\| < 1.0868 \dots \approx 1,\tag{4}$$

 $\Omega^{(m)}$ decreases in importance as m increases. A DD scheme such that $\Omega^{(m)}_{DD}$ acts trivially on the system (i.e., acts as the identity on the system) for all $m \le n$ is said to achieve nth-order decoupling. For such a scheme, the system sees an effectively weakened noise, of strength $\|\Omega^{(n)}_{DD}\| \sim \|\tau H\|^n$, compared with $\|\tau H\|$ without DD.

B. The PDD scheme

We specialize here to the case of interest, that of the single-qubit PDD scheme based on the universal decoupling sequence [3]. We refer to this case simply as PDD for brevity. The single qubit interacts with a bath, via a joint Hamiltonian (in the absence of DD) that can be written, without loss of generality, as

$$H \equiv I \otimes B_I + X \otimes B_X + Y \otimes B_Y + Z \otimes B_Z. \tag{5}$$

Here $I \equiv \sigma_0$ is the identity operator on the qubit, $X \equiv \sigma_1$, $Y \equiv \sigma_2$, and $Z \equiv \sigma_3$ are the Pauli operators on the qubit, and the B_{α} s are operators on the bath. We identify $I \otimes B_I$ as the bathonly Hamiltonian H_B , and $X \otimes B_X + Y \otimes B_Y + Z \otimes B_Z$ as the interaction Hamiltonian H_{SB} .

The universal decoupling sequence refers to a simple four-pulse sequence "Z-X-Z-X-." The corresponding Ω_{DD} of Eqs. (1), now relabeled as Ω_{PDD} , is expressible as the series $\Omega_{PDD} = \Omega_{PDD}^{(1)} + \Omega_{PDD}^{(2)} + \cdots$ whose first two terms can be worked out to be

$$\Omega_{\mathsf{PDD}}^{(1)} = (4\tau)I \otimes B_I, \tag{6}$$

$$\Omega_{PDD}^{(2)} = -(2\tau^2)\{X \otimes 2i [B_I, B_X]
+ Y \otimes (i [B_I, B_Y] + \{B_X, B_Z\})\},$$
(7)

where $\{\cdot, \cdot\}$ is the anticommutator.

Let us compare the PDD evolution with the evolution without DD over the same time period of 4τ (i.e., in the memory setting): $U=e^{-i4\tau H}\equiv e^{-i\Omega}$, with

$$\Omega = (4\tau)H = (4\tau)(I \otimes B_I + H_{SB}). \tag{8}$$

Comparing $\Omega_{\text{PDD}}^{(1)}$ —usually the dominant term—with Ω , we see that the H_{SB} in Ω no longer appears in $\Omega_{\text{PDD}}^{(1)}$ and $\Omega_{\text{PDD}}^{(1)}$ is trivial on the system. This corresponds to the fact that PDD is able to remove the lowest-order noise and that it achieves first-order decoupling.

C. Scaling up the protection with concatenation

A given DD sequence yields a given decoupling order, setting a limit on the scheme's ability to reduce noise in the

system. To increase the power of the DD scheme, one can employ the method of concatenation introduced in Ref. [5]. In that work, CDD was built upon the basic PDD scheme; the same procedure of concatenation, however, can be applied to other basic DD sequences as well [8,9]. The idea is to make use of concatenation to increase the decoupling order of the resulting DD sequence, hence reducing the residual noise.

CDD can be described in a recursive manner. We begin with the bare evolution, without any DD sequence, writing the evolution operator over time interval t as

$$U_0(t) \equiv e^{-itH_0},\tag{9}$$

where $H_0 \equiv H$, with the subscript 0 added here in preparation for concatenation to higher levels. With a DD scheme of L pulses, applied at time intervals (assuming regular-interval DD) τ_0 , the evolution operator is

$$U_1(\tau_1) \equiv P_L U_0(\tau_0) \cdots P_2 U_0(\tau_0) P_1 U_0(\tau_0), \tag{10}$$

where $\tau_1 \equiv L\tau_0$. The subscript 1 is to be understood as indicating that this is for concatenation level 1. To concatenate further, $U_k(\tau_k)$ is defined recursively,

$$U_k(\tau_k) \equiv P_L U_{k-1}(\tau_{k-1}) \cdots P_1 U_{k-1}(\tau_{k-1})$$

$$\equiv e^{-i\tau_k H_k} = e^{-i\Omega_{\text{CDD}k}} \quad (\text{for } k \geqslant 1), \qquad (11)$$

with $\tau_k \equiv L\tau_{k-1}$. For each $U_k(\tau_k)$, we also associate an effective Hamiltonian H_k , and a dimensionless Hamiltonian $\Omega_{\text{CDD}k} \equiv \tau_k H_k$. Each CDD scheme is determined by specifying the maximal concatenation level n, and either the value of τ_0 or τ_n . CDD at level n, denoted as CDDn, is then a sequence of $L_n \equiv L^n$ pulses separated by time interval τ_0 and taking total time $\tau_n = L^n \tau_0$ to complete.

In the remainder of the paper, we will restrict our discussion to CDD built upon the basic PDD scheme. The authors of Ref. [5] showed that CDD*n* achieves *n*th-order decoupling. This quantifies the benefit of scaling up the noise protection by concatenation. Appendix D rederives this conclusion with a different argument than that in the original reference.

D. Quantifying the efficacy of DD

We need concrete figures of merit to quantify the performance of DD. For most of the discussion, we will employ the error phase, which measures the strength of the effective noise Hamiltonian. In a specific example, we will also examine the infidelity measure, and compare the conclusions to those from error phase considerations.

1. Error phase

To gauge the efficacy of DD, we quantify the deviation of the actual state of the quantum system, with and without DD, from the ideal, no-noise state. Following Ref. [5], we make use of the *error phase*, which measures the strength of the system-bath interaction, the source of noise on the system. The system and bath evolve jointly for some specified time T according to the evolution operator U(0,T). The underlying joint Hamiltonian generating the dynamics can be time-dependent, and can include—or not—the DD pulses on the system. We write $U(0,T) \equiv e^{-i\Omega}$, for some effective dimensionless Hamiltonian Ω ; this can be thought of

as evolution according to a time-independent effective Hamiltonian Ω/T , for time T. Ω can be split into two pieces: $\Omega \equiv \Omega_{\rm B} + \Omega_{\rm SB}$, where $\Omega_{\rm B} \equiv \frac{1}{d_{\rm S}} I_{\rm S} \otimes {\rm tr}_{\rm S}(\Omega)$ ($d_{\rm S}$ is the dimension of the system) acts on the bath alone, while $\Omega_{\rm SB} \equiv \Omega - \Omega_{\rm B}$ contains all the pieces that act nontrivially on the system. $\Omega_{\rm SB}$ can be thought of as the effective system-bath interaction over this time T. We define the error phase $\Phi_{\rm SB}$ as the norm of $\Omega_{\rm SB}$, $\Phi_{\rm B}$ as the norm of $\Omega_{\rm SB}$, which as we will see, will also enter our analysis:

$$\Phi_{\rm B} \equiv \|\Omega_{\rm B}\| \quad \text{and} \quad \Phi_{\rm SB} \equiv \|\Omega_{\rm SB}\|.$$
(12)

We choose to employ the operator norm, i.e., the maximal singular value of an operator, for all norm symbols appearing in this paper. For this choice, we have unit norm for the identity operator in any dimension, so that the bath dimension is irrelevant in the definition of the error phase, a fact that will come in useful later. In what follows, we will use ϕ for the bare Hamiltonian associated with the no-DD situation. Specifically, we write

$$\phi_{\rm B} \equiv \|\tau H_{\rm B}\| \quad \text{and} \quad \phi_{\rm SB} \equiv \|\tau H_{\rm SB}\|,$$
 (13)

and exclusively reserve the uppercase Φ for the effective Hamiltonian after DD. The Hamiltonian operators are assumed to be bounded throughout this work, but there is otherwise no restriction on the dimension of the bath Hilbert space.

2. Infidelity measure

We mention another natural measure of noise, namely, the infidelity between the noisy (with or without DD) and ideal no-noise system states. Since we are interested only in how the system state is affected by noise, we care only about the effective noise channels ${\cal N}$ acting on the system, with or without the DD pulses, with the bath degrees of freedom discarded, i.e.,

$$\mathcal{N}(\,\cdot\,) \equiv \operatorname{tr}_{B} \{ e^{-i\Omega}(\,\cdot\,\otimes \rho_{B}) e^{i\Omega} \}$$
 and
$$\mathcal{N}_{DD}(\,\cdot\,) \equiv \operatorname{tr}_{B} \{ e^{-i\Omega_{DD}}(\,\cdot\,\otimes \rho_{B}) e^{i\Omega_{DD}} \}. \tag{14}$$

With these channels, we can define our infidelity measure as

$$InF \equiv \max_{\mathcal{N}, \psi} InF(\mathcal{N}, \psi), \tag{15}$$

with $InF(\mathcal{N}, \psi) \equiv \sqrt{\langle \psi | (\mathcal{I} - \mathcal{N})(\psi) | \psi \rangle}$, noting $\langle \psi | (\mathcal{I} - \mathcal{N})(\psi) | \psi \rangle = 1 - \langle \psi | \mathcal{N}(\psi) | \psi \rangle \in [0, 1]$, so that InF and $InF(\mathcal{N}, \psi) \in [0, 1]$. We have an analogous expression for InF_{DD} computed from \mathcal{N}_{DD} . Here $\psi \equiv |\psi\rangle\langle\psi|$ is a pure system-only state. InF is the square root (as we will see, the square root gives the proper comparison with the error phase) of the deviation from 1 of the square of the fidelity between the postnoise state and the initial state. The maximization over $\mathcal N$ (and $\mathcal N_{DD}$) refers to a maximization over all choices of the B operators that enter $H = H_B + H_{SB}$, with fixed ϕ_B and ϕ_{SB} values. The maximization—resulting in a worst-case measure—over all pure system states provides a state-independent quantification, while the maximization over the Ns reflects our typical lack of knowledge of the precise forms of the B operators, even if we are given the $\phi_{\rm B}$ and ϕ_{SB} values.

One can obtain analytical bounds between the error phase and infidelity measures. Appendix A derives such a bound by writing the output state $\mathcal{N}(\rho_S)$ as a power series in Ω . We find the relation

$$0 \leqslant \text{InF}(\mathcal{N}, \psi) \lesssim \sqrt{2}\phi_{\text{SB}}.$$
 (16)

Since we allow any general channel \mathcal{N} here, analogous bounds also hold for InF_{DD} and Φ_{SB} .

III. LIMITS OF DECOUPLING: THE BREAK-EVEN CONDITIONS

Any noise mitigation strategy is effective only if its costs, i.e., the increased complexity of carrying out the computational task with noise mitigation, are lower than its benefits, i.e., the increased ability to reduce the adverse effects of noise on the computation. The costs enter not just because any noise mitigation approach requires the use of more resources, e.g., more gates, more qubits, etc., but also that those added resources are themselves imperfect in practice, such that the unmitigated noise of the resulting larger system is unavoidably larger than if no noise mitigation strategy was adopted. There is then a net benefit only if the added noise is low enough to not overwhelm the added noise-removal capabilities. Such is the content of any fault-tolerance analysis (see, for example, Ref. [21]), usually applied to quantum computing tasks protected by quantum error correction. Here we apply the same logic to DD, and ask when the added benefit of averaging away part of the noise outweighs the added cost of having to do additional pulses which are themselves noisy.

For a proper cost-benefit analysis, we distinguish between two operational scenarios: to better preserve states in a quantum memory (no computational gates), or to reduce noise in the course of carrying out a quantum computation. In the memory setting, the goal is to preserve an arbitrary quantum state for some storage time T. During this period, DD pulses are applied, and we can ask if the resulting noise is lower compared with the no-DD case over the same time period. In other words, for DD to work well, we require

Memory setting:
$$\epsilon_{DD}(T) < \epsilon(T)$$
, (17)

where $\epsilon(t)$ is some chosen figure of merit quantifying the noise associated with the time evolution over period t, with or without DD as specified by the subscript. In particular, for the quantum memory protected with an L-pulse DD scheme with constant interval τ , we have $T=L\tau$. In the computational setting, instead of the same total evolution time with and without DD, we need to pay attention to the gate time, assumed to be common to both computational and DD gates. In the absence of DD, we assume computational gates are applied at the same rate as DD pulses, i.e., every time step τ . With DD, however, the protected computational gates are further separated in time by a factor L, the number of pulses in the DD cycle. The condition for DD to be effective in this case is then

Computational setting:
$$\epsilon_{DD}(L\tau) < \epsilon(\tau)$$
. (18)

In our work, we are more interested in the computational setting, the more stringent one among the two, though it is straightforward to adapt our analyses to the memory case, as we do in some of the situations below. As we will see, condition (18) will yield a requirement on the noise parameters characterizing the noise in the system and the DD pulses. We refer to this requirement as the break-even condition for DD, borrowing terminology from quantum error correction and fault-tolerant quantum computing.

A. Ideal case

Let us first discuss the break-even condition for ideal PDD. We expect to recover the usual statement of when DD works at all, namely when the noise changes slowly compared with the time taken for a complete DD sequence. We choose the error phase as our figure of merit: $\epsilon \equiv \Phi$. For PDD to be useful, we require, under the computational setting,

$$\Phi_{\rm SB} \leqslant \phi_{\rm SB},\tag{19}$$

where Φ_{SB} is the error phase with PDD, while ϕ_{SB} , as defined earlier, is $\|\tau H_{SB}\|$, the error phase without DD, for a single time step τ . From our earlier discussion of PDD, we have $\Phi_{SB} = \|\Omega_{PDD}^{(2)} + \Omega_{PDD}^{(3)} + \cdots\|$. The full Magnus series is difficult to write out, but we can employ the dominant nontrivial (on the system) term, $\Omega_{PDD}^{(2)}$, for the approximate condition:

$$\Phi_{\rm SB} \simeq \|\Omega_{\rm PDD}^{(2)}\| \leqslant \phi_{\rm SB}.\tag{20}$$

Using the expression for $\Omega^{(2)}_{PDD}$ from Eq. (7), we have

$$\|\Omega_{\mathsf{PDD}}^{(2)}\| \leq 2\tau^{2}(2\|[B_{I}, B_{X}]\| + \|[B_{I}, B_{Y}]\| + \|\{B_{X}, B_{Z}\}\|)$$

$$\leq 4\phi_{\mathsf{B}}\tau(2\|B_{X}\| + \|B_{Y}\|) + 4\tau^{2}\|B_{X}\|\|B_{Z}\|$$

$$\leq 12\phi_{\mathsf{B}}\phi_{\mathsf{SB}} + 4\phi_{\mathsf{SR}}^{2}, \tag{21}$$

noting that $\phi_B \equiv ||\tau H_B|| = ||\tau B_I||$, and that $\phi_{SB} \equiv ||\tau H_{SB}|| \ge ||\tau B_i||$ for i = X, Y, Z (see Appendix B). For the break-even condition (19), it then suffices to require

$$12\phi_{\rm B} + 4\phi_{\rm SB} \leqslant 1. \tag{22}$$

Condition (22) amounts to a requirement that both ϕ_{SB} and $\phi_{\rm B}$ be small for PDD to work well. We note that this condition is also sufficient for the convergence criterion, Eq. (4), as $4\|\Omega\| \le 4\phi_{\rm B} + 4\phi_{\rm SB} \le 1$, which in turn justifies the approximation of Eq. (20). The coefficients for ϕ_B and ϕ_{SB} contain a factor of 4 from the length of the PDD sequence. In general, a longer sequence will have a larger prefactor and hence put a more stringent requirement on the noise to be small. That ϕ_{SB} has to be small comes as no surprise: ϕ_{SB} quantifies the strength of the noise on the system, and DD is expected to work well as long as the noise is weak enough so that the remnant noise is small. The requirement that $\phi_{\rm B}$ be small is perhaps more surprising. After all, it bounds the bath-only term which does not directly lead to noise on the system. Nevertheless, $H_{\rm B}$ determines the evolution rate of the bath (which can be made explicit in the Heisenberg picture of H_{SB} defined by $H_{\rm B}$), while τ is the inverse rate for the control pulse. Hence $\phi_{\rm B}$ quantifies how rapidly the bath evolves relative to the pulse, or how fast the noise is compared with the control. That $\phi_{\rm B}$ should also be small should be understood as the additional requirement that the characteristic frequency of the bath should be low compared with the control, for good averaging over the entire sequence. These two requirements

are consistent with the understanding of DD from the filter-function formalism [19].

B. Noisy DD gates

Next, we accommodate the possibility of noise in the application of the DD pulses. A realistic—imperfect or noisy—DD pulse takes finite time to complete, during which the always-on background $H_{\rm SB}$ interaction acts, leading to noise in the applied pulse. Control errors in the course of the gate application contribute to additional imperfections.

1. Noise model

We describe a noisy DD pulse in the following manner. We denote the actual ith DD pulse by $\widetilde{\mathcal{P}}_i$, now regarded as a map on states, not just a unitary operator. $\widetilde{\mathcal{P}}_i$ can be obtained in an experiment through the use of process tomography methods, giving generally a completely positive (CP) and trace-preserving (TP) map [45]. $\widetilde{\mathcal{P}}_i$ can in reality be time-dependent in that the noisy pulse may be different in each DD cycle, but for simplicity, we assume no such time dependence. Each pulse is assumed to take time τ_P (the pulse width) to complete, during which a gate Hamiltonian H_i acts. In the ideal situation, $P_i = e^{-i\tau_P H_i}$; in reality, during the gate action, the background Hamiltonian $H = H_{\rm B} + H_{\rm SB}$ acts as well, and there can be additional gate control noise. We hence write the noisy pulse as

$$\widetilde{\mathcal{P}}_i = \mathcal{U}_i \circ \mathcal{E}_i. \tag{23}$$

Here $\mathcal{U}_i(\cdot) \equiv U_i(\cdot)U_i^\dagger$ with $U_i \equiv e^{i\tau_P(H_i+H)}$, is the unitary map that accounts for the finite width of the DD pulse, with noise coming from the background Hamiltonian H. $\mathcal{E}_i \equiv \mathcal{U}_i^\dagger \circ \widetilde{\mathcal{P}}_i$ is a CPTP "error" map that captures the gate control noise. \mathcal{E}_i acts only on the system in the typical case. Splitting $\widetilde{\mathcal{P}}_i$ into the two maps as in Eq. (23) reflects the physical nature of the two noise sources and what we typically know about them: The noise arising from the background Hamiltonian scales as the pulse width, while the control noise (e.g., finite turn-on and -off times, misalignment issues, etc.) often does not. While one can view \mathcal{E}_i (as we do below, mathematically) as one generated by a noise Hamiltonian, one often does not know that Hamiltonian source of noise but obtains \mathcal{E}_i from tomography alone, as a map for the time step associated with the pulse.

Now, if \mathcal{E}_i is a unitary map, then $\mathcal{E}_i(\cdot) = V_i(\cdot)V_i^{\dagger}$ with V_i a unitary operator on the system. We can write $V_i \equiv e^{-i\Gamma_i}$, where Γ_i is a Hermitian operator to be viewed as the dimensionless effective Hamiltonian for the control noise in the ith pulse. Even for a nonunitary CPTP map \mathcal{E}_i (the typical case), we can "dilate" the system to include an ancillary system such that \mathcal{E}_i arises from a unitary map on the system-ancilla composite, with \mathcal{E}_i obtained after tracing over the ancilla. According to the Stinespring dilation theorem of a quantum channel, it suffice to incorporate at most a d^2 -dimensional ancilla space for the d-dimensional system [46]. For the analytical treatment below using the error phase, we will always assume this dilation, and treat all noise as unitary \mathcal{E}_i with a dimensionless effective Hamiltonian Γ_i . The ancillary system needed for the dilation is treated as part of the bath. We thus write the noisy pulse finally as $\widetilde{\mathcal{P}}_i(\cdot) = \widetilde{P}_i(\cdot)\widetilde{P}_i^{\dagger}$, with \widetilde{P}_i a unitary operator on the system and bath defined as

$$\widetilde{P}_i \equiv e^{-i\tau_P(H_i + H)} e^{-i\Gamma_i}.$$
(24)

Note that every Γ_i can be taken to have no pure-bath term, i.e., it does not have a term proportional to I on the system. Such a term gives rise only to an overall phase on the system and cannot result in observable imperfections in the pulses.

We assume that $\|\Gamma_i\| \leqslant \eta$ for all i, where η is a dimensionless parameter that carries the meaning of a control-noise strength. In experimentally relevant scenarios, we expect η to be small. Also, τ_P is usually chosen to be small compared to τ , for fast pulses that minimize the effect of H while the pulse Hamiltonian acts. We define $\delta \equiv \tau_P/\tau \ll 1$. There is no particular relation between the small parameters η and δ , and we will consider the leading-order contributions from both terms.

2. Noisy PDD

With noisy and finite-width pulses, the evolution of the system and bath under DD can be described using the time evolution operator,

$$\widetilde{U}_{\text{DD}} \equiv \widetilde{P}_{K} e^{-i(\tau - \tau_{P})H} \widetilde{P}_{K-1} \cdots \widetilde{P}_{2} e^{-i(\tau - \tau_{P})H} \widetilde{P}_{1} e^{-i(\tau - \tau_{P})H}
\equiv e^{-i\widetilde{\Omega}_{\text{DD}}}.$$
(25)

This reduces to the ideal case of Sec. III A when Γ_i and τ_P both vanish.

For PDD, the P_i sequence comprises only X and Z pulses. We assume that the X pulses suffer the same noise each time they are applied, as do the Z pulses. We write the noisy X and Z pulses as

$$\widetilde{Z} = e^{-i\tau_P(H_Z + H)} e^{-i\Gamma_Z}$$
 and $\widetilde{X} = e^{-i\tau_P(H_X + H)} e^{-i\Gamma_X}$, (26)

with $H_X \equiv \frac{\pi}{2\tau_P} X$ and $H_Z \equiv \frac{\pi}{2\tau_P} Z$. We can rewrite the time evolution as

$$\widetilde{U}_{PDD} = \widetilde{Z}e^{-i(\tau - \tau_P)H}\widetilde{X}e^{-i(\tau - \tau_P)H}\widetilde{Z}e^{-i(\tau - \tau_P)H}\widetilde{X}e^{-i(\tau - \tau_P)H}$$

$$\equiv e^{-iK_3}e^{-i\Omega_3}e^{-iK_2}e^{-i\Omega_2}e^{-iK_1}e^{-i\Omega_1}e^{-iK_0}e^{-i\Omega_0}, \quad (27)$$

where $\Omega_{\alpha} \equiv \tau \sigma_{\alpha} H \sigma_{\alpha}$ for $\alpha = 0, 1, 2, 3$, and the K_{α} s are defined as $e^{-iK_0} \equiv X \widetilde{X} e^{i\tau_P H}$, $e^{-iK_1} \equiv Y \widetilde{Z} e^{i\tau_P H} X$, $e^{-iK_2} \equiv Z \widetilde{X} e^{i\tau_P H} Y$, and $e^{-iK_3} \equiv \widetilde{Z} e^{i\tau_P H} Z$. Following the calculation detailed in Appendix C, we find, to the lowest order in η and $\phi_{\rm SB} \delta (= \|\tau_P H_{\rm SB}\|)$,

$$K_{0} \simeq \Gamma_{X} + \tau_{P} \{-Y \otimes [B_{Y} - (2/\pi)B_{Z}]$$

$$-Z \otimes [B_{Z} + (2/\pi)B_{Y}] \},$$

$$K_{1} \simeq X\Gamma_{Z}X + \tau_{P} \{-X \otimes [B_{X} - (2/\pi)B_{Y}]$$

$$+ Y \otimes [B_{Y} + (2/\pi)B_{X}] \},$$

$$K_{2} \simeq Y\Gamma_{X}Y + \tau_{P} \{-Y \otimes [B_{Y} - (2/\pi)B_{Z}]$$

$$+ Z \otimes [B_{Z} + (2/\pi)B_{Y}] \},$$

$$K_{3} \simeq Z\Gamma_{Z}Z + \tau_{P} \{X \otimes [B_{X} - (2/\pi)B_{Y}]$$

$$+ Y \otimes [B_{Y} + (2/\pi)B_{X}] \}.$$
(28)

With this, we can calculate the first-order Magnus term for noisy PDD as

$$\widetilde{\Omega}_{PDD}^{(1)} = \sum_{\alpha=0}^{3} (\Omega_{\alpha} + K_{\alpha}) \simeq (4\tau)I \otimes B_{I} + (4/\pi)\tau_{P}Y$$

$$\otimes (B_{X} + B_{Z}) + \Gamma_{X} + X\Gamma_{Z}X + Y\Gamma_{X}Y + Z\Gamma_{Z}Z.$$
(29)

Unsurprisingly, we no longer have exact first-order decoupling, with deviations of order $\phi_{SB}\delta$ and η .

Observe that, if $\Gamma_X = \Gamma_Z$, the Γ_i terms in Eq. (29) cancel, leaving only the finite-width term that goes as τ_P as the sole deviation from exact first-order decoupling. Thus, gate-independent control noise affects PDD only at order η^2 and higher. We can understand this as the PDD sequence averaging away also the constant—over the PDD sequence—noise that comes from the imperfect pulses, just as it does the time-independent $H_{\rm SB}$. Put differently, the deviation from first-order decoupling from control noise enters only as the difference between the X and Z control noise, so that we can rewrite $\widetilde{\Omega}_{\rm PDD}$ as

$$\widetilde{\Omega}_{PDD}^{(1)} \simeq (4\tau)I \otimes B_I + (4/\pi)\tau_P Y \otimes (B_X + B_Z)$$

$$+ X(\Gamma_Z - \Gamma_X)X + Z(\Gamma_Z - \Gamma_X)Z.$$
(30)

Ignoring higher-order Magnus terms, we can extract a sufficient break-even condition,

$$(8/\pi)\phi_{SB}\delta + 4\eta \leqslant \phi_{SB}. \tag{31}$$

Here we have upper-bounded $\|\Gamma_Z - \Gamma_X\|$ by 2η , the most general bound when there is no particular relation between Γ_Z and Γ_X , as would be the case for platforms where the two gates are applied by different methods. If, on the other hand, the control noise is only weakly gate-dependent, namely, that $\|\Gamma_Z - \Gamma_X\| \le \eta' \ll \eta$, one can replace the 4η in the break-even condition Eq. (31) by $2\eta'$. In this case, η' may also be comparable to the second-order Magnus terms (e.g., compared with $\phi_{\rm SB}^2$), and higher order corrections should be considered and included in the break-even condition. We explore a concrete example of this in Sec. III C.

In the absence of gate control noise, i.e., $\Gamma_X = 0 = \Gamma_Z$, or, equivalently, $\eta = 0$, we return to the case of finite-width pulses already discussed in past literature. In this scenario, the break-even condition becomes simply $\delta \leqslant \frac{\pi}{8}$, or, in terms of τ_P itself,

$$\tau_P \leqslant \frac{\pi}{8} \tau \simeq 0.4 \tau.$$
(32)

In the opposite limit, if there is control noise, but pulses are instantaneous, i.e., $\tau_P = 0$, we have the break-even condition,

$$\eta \leqslant \frac{1}{4}\phi_{\rm SB}.\tag{33}$$

C. Example: Unitary error and cross-measure consistency

As a concrete example of our discussion of break-even conditions for PDD, let us examine the situation of a unitary—on the system only—gate control noise, arising, for example, from a systematic calibration error in the pulse control. At the same time, we investigate the infidelity measure for ϵ and

 ϵ_{DD} , as an alternative to the error phase, and compare the conclusions about the break-even conditions.

For simplicity, we assume instantaneous pulses $\tau_P = 0$ (since finite-width errors are already explored in existing literature) and that the noise is gate independent. Specifically, we write

$$\Gamma_X = \Gamma_Z = \eta \, \mathbf{n} \cdot \boldsymbol{\sigma},\tag{34}$$

with $\sigma \equiv (\sigma_X, \sigma_Y, \sigma_Z) = (X, Y, Z)$, $\mathbf{n} \equiv (n_X, n_Y, n_Z)$ is a real unit vector, and η is a nonnegative constant quantifying the strength of the unitary error. Here the Γ s act only on the system and not on the bath. Since $\tau_P = 0$ and the noise is gate independent, the first-order Magnus contribution to the effective with-DD interaction vanishes: $\widetilde{\Omega}_{PDD}^{(1)}$ of Eq. (30) reduces to a bath-only term. The error phase is then second order in small quantities, and we need the second-order Magnus term to deduce the effective interaction Hamiltonian. Direct calculation yields

$$\widetilde{\Omega}_{PDD}^{(2)} = -4X \otimes i\tau^{2}[B_{I}, B_{X}] - 2Y \otimes (i\tau^{2}[B_{I}, B_{Y}] + \{\tau B_{X} + \eta n_{X}I, \tau B_{Z} + \eta n_{Z}I\}).$$
(35)

We can then bound the error phase in a straightforward manner, with triangle inequalities, as

$$\Phi_{SB} \simeq \|\widetilde{\Omega}_{PDD}^{(2)}\| \leqslant 2\eta^2 + 8\eta\phi_{SB} + 12\phi_B\phi_{SB} + 4\phi_{SB}^2
= 2(\eta + 2\phi_{SB})^2 + 4\phi_{SB}(3\phi_B - \phi_{SB}),$$
(36)

where we have used the fact that $|n_X n_Z| \le \frac{1}{2}$ since **n** is a unit vector. A sufficient break-even condition, neglecting higher-order corrections, can be obtained by restricting the above upper bound on the error phase Φ_{SB} to be no larger than ϕ_{SB} . This gives the regions bounded by the white-dashed lines in the lower-left corners of Figs. 2(a)–2(c); for Figs. 2(d) and 2(e), the break-even condition cannot be satisfied for the chosen parameters.

One can regard $\Phi_{SB} \leqslant \phi_{SB}$ as a threshold condition for the gate control error η , for given ϕ_{SB} and ϕ_{B} . Using the approximation in Eq. (36) for the error phase, this translates into the condition

$$\eta \leqslant \sqrt{\frac{1}{2}\phi_{\rm SB}}[1 - 4(3\phi_{\rm B} - \phi_{\rm SB})]^{-1/2} - 2\phi_{\rm SB}$$
(37)

for the break-even point. For ϕ_B and $\phi_{SB} \ll 1$, as is usually the case in practice, this becomes $\eta \lesssim \sqrt{\phi_{SB}/2}$. The square-root relation between η and ϕ_{SB} comes directly from the fact that the gate-independent noise matters only at the second order when there is DD, compared with ϕ_{SB} that enters in the first order without DD.

To gauge how tight our analytical bounds are, which are general and applicable for all H and unitary Γ , we numerically simulate the action of PDD for specific instances to find the break-even conditions. At the same time, numerical analysis permits the investigation of figures of merit to quantify the performance of PDD, beyond the analytically accessible error phase that we have used thus far. Our conclusions about the break-even conditions are useful only if they are reasonably consistent across different measures. Specifically, we examine

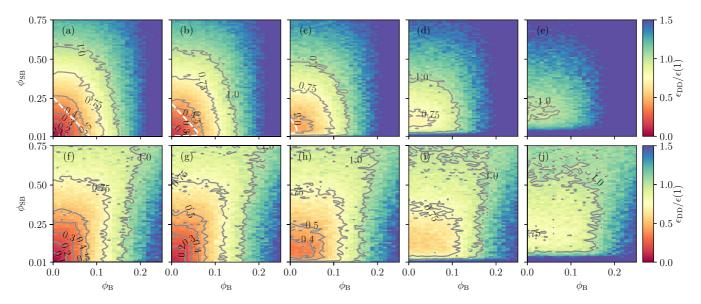


FIG. 2. Ratio $\epsilon_{DD}/\epsilon(1)$ for PDD, with unitary system-only gate control noise [Eq. (34)] and instantaneous pulses, for ϵ taken to be the error phase (top row) and infidelity (bottom row). The strengths of gate control noise are, from left to right, $\eta = 0, 0.02, 0.06, 0.1$, and 0.14. The ratio = 1 contours give the break-even conditions. The white-dashed lines in (a)–(c) mark the analytical break-even condition of Eq. (37).

the infidelity measure, already introduced in Sec. II D 2, in addition to the error phase figure of merit.

We numerically simulate the action of PDD on a single system qubit, for the gate-independent unitary noise of Eq. (34). The bath is also taken to be a single qubit. The system-bath Hamiltonian takes the form of Eq. (5), with the bath operators chosen randomly, with specified values of the norms, $\phi_B \equiv \|H_B\| = \|I \otimes B_I\|$ and $\phi_{SB} \equiv \|H_{SB}\| = \|\sum_i \sigma_i \otimes B_i\|$. B_I is chosen such that $\operatorname{tr}(B_I)$ vanishes, corresponding to fixing a choice for the zero-energy level. Specifically, we write $B_I = \phi_B \mathbf{v} \cdot \boldsymbol{\sigma}$ where \mathbf{v} is a real 3D unit vector chosen uniform-randomly over the surface of a 3-sphere. In addition, we write $B_i = R_i + R_i^{\dagger}$, where R_i is a complex random matrix with normally distributed entries. The resulting H_{SB} is normalized and then multiplied by a chosen ϕ_{SB} to give the desired magnitude.

The error phase approach requires no specification of the initial bath state, as it quantifies the size of the full systembath interaction Hamiltonian. The infidelity measure that we compute here, however, targets only the resulting system-only state, and the effective system-only noise channels depend on the choice of initial bath state ρ_B . In those cases, we consider the infinite temperature state, i.e., the maximally mixed state, as a symmetric choice of bath state. The maximization contained in the definition of InF is carried out numerically with 10 000 samples over $\mathcal N$ and ψ .

In Fig. 2 we plot the numerical values of the ratio $\epsilon_{\text{PDD}}/\epsilon(1)$, for the two different choices of the figure of merit ϵ , over the parameter space (ϕ_B, ϕ_{SB}) , and for different η values. Figures 2(a)–2(e) (top row), plotting the error phase results, show a rapid shrinking of the break-even region, where DD remains effective, as η increases from 0 to 0.14. The numerical break-even regions are much larger than those predicted by the analytical condition of Eq. (37). The infidelity measure is plotted in Figs. 2(f)–2(j) (bottom row). With InF, the break-even regions are larger, indicating a difference in detailed conclusion from the error-phase measure. The contour

shapes nevertheless follow a similar pattern as that of the error phase measure, suggesting a qualitative agreement between the two measures.

IV. SCALING UP PROTECTION: LIMITS OF CONCATENATION

Next, we turn to the situation of CDD, where we attempt to scale up the noise protection by concatenating a basic decoupling sequence to longer and more sophisticated sequences. The scaling strategy differs between the quantum memory scenario and the quantum computation scenario, as illustrated in Fig. 3. In the memory setting, the natural strategy is to slice the fixed evolution time T into finer and finer intervals to accommodate more and more DD pulses, until some minimal gate time is reached. In the computational setting, with the working gate time τ fixed, increasing the concatenation level will inevitably bring about a longer time between computational gates. Past analytical and numerical studies gave rather optimistic assessments of the scaling performance using the memory setting [5,25,27], while experimental and numerical studies in the computational setting produced mixed conclusions [26,30,32]. Below we put our emphasis on the computational setting, the more relevant scenario for the current pursuit of high-fidelity gates for quantum computational tasks. We also adapt our analyses to the memory case for comparison.

One can certainly discuss the break-even point for a particular CDDn sequence, treating it simply as a "flattened" sequence of pulses, and asking when the error phase after CDDn is smaller than ϕ_{SB} as was done for PDD in the previous section. However, a more interesting question, given the CDD scheme of scaling up the protection by increasing the concatenation level, is the gain in noise-removing power for every increase in CDD level. A key concept in fault-tolerant quantum computing is the accuracy threshold, the noise strength below which scaling up the QEC code always gives

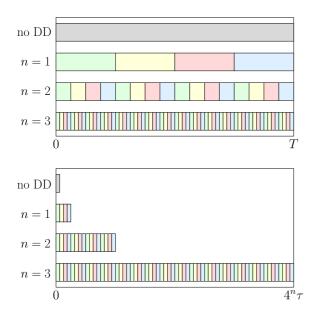


FIG. 3. An illustration of the CDD sequences for different concatenation levels. We compare the memory setting (top panel) with the computational setting (bottom panel).

improved protection against noise, and hence more accurate quantum computation (see, for example, Ref. [21]). Here we ask the analogous question of CDD: Is there a condition on the noise parameter of the problem such that increasing the CDD level always leads to improved noise removal capabilities? Concretely, an accuracy threshold for CDD exists if the physical noise strength—denoted symbolically here by η , though there can be many noise parameters that characterize the noise strength—satisfies a condition $\eta < \eta_{\text{thres}}$, for some $\eta_{\text{thres}} > 0$, such that

$$\Phi_{SB,n+1} < \Phi_{SB,n}, \tag{38}$$

for every $n=1,2,\ldots$ Here $\Phi_{\mathrm{SB},n}$ denotes the error phase for CDDn. The existence of an accuracy threshold means that every increase in CDD level is accompanied by a decrease in the resulting error phase, as long as the physical noise strength is below the threshold level. Note that Eq. (38) automatically implies the break-even condition by recognizing that $\Phi_{\mathrm{SB},0}=\phi_{\mathrm{SB}}$.

Unfortunately, as we will show below, for CDD built from concatenating the PDD scheme, there is no such accuracy threshold in the computational setting, i.e., there is no nonzero η_{thres} for which Eq. (38) holds for every n. Instead, we will see that the error phase initially decreases as n increases but eventually, this decrease turns around, and there is a maximal concatenation level beyond which the error phase actually grows with n. This maximal concatenation level hence quantifies the limits to the power of CDD, for given physical noise parameters. Below we show this general behavior for both the ideal CDD case where pulses are perfect, and the noisy CDD case where imperfections in the pulses are allowed.

As mentioned earlier, both ϕ_B and ϕ_{SB} have to be small for DD to offer benefits. Depending on the relative sizes of these two terms, there are two very different regimes for understanding the CDD performance: $\phi_B/\phi_{SB} \ll 1$ and

 $\phi_{\rm B}/\phi_{\rm SB}\gtrsim 1$. Reference [5], the original work that introduced CDD, focused on the limit of $\phi_{\rm B}/\phi_{\rm SB}\gtrsim 1$ (specifically, they assume $\phi_{\rm B} > \phi_{\rm SB}$ while both are small), a limit they argue to be relevant when the system is coupled only to a small portion of the bath while the whole bath, with many degrees of freedom, has a nonzero self-interaction. The other limit of $\phi_B/\phi_{SB} \ll 1$ is also potentially of physical interest, and represents an arguably better situation for DD, where the noise evolution is of secondary importance to the actual interaction between the system and bath. Here we reexamine the situation of $\phi_{\rm B}/\phi_{\rm SB}\gtrsim 1$ as done in [5] but now with gate-control noise. We also consider the opposite limit of $\phi_B/\phi_{SB} \ll 1$ for the ideal case; we will see that the qualitative behavior is, in fact, not that different. In that case, to obtain the leadingorder behavior, we can simply assume $B_I = 0$ in the physical Hamiltonian. The practical situation will likely be somewhere between the two limits.

A. Ideal case

1. $\phi_{\rm B}/\phi_{\rm SB}\gtrsim 1$

We first work in the limit of $\phi_B/\phi_{SB} \gtrsim 1$ examined in Ref. [5], while distinguishing between the computational and memory settings. Here the \gtrsim symbol specifies that $\phi_B > \phi_{SB}$ but both are similar in magnitude so that both can be considered to be of the same order in a perturbative expansion.

We focus on the computational setting for now. Following the analysis of Ref. [5], the effective Hamiltonian for CDDn can be written as

$$\Omega_{\text{CDD}n} = I \otimes 4^{n} \tau B_{I} + X \otimes (-i)^{n} 2^{n(n+1)} \tau^{n+1} a d_{B_{I}}^{n}(B_{X})
+ Y \otimes (-i)^{n} 2^{n^{2}} \tau^{n+1} a d_{B_{I}}^{n-1}([B_{I}, B_{Y}] - i\{B_{X}, B_{Z}\})
+ (higher order terms).$$
(39)

We confirm this leading-order behavior in Appendix D, an analysis needed also for our $B_I = 0$ discussion below. CDDn thus indeed achieves nth-order decoupling in that the lowest-order system-bath interaction term is (n + 1)th order in B_I and/or B_{α} ; lower-order terms have been eliminated by the DD sequence. The pure-bath part, whose size can be approximated by

$$\Phi_{B,n} \simeq 4^n \phi_B, \tag{40}$$

is of the first order. The remnant interaction part can be bounded in a straightforward manner using the leading-order terms in Eq. (39), giving the error phase

$$\Phi_{\mathrm{SB},n} \lesssim 2^{(n+1)^2} \phi_{\mathrm{R}}^n \phi_{\mathrm{SB}}.\tag{41}$$

That ϕ_B enters the error phase should again be of no surprise. As in PDD, ϕ_B determines how quickly the noise seen by the system evolves, and hence affects the efficacy of DD designed to eliminate noise that remains unchanged for the full DD sequence. ϕ_B has to be small for DD to work well. What is perhaps more surprising is the $2^{(n+1)^2}$ factor in $\Phi_{SB,n}$. The origin of this factor lies in the exponentially increasing temporal length of the CDD pulse sequence as n increases, for fixed $\tau_0 \equiv \tau$ (implicit in the ϕ_B and ϕ_{SB} quantities). This means that the error phase eventually increases for large enough n, as the exponentially decreasing ϕ_B^n factor

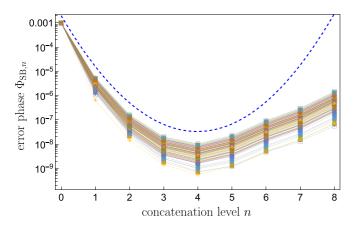


FIG. 4. The error phase $\Phi_{SB,n}$ as a function of the CDD concatenation level. The thin colored lines are obtained from random samples of the Hamiltonian H satisfying $\phi_B = \phi_{SB} = 0.001$. The blue dashed line is our theoretical upper bound for the error phase. The maximal concatenation level occurs at 4 for these parameter values.

is eventually overcome by the super-exponentially increasing $2^{(n+1)^2}$ factor. This tells us, as anticipated in the introductory paragraphs to this section, that *there is no accuracy threshold*: There are no nonzero values for ϕ_B and ϕ_{SB} below which $\Phi_{SB,n+1} \leqslant \Phi_{SB,n}$ for *all n*. Instead, there is a maximal useful level of CDD, beyond which further concatenation actually increases the noise seen by the system. Figure 4 plots this situation of fixed τ_0 , for increasing concatenation level *n*. We observe the initial decrease of $\Phi_{SB,n}$ as *n* increases, but this turns around eventually (at n=4 for the plotted situation).

Of course, the above discussion is based on the upper bound for the error phase. We can only conclude that there is no threshold level of noise below which the *upper bound* on error phase decreases as n increases. A more careful discussion of the threshold should look at the original $\Omega_{\text{CDD}n}$ expression, rather than a bound. For that, we examine the actual map $\Omega_{\text{CDD}n+1} \equiv \mathcal{D}(\Omega_{\text{CDD}n})$ that recursively gives the evolution as the CDD level increases, leading to Eq. (39). Stated in terms of the bath operators and neglecting higher-order terms, the map \mathcal{D} is given by (see Appendix D), for $n \geqslant 0$,

$$\tau B_{n+1,I} = 4\tau B_{n,I} = 4^{n+1}\tau B_{I},
\tau B_{n+1,X} = -4i[\tau B_{n,I}, \tau B_{n,X}],
\tau B_{n+1,Y} = -2i[\tau B_{n,I}, \tau B_{n,X}],
\tau B_{n+1,Z} = 0,$$
(42)

where $B_{n,\alpha}$, for $\alpha = I, X, Y, Z$, refers to the bath operators in $\Omega_{\text{CDD}n}$ associated with σ_{α} on the system for CDDn: $\Omega_{\text{CDD}n} = \sum_{\alpha} \sigma_{\alpha} \otimes \tau B_{n,\alpha}$, with $\Omega_{\text{CDD}0} \equiv \Omega$, and $B_{0,\alpha} = B_{\alpha}$; the factor of τ that accompanies every B operator makes it a dimensionless quantity. Since $B_{n,Z} = 0$ at this order of approximation for $n \geqslant 1$, for the norm of the interaction part of $\Omega_{\text{CDD}n}$ to decrease as n increases, it suffices to require $\tau B_{n,X}$ and $\tau B_{n,Y}$ to each decrease in size as n grows. This happens if the map $-4i[\tau B_{n,I}, \cdot]$ is contractive, satisfied if $8\|\tau B_{n,I}\| < 1$. This

gives a maximal concatenation level,

$$n_{\text{max}} = \left[-\log_4(\phi_{\text{B}}) - \frac{3}{2} \right],\tag{43}$$

where $\lceil \cdot \rceil$ denotes the ceiling function. At $n=n_{\rm max}$, the update rules (42) no longer give a contractive map, and the error phase can grow upon further concatenation. For $\phi_{\rm B}=0.001$, this sufficient condition successfully predicts the $n_{\rm max}=4$ numerically observed in Fig. 4. The maximal concatenation level depends only on the norm of τB_I here, but this need not hold in all settings. It is worth mentioning that in Refs. [26], the authors find through numerical investigations that an optimal decoupling strategy is CDD2, when compared with other schemes like PDD, Symmetric DD (a time-symmetrized version of PDD), and CDD4. The existence of such an optimal strategy is consistent with our observation of a maximal concatenation level here.

The existence of a maxima concatenation level, beyond which further concatenation increases—rather than decreases—the noise strength, does not technically contradict the statement that higher-order decoupling is achieved by higher-level CDD. Using decoupling order to quantify the degree of noise removal requires, in the first place, the convergence of the Magnus series, so that the (n + 1)-th and higher-order terms in the series are small corrections to the first n terms. However, if the DD scheme is designed such that the total time for the sequence grows exponentially with n, as is the case for CDD if τ_0 is fixed as n increases, eventually, we violate the convergence criterion and the decoupling order stops being a reasonable indicator of successful noise removal.

That the total sequence time grows exponentially with n also provides the intuition to the nonexistence of an accuracy threshold for CDD, and that CDD fails after some maximal level, even in the ideal (perfect pulses) case. As n grows, the sequence time grows as $4^n\tau$ in this computational setting where the per-pulse time interval is some fixed value τ independent of n. Recalling that DD works to average away noise that remains constant for the full sequence time, as n grows, the sequence time eventually becomes long compared to the timescale of evolution of the noise, reducing the efficacy of the DD averaging. Eventually, the additional concatenation adds only to the total sequence time, without adding much noise-removal power, and we reach the maximal useful concatenation level beyond which further concatenation makes the total (over the full sequence time) noise worse.

To further confirm this intuition, it is useful to momentarily consider the memory setting appropriate for a a different physical situation: to have fixed $\tau_n = T \equiv \tau$, so that the CDD sequence takes the same amount of time, regardless of n. This requires per-pulse time interval $\tau_0 = \frac{\tau}{4^n}$ for each n, so that the physical pulses are applied at shorter and shorter time intervals as the concatenation level increases, up to some practical limit in the pulse rate. In this case, the evolution can be written as

$$\Omega_{n+1} = \mathcal{D}(\Omega_n/4),\tag{44}$$

so that the recursive rules of Eq. (42) become

$$\tau B_{n+1,0} = \tau B_{n,0},$$

$$\tau B_{n+1,1} = -\frac{1}{4}\tau^2 i[B_{n,0}, B_{n,1}],$$

$$\tau B_{n+1,2} = -\frac{1}{8}i\tau^2 [B_{n,0}, B_{n,1}],$$

$$\tau B_{n+1,3} = 0.$$
(45)

This gives

$$\Omega_n \simeq \sigma_0 \otimes \tau B_I + \sigma_1 \otimes (-i)^n 2^{-n(n+1)} \tau^{n+1} a d_{B_I}^n (B_X)$$

$$+ \sigma_2 \otimes (-i)^n 2^{-n(n+2)} \tau^{n+1} a d_{B_I}^{n-1} ([B_I, B_Y] - i \{B_X, B_Z\}).$$
(46)

 $\Phi_{\mathrm{B},n} = \phi_{\mathrm{B}}$ thus remains unchanged, while the error phase is reduced to

$$\Phi_{\mathrm{SB},n} \lesssim 2^{-n^2} \phi_{\mathrm{B}}^n \phi_{\mathrm{SB}},\tag{47}$$

where we have dropped the subleading-order term. For reasonably bounded noise ($\phi_{\rm B} < 2$), the condition $\Phi_{{\rm SB},n+1} < \Phi_{{\rm SB},n}$ automatically holds for all $n \geqslant 0$. The error phase is super-exponentially decreasing in n, suggesting—in sharp contrast to the computational setting—that the noise can be arbitrarily suppressed by increasing the CDD level, limited only by practical constraints on how small τ_0 can be in the experiment.

2. $\phi_{\rm B}/\phi_{\rm SB}\ll 1$

Next, we consider the situation where B_I is negligible compared with B_{α} so that we effectively set $B_I = 0$ (and hence ϕ_B) to the approximation order considered below. As mentioned earlier, one expects CDD to perform better in this case: DD is intended to remove slow—compared with the gate timesnoise, the limiting situation being one where $B_I = 0$ so that the noise in fact does not evolve by itself in the absence of the system. As we will see below, this $B_I = 0$ limit, however, does not mean that an accuracy threshold for CDD exists. The behavior of CDD as the concatenation level increases in fact eventually resembles that of CDD with $B_I \neq 0$. This can be understood intuitively once we recognize—as we will see below—that the CDD pulses cause a back-action—through the system-bath interaction—on the bath itself, such that the effective B_I after adding the DD pulses becomes nonzero, and eventually we get back the situation of the previous subsection.

When $B_I = 0$ (or of higher order in our approximation compared with B_{α}), the terms in Eq. (39) for the effective Hamiltonian for CDDn vanish, except when n = 1 (the PDD case; there, we retain the $\sigma_2 \otimes \{B_X, B_Z\}$ term). That expression is hence no longer useful beyond PDD, as is the case of the recurrence map Eq. (42). Both expressions came from considering only the first and second-order Magnus terms for every n (see Appendix D).

For a nonvanishing contribution for n > 1, we first consider the third-order term for n = 1, the PDD case; we will see why momentarily. Straightforward algebra gives, for negligible B_I ,

$$\Omega_{\mathsf{PDD}}^{(3)} = \frac{2}{3}\tau^{3}[iI \otimes ([B_{Z}, \{B_{X}, B_{Y}\}] - [B_{Y}, \{B_{X}, B_{Z}\}]) + 3\sigma_{z} \otimes \{B_{X}, \{B_{X}, B_{Z}\}\}]. \tag{48}$$

For PDD, this third-order correction is subleading order, compared with the nonvanishing second-order $\Omega_{PDD}^{(2)}$, and can be neglected in the PDD analysis. However, we see that this third-order term gives rise to a nonzero $B_{1,0}$, the bath operator

associated with the identity on the system in Ω_{PDD} . This term arises solely from the H_{SB} bath operators B_X , B_Y , and B_Z , and can be thought of as a back-action on the bath due to its interaction with the system.

Given the recursive structure of CDD, this means that, while at level n=1, the PDD analysis has $B_I=B_{0,0}=0$, at the next level n=2, we no longer have this no-pure-bath-evolution situation, since $B_{1,0} \neq 0$, albeit of a higher order. If we begin our recursion map \mathcal{D} at n=2, rather than n=1, using Ω_{PDD} as the base dimensionless Hamiltonian in place of the physical τH , we are in essence back in the previous case where the pure-bath evolution is no longer vanishing. The former form of \mathcal{D} as specified in Eq. (42) thus holds for $n \geq 2$, with now the base case as n=1, not n=0, with the $B_{1,\alpha}$ s given by

$$\tau B_{1,0} \equiv i\frac{2}{3}\tau^{3}([B_{Z}, \{B_{X}, B_{Y}\}] - [B_{Y}, \{B_{X}, B_{Z}\}]) + O(\phi_{SB}^{4})$$

$$\tau B_{1,1} \equiv 0 + O(\phi_{SB}^{4})$$

$$\tau B_{1,2} \equiv -2\tau^{2}\{B_{X}, B_{Z}\} + O(\phi_{SB}^{4})$$

$$\tau B_{1,3} \equiv 2\tau^{3}\{B_{X}, \{B_{X}, B_{Z}\}\} + O(\phi_{SB}^{4}).$$
(49)

The dimensionless Hamiltonian for CDDn is then estimated for $n \ge 2$, following Eq. (39), as

$$\Omega_{\text{CDD}n} \simeq \sigma_0 \otimes 4^{n-1} \tau B_{1,0} + \sigma_2 \otimes (-i)^{n-1} 2^{(n-1)^2} \tau^n a d_{B,0}^{n-1} (B_{1,2}).$$
 (50)

Here we have kept only the lowest-order terms that contribute to $H_{\rm B,n}$ and $H_{\rm SB,n}$, keeping in mind the various $\tau B_{1,\alpha}$ s are of different orders of magnitude: $\tau B_{1,0}$ and $\tau B_{1,3} \sim \phi_{\rm B}^3$, $\tau B_{1,1} \sim O(\phi_{\rm SB}^4)$, and $\tau B_{1,2} \sim O(\phi_{\rm SB}^2)$.

In this case, the error phase for level-n CDD can be bounded as

$$\Phi_{SB,n} \lesssim 2^{n(n-1)} \|\tau B_{1,0}\|^{n-1} \|\tau B_{1,2}\|
\lesssim \frac{2^{n^2+3n-2}}{3^{n-1}} \phi_{SB}^{3n-1} = \left(\frac{8}{3}\right)^{n-1} 2^{n^2+1} \phi_{SB}^{3n-1}, \qquad (51)$$

noting that $\|\tau B_{1,0}\| \lesssim \frac{16}{3}\phi_{\mathrm{SB}}^3$ and $\|\tau B_{1,2}\| \lesssim 4\phi_{\mathrm{SB}}^2$. Comparing this with the error phase bound of $\Phi_{\mathrm{SB},n} \sim \phi_{\mathrm{B}}^n \phi_{\mathrm{SB}}$, the current situation of $B_I = 0$ gives a faster suppression of ϕ_{SB}^{3n} as n increases. The numerical prefactor still grows exponentially with n, however, so we still expect the accuracy threshold for CDD to vanish even in this case. Figure 5 plots the error-phase bound for this $B_I = 0$ case for $\phi_{\mathrm{SB}} = 0.001$. This shows the same qualitative behavior as in Fig. 4, where $B_I \neq 0$, though the maximal n value is larger in this $B_I = 0$ situation.

A similar analysis as in the $\phi_B/\phi_{SB} \gtrsim 1$ regime leads to an analogous maximal concatenation level for CDD to offer benefit in this $B_I = 0$ setting,

$$n_{\text{max}} = \left[-3\log_4(\phi_{\text{SB}}) + \log_4 3 - \frac{5}{2} \right],$$
 (52)

where we have used the fact that $B_{n+1,0} = 4^n B_{1,0}$ for $n \ge 2$. For $\phi_{\rm SB} = 0.001$, this gives $n_{\rm max} = 14$, which agrees with the maximal n level observed in Fig. 5. We note that, compared with the $\phi_{\rm B}/\phi_{\rm SB} \gtrsim 1$ regime of the previous subsection, this $\phi_{\rm B} \ll 1$ situation presents much more favorable conditions for the performance of CDD, in terms of both the smallest achievable error phase and the maximal concatenation level

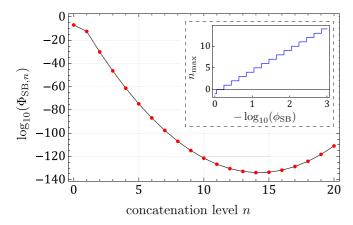


FIG. 5. The error phase as a function of the concatenation level using the theoretical upper bound of Eq. (51), for $\phi_B = 0$ and $\phi_{SB} = 0.001$. The maximal concatenation level occurs at n = 14 for this particular configuration. The inset shows the maximal concatenation level as a function of the error phase, as predicted by (52). Nonpositive n values here indicate that even PDD (n = 1) does not offer any benefit.

(cf. Figs. 4 and 5). The inset to Fig. 5 shows the maximal useful CDD level as a function of ϕ_{SB} , using Eq. (52). We see that DD becomes useful when $\phi_{SB} \lesssim 10^{-0.25} \simeq 0.56$, an improvement compared with the prediction of the generic condition (22) (with $\phi_B = 0$, giving $\phi_{SB} \lesssim 0.25$) based on a loose bound. These results support the intuition that DD performs better in the $\phi_B \ll \phi_{SB}$ regime, although the quantitative behavior, in particular the existence of a maximal concatenation level, is similar in both.

B. Noisy CDD

Next, we discuss the CDD performance when the DD gates themselves are noisy. In particular, we are interested in the strength of the combined noise by studying how the error phase upper bound will evolve with the CDD concatenation level. Since we want to understand the limitations of CDD, we focus only on the $\phi_B/\phi_{SB}\gtrsim 1$ regime, the scenario with poorer CDD performance, as explained earlier.

With imperfect control, the iterative map \mathcal{D} of the ideal situation gets modified by the noise, and we write $\widetilde{\Omega}_{\text{CDD}n+1} = \widetilde{\mathcal{D}}(\widetilde{\Omega}_{\text{CDD}n})$ for $n \geqslant 0$, where, as before, a tilde indicates the noisy version of the quantity, and we have $\widetilde{\Omega}_{\text{CDD}1} \equiv \widetilde{\Omega}_{\text{PDD}}$ and $\widetilde{\Omega}_{\text{CDD}0} \equiv \Omega$. The use of the same noisy map $\widetilde{\mathcal{D}}$ stems from the fact that the same physical gates—with the same noise—are used to implement CDD at every level. The error phase, namely, the norm of the interaction part of each $\widetilde{\Omega}_{\text{CDD}n}$, is denoted as $\widetilde{\Phi}_{\text{SB},n}$.

The accuracy threshold condition under noisy control is given by

$$\widetilde{\Phi}_{\mathrm{SB},n+1} \leqslant \widetilde{\Phi}_{\mathrm{SB},n},$$
 (53)

for every n. From our earlier discussion, we understand that there is no accuracy threshold even with perfect control, but it is still meaningful to study how noise can adversely impact the error phase at each level. For each n, the recursive nature of CDD allows us to view Eq. (53) as the break-even condition for noisy PDD, with Ω replaced by $\widetilde{\Omega}_{\text{CDD}n-1}$. We thus simply

need to understand how $\widetilde{\Phi}_{SB,n}$ and $\widetilde{\Phi}_{B,n}$ evolve as n increases, and make use of the PDD break-even conditions discussed in Sec. III.

We consider the computational setting, with only gatecontrol noise and, for simplicity, instantaneous pulses ($\tau_P = 0$). The noisy X and Z gates are written as

$$\widetilde{X} = Xe^{-i\Gamma_X}$$
 and $\widetilde{Z} = Ze^{-i\Gamma_Z}$, (54)

where we recall that Γ_X and Γ_Z describe the gate-control noise, and are generally operators on both the system and the bath. We write $\Gamma_{X(Z)} \equiv \sum_{i=1}^3 \sigma_i \otimes \Gamma_{X(Z),i}$ where $\Gamma_{X(Z),i}$ acts only on the bath. $\Gamma_{X(Z)}$, as previously noted, can be taken to have no pure-bath term, and $\|\Gamma_{X(Z)}\| \leq \eta$. Our earlier calculation [Eq. (30)] for noisy PDD gives, as the first-order Magnus term,

$$\widetilde{\Omega}^{(1)}_{\mathsf{PDD}} \simeq (4\tau)I \otimes B_I + 2Y \otimes (\Gamma_{X,2} - \Gamma_{Z,2}).$$
 (55)

From this, we know that the pure-bath term approximately (ignoring higher-order corrections) quadruples in size with every concatenation level,

$$\widetilde{\Phi}_{B,n} \simeq 4\widetilde{\Phi}_{B,n-1} = 4^n \phi_B. \tag{56}$$

For the interaction part, the first-order Magnus expression gives a term with norm bounded by 4η . Since CDDn is just PDD on CDD(n-1) with the same noisy gates, this term will occur unchanged at all concatenation levels, giving (again, ignoring higher-order corrections) $\widetilde{\Phi}_{SB,n} \lesssim 4\eta$, an n-independent bound. This is the appropriate estimate of the error phase if η is dominant over the ϕ_{SB} and ϕ_{B} terms that arise in the second-order Magnus term. In the opposite limit, where $\eta \ll \phi_{SB}$, ϕ_{B} , we should recover the ideal PDD behavior, where the first-order Magnus term has no interaction piece, and the leading order correction is the second-order Magnus term [Eq. (7)] for ideal PDD. In this case, we estimate $\widetilde{\Phi}_{SB,n}$ as in Eq. (21), with ϕ_{SB} and ϕ_{B} replaced by $\widetilde{\Phi}_{SB,n-1}$ and $\widetilde{\Phi}_{B,n-1}$, respectively, the noise parameters from CDD(n-1). Putting the two limits together, we can estimate the error phase

$$\widetilde{\Phi}_{\mathrm{SB},n} \lesssim 12 \widetilde{\Phi}_{\mathrm{B},n-1} \widetilde{\Phi}_{\mathrm{SB},n-1} + 4 \widetilde{\Phi}_{\mathrm{SB},n-1}^2 + 4 \eta, \tag{57}$$

now valid for all values of η . Equations (56) and (57) give a pair of recurrence relations that tells us how the noise parameters evolve as n increases.

Intuitively, we can understand the origin of the maximal useful CDD level of concatenation by combining the recurrence relations with the PDD break-even condition. From our understanding of PDD in Sec. III, we know that the break-even condition for PDD identifies a bounded region for ϕ_B and ϕ_{SB} . Yet, we see here that $\widetilde{\Phi}_{B,n}$ —which enters the bound for $\widetilde{\Phi}_{SB,n}$ —grows without bound, indicating that there is no accuracy threshold, i.e., no values of the noise parameters for which condition (53) is satisfied for all n. Such is to be expected, given our conclusion for ideal CDD. We can, moreover, determine the maximal useful CDD level of concatenation from this: The maximal level corresponds to the n such that the $(\widetilde{\Phi}_{B,n},\widetilde{\Phi}_{SB,n})$ pair just crosses the boundary determined by the PDD break-even condition. We illustrate this point in Fig. 6(a), where three different systems with

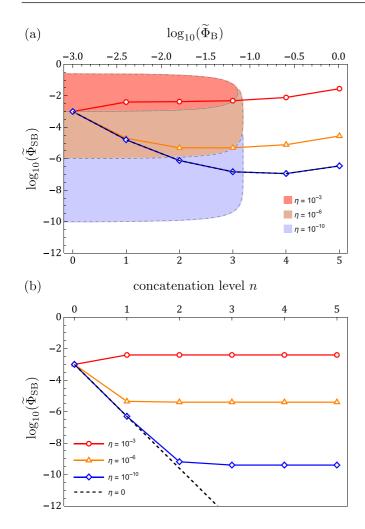


FIG. 6. The evolution of the $(\widetilde{\Phi}_B, \widetilde{\Phi}_{SB})$ pair with increasing concatenation level for (a) the computational setting and (b) the memory setting. In both plots, the three colored lines, from top to bottom, are for the overall pulse errors of $\eta=10^{-3}$, 10^{-6} , and 10^{-10} , respectively, while the black dashed line gives the noiseless limit [in (a) it is nearly coincident with the $\eta=10^{-10}$ line]. The leftmost marker for each line is the n=0 (i.e., no DD) point, and we have $n=1,2,3,\ldots$ for the subsequent markers as we move rightwards on the line. The shaded regions in (a) are the PDD (n=1) break-even regions for the different values of η in the computational setting.

control noise levels $\eta=10^{-3}$, 10^{-6} , and 10^{-10} are explored, indicated in red, orange, and blue shades, respectively. Figure 6(a) plots the regions bounded by the break-even conditions for those systems, and simulates the evolution of $(\tilde{\Phi}_B, \tilde{\Phi}_{SB})$ for the effective Hamiltonian as one increases the concatenation level, for the same starting values of $\phi_B=\phi_{SB}=0.001$. Since $\log(\tilde{\Phi}_B)$ is linear in n [see Eq. (56)], the concatenation level can be alternatively read off from the horizontal axis. For small enough η , we observe a similar behavior as in the ideal CDD case: $\Phi_{SB,n}$ first decreases then increases as n grows, and we can identify a maximal n beyond which scaling up CDD gives no further benefit. That maximal n value is the very last level where its previous level still lies within the break-even region (not counting the boundary). For $\eta=10^{-3}$, we see that even n=1 is of no use: The error phase

 $\Phi_{SB,n}$ for every n is larger than if no DD is used, as the bare Hamiltonian lies on the break-even boundary to begin with.

Again, we can examine the memory setting for comparison. In that case, $\widetilde{\Phi}_{B,n} \simeq \phi_B$ is approximately constant across the concatenation levels, while the error phase, following a similar logic as above, updates according to the rule

$$\widetilde{\Phi}_{\mathrm{SB},n} \lesssim \frac{1}{2} \phi_{\mathrm{B}} \widetilde{\Phi}_{\mathrm{SB},n-1} + 4\eta.$$
 (58)

Assuming the upper bound provides a good estimate of the actual quantity, we solve this recursive relation to obtain

$$\widetilde{\Phi}_{\mathrm{SB},n} \simeq \left(\frac{1}{2}\phi_{\mathrm{B}}\right)^{n}\phi_{\mathrm{SB}} + \frac{4\eta}{1 - \frac{1}{2}\phi_{\mathrm{B}}},\tag{59}$$

keeping terms up to order n + 1 in small quantities. As long as $\phi_{\rm B} < 2$, $\Phi_{{\rm SB},n}$ decreases as n increases. However, unless the control pulses are perfect, there will always be an n-independent remnant—the second term in Eq. (59) proportional to η —that will limit how small $\Phi_{SB,n}$ can be, not to mention the eventual limit in n when $\tau (= T/4^n)$ for this memory setting gets too short to be feasible. As illustrated in Fig. 6(b), the error phase reduces and plateaus to the same order as η as one increases the concatenation level. Thus, in this case of noisy pulses, there is a limit in the error suppression capability of CDD, determined entirely by the size of the gate control noise. Hence, arbitrary noise suppression with CDD is not possible with imperfect control even in the memory setting. Note that this conclusion does not rely on us equating $\Phi_{SB,n}$ with its upper bound above, but simply results from the linear-in- η term in Eq. (55).

V. CONCLUSIONS

In this work, we studied the question of the fault tolerance of DD, namely, the efficacy of DD schemes in the presence of imperfections in the very pulses that carry out the DD operations. Our results are summarized in Table I. We examined first the break-even conditions on the noise parameters for single-level PDD, and then we analyzed the performance of CDD. We saw that, in the computational setting, of primary relevance today, there is generally a limit to the CDD concatenation level, beyond which further concatenation offers no added benefit. This is in contrast to the memory setting, where unlimited error suppression is possible in some cases: Here we saw this for the case of ideal CDD, while the case of finite-width pulses gave similar conclusions in Ref. [5]. In the case where gate control noise dominates, CDD faces a limit even in the memory setting, determined by the strength of the gate control noise.

That CDD faces a limit eventually is simply a reflection of the fact that the benefit, namely the error suppression capability, of CDD fails to grow sufficiently fast to compensate for the added cost, namely, the additional gate control noise, from the increased number of pulses needed for larger n. It is easy to understand how this arises in the computational setting: Even in the case of ideal pulses, the exponential lengthening of the pulse sequence time as the CDD level increases means a relative increase in the noise evolution rate [see Eq. (56)]—whether the noise changes quickly or slowly is relative to the pulse sequence time—until eventually, for large enough n, the noise changes much faster than the

Assumptions	Setting	Condition	Eq(s). in text
	PDD, brea	ak-even conditions	
Ideal	Computational	$12\phi_{\mathrm{B}} + 4\phi_{\mathrm{SB}} \leqslant 1$	(22)
Noisy, general	Computational	$(8/\pi)\phi_{ m SB}\delta + 4\eta \leqslant \phi_{ m SB}$	(31)
Noisy, $\eta = 0$	Computational	$ au_P/ au \leqslant 8/\pi \simeq 0.4$	(32)
Noisy, $\tau_P = 0$	Computational	$\eta\leqslant\phi_{ m SB}/4$	(33)
Unitary control errors only, $\tau_P = 0$	Computational	$ \eta \leqslant \sqrt{\phi_{\rm SB}/2} [1 - 4(3\phi_{\rm B} - \phi_{\rm SB})]^{-\frac{1}{2}} - 2\phi_{\rm SB} \approx \sqrt{\phi_{\rm SB}/2} \text{ (when } \phi_{\rm B}, \phi_{\rm SB} \ll 1) $	(37)
	CDD, maxim	al concatenation level	
Ideal, $\phi_{ m B} \gtrsim \phi_{ m SB}$	Computational	$\lceil -\log_4(\phi_{\rm B}) - 3/2 \rceil$	(43)
Ideal, $\phi_{ m B} \gtrsim \phi_{ m SB}$	Memory	no maximal n , with $\Phi_{\rm SB} \to 0$	_
Ideal, $\phi_{\rm B} = 0$	Computational	$[-3\log_4(\phi_{\rm SB}) + \log_4 3 - 5/2]$	(52)
Noisy, $\tau_P = 0$	Computational	recurrence relations for maximal n	(56) and (57)
Noisy, $\tau_P = 0$	Memory	no maximal n , with $\widetilde{\Phi}_{\mathrm{SB},n} \to 4\eta/(1-\phi_{\mathrm{B}}/2)$	(59)

TABLE I. Summary of results.

pulse sequence time, and DD stops being effective. This is exacerbated by imperfections in the pulses, with the limiting n reached earlier than in the ideal case; see Fig. 6(a) as an example. In contrast, the memory setting sees no such relative increase in noise evolution rate as, in this case, all CDDn sequences, for any n, are assumed to take the same total time.

That gate control noise sets a limit to the error suppression capability of CDD can also be understood intuitively. Any gate-dependent errors is equivalent to noise that changes during the pulse sequence and hence cannot be effectively averaged away by the DD sequence. This is in contrast to any gate-independent noise, which includes noise that arises from the finite-width pulses due to the always-on and time-independent H_{SB} during the pulse application. Such static noise—referred to as "systematic errors" in Refs. [5] can be removed by the DD sequence. Note that Refs. [5] hinted at some robustness of CDD against what they termed "random errors," as opposed to systematic errors. Our gatedependent gate control noises are closer to the random errors mentioned in Refs. [5], but they are not quite random enough for truly randomized averaging effects, hinted at in Ref. [5], to kick in. Yet such kinds of gate-dependent errors are generic in experiments today, with the same noise statistics manifesting each time the same gate is applied.

While we chose PDD and CDD as our focus here, the insights gained from them apply to arbitrary scalable DD schemes. As long as the DD pulse sequence gets longer as one scales up, the above intuition that the noise evolution rate increases relatively continues to hold true, and one again expects a limit in the efficacy of CDD.

Last, it is worthwhile to note that, even though we conclude here that there is a limit to the usefulness of CDD concatenation in the realm of realistic pulses, we should remember that CDD is anyway not meant as the final solution to noise in a quantum computing device. Instead, it is a lower-level means to remove slow noise, with higher-level methods like error correction taking over eventually to eliminate any remnant noise. The advantage of CDD, over that of error correction, is its much lower-resource requirements, and this remains a key consideration in near- to middle-term quantum

devices. One can simply employ CDD to the maximum n limit, weakening the noise as much as possible, before switching over to more expensive, but more powerful, quantum error correction.

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APPENDIX A: RELATING THE ERROR PHASE AND INFIDELITY MEASURES

Here we relate the error phase (see Sec. II D 1) and infidelity (see Sec. II D 2) measures. We consider the system and bath evolution for some time T, with effective dimensionless Hamiltonian Ω ($U(T,0)=e^{-i\Omega}$), using the appropriate Ω for the bare evolution without DD, or in the case with DD. Given our context of weak noise and slow evolution of the bath for good DD performance, we assume Ω to be small in norm and write the system-bath evolution as a power series in Ω using the Baker-Hausdorff lemma [47], so that we have

$$\mathcal{N}(\rho_{S}) = \operatorname{tr}_{B} \{ U(T, 0)(\rho_{S} \otimes \rho_{B}) U(T, 0)^{\dagger} \}$$

$$= \sum_{n=0}^{\infty} \frac{(-i)^{n}}{n!} \operatorname{tr}_{B} \{ ad_{\Omega}(\rho_{S} \otimes \rho_{B}) \}, \tag{A1}$$

where ad_{Ω} is the map $ad_{\Omega}(\cdot) \equiv [\Omega, \cdot]$. With this [see Eq. (15)], we can write

$$InF(\mathcal{N}, \psi)^2 = \langle \psi | (\mathcal{I} - \mathcal{N})(\psi) | \psi \rangle \equiv \sum_{n=1}^{\infty} f_n(\mathcal{N}, \psi), \quad (A2)$$

for $f_n(\mathcal{N}, \psi) \equiv -\frac{(-i)^n}{n!} \langle \psi | \text{tr}_{B} \{ ad_{\Omega}^n(\psi \otimes \rho_{B}) \} | \psi \rangle = -\frac{(-i)^n}{n!} \text{tr} \{ \psi \ ad_{\Omega}^n(\psi \otimes \rho_{B}) \}$, from the order-*n* term in the power series (A1). This allows us to examine the infidelity measure

as a series also in Ω , with the leading orders giving an approximate expression.

To further evaluate the f_n s, we note two identities,

$$\operatorname{tr}\{\psi ad_A(\psi \otimes \rho_{\mathbf{B}})\} = 0, \tag{A3}$$

and
$$\operatorname{tr}_{\mathbf{B}} a d_{\mathbf{B}} = 0$$
, (A4)

where A is an arbitrary system-bath operator, while B is a bath-only operator, i.e., acts nontrivially only on the bath. Eq. (A3) immediately tells us that the first-order term $f_1(\mathcal{N}, \psi)$ always vanishes. For the second-order term, we first write $\Omega = \Omega_{\rm B} + \Omega_{\rm SB}$ with $\Omega_{\rm B} \equiv I \otimes {\rm tr}_{\rm S}\Omega$ as the bath-only part, and $\Omega_{\rm SB} \equiv \Omega - \Omega_{\rm B}$ is the system-bath term. Then, invoking the Jacobi identity, $ad_Aad_{A'} = ad_{A'}ad_A + ad_{[A,A']}$, together with Eqs. (A3) and (A4), we find

$$f_2(\mathcal{N}, \psi) = \frac{1}{2} \operatorname{tr} \{ \psi a d_{\Omega_{SB}}^2(\psi \otimes \rho_B) \}.$$
 (A5)

 $f_2(\mathcal{N}, \psi)$ hence does not depend on Ω_B and scales as the square of the error phase $\|\Omega_{SB}\|^2$. In fact, from Eq. (A5), we can write the following bound:

$$f_2(\mathcal{N}, \psi) \leqslant \frac{1}{2} \|\rho_{\mathcal{S}} \otimes I\| \|ad_{\Omega_{\mathcal{S}B}}^2(\rho_{\mathcal{S}} \otimes \rho_{\mathcal{B}})\|_{\text{tr}}$$

$$\leqslant 2 \|\Omega_{\mathcal{S}B}\|^2 \|\rho_{\mathcal{S}} \otimes \rho_{\mathcal{B}}\|_{\text{tr}} = 2 \|\Omega_{\mathcal{S}B}\|^2. \tag{A6}$$

Assuming $f_2(\mathcal{N}, \psi)$ nonvanishing, we thus have that $InF(\mathcal{N}, \psi)^2 \simeq f_2(\mathcal{N}, \psi) \leqslant 2\|\Omega_{SB}\|^2$ for all \mathcal{N} and ψ , so that

$$0 \leqslant \operatorname{InF} \lesssim \sqrt{2} \|\Omega_{SB}\|. \tag{A7}$$

APPENDIX B: NORM INEQUALITY

Here we prove an inequality used in the main text,

$$\max_{i \in \{1,2,3\}} \|B_i\| \leqslant \left\| \sum_{i=1}^3 \sigma_i \otimes B_i \right\|,\tag{B1}$$

for B_i s Hermitian operators, σ_i the usual Pauli operators, and $\|\cdot\|$ is the operator norm $\max_{x:\|x\|=1}|x^\mathsf{T}\cdot x|$. In the main text, we had used this inequality in the context where B_i s are the bath operators associated with σ_i s on the system in the interaction Hamiltonian H_{SB} .

Proof. We first prove a lemma: For A_i Hermitian and A_0 positive semidefinite, we show that $\|I \otimes A_0 + \sum_{i=1}^3 \sigma_i \otimes A_i\| \geqslant \|A_0\|$. Let $M \equiv I \otimes A_0 + \sum_{i=1}^3 \sigma_i \otimes A_i$. It can be written in a block form, following the standard matrix representation of σ_i s, as

$$M = \begin{pmatrix} A_0 + A_3 & A_1 - iA_2 \\ A_1 + iA_2 & A_0 - A_3 \end{pmatrix}.$$
 (B2)

Following the same block form, an arbitrary vector can be written as $(x, y)^T$, with x and y vectors (columns) themselves. Let x be the (unit-length) eigenvector corresponding to the greatest eigenvalue of A_0 . Since $A_0 \ge 0$, $||A_0|| = x^T A_0 x$. We calculate the following:

$$(x^{\mathsf{T}} \quad 0)M \begin{pmatrix} x \\ 0 \end{pmatrix} = x^{\mathsf{T}} A_0 x + x^{\mathsf{T}} A_3 x = ||A_0|| + x^{\mathsf{T}} A_3 x.$$
 (B3)

This means that $||M|| \ge ||A_0|| + x^T A_3 x$. A similar argument with $(0 \ x)^T$ in place of $(x \ 0)^T$ gives $||M|| \ge ||A_0|| - x^T A_3 x$.

Combining the two inequalities gives the desired result: $||M|| \equiv ||I \otimes A_0 + \sum_{i=1}^{3} \sigma_i \otimes A_i|| \geqslant ||A_0||$.

With this, we can prove Eq. (B1). We first note that $||A||^2 = ||A^{\dagger}A||$ for any operator A. Since σ_i are Hermitian, we have

$$\left\| \sum_{i=1}^{3} \sigma_{i} \otimes B_{i} \right\|^{2} = \left\| \sum_{ij} \sigma_{i} \sigma_{j} \otimes B_{i}^{\dagger} B_{j} \right\|$$

$$= \left\| I \otimes \sum_{i} B_{i}^{\dagger} B_{i} + \sum_{i \neq j} \sigma_{i} \sigma_{j} \otimes B_{i}^{\dagger} B_{j} \right\|$$

$$\geqslant \left\| \sum_{i} B_{i}^{\dagger} B_{i} \right\| \geqslant \max_{i} \|B_{i}\|^{2}, \tag{B4}$$

where, in the last line, we have used the above-proven lemma, with $A_0 \equiv \sum_i B_i^\dagger B_i$ and recognizing that $\sum_{i \neq j} \sigma_i \sigma_j \otimes B_i^\dagger B_j$ can be written in the form of $\sum_i \sigma_i \otimes A_i$. The final inequality follows from the fact that $\langle \psi | O | \psi \rangle \leqslant \langle \psi | O | \psi \rangle + \langle \psi | O' | \psi \rangle$, for any O and O' positive semidefinite operators.

APPENDIX C: NOISY PDD DERIVATION

Here we provide the derivation leading to Eq. (28), for the effective Hamiltonians K_i s (i = 0, 1, 2, 3) for PDD with noisy gates as detailed in the main text. We recall the definitions of the K_i here:

$$e^{-iK_0} \equiv X\widetilde{X}e^{i\tau_{\rm P}H},$$

$$e^{-iK_1} \equiv Y\widetilde{Z}e^{i\tau_{\rm P}H}X,$$

$$e^{-iK_2} \equiv Z\widetilde{X}e^{i\tau_{\rm P}H}Y,$$

$$e^{-iK_3} \equiv \widetilde{Z}e^{i\tau_{\rm P}H}Z,$$
(C1)

with $\widetilde{Z}(X) \equiv e^{-i\tau_{\rm P}(H_{Z(X)}+H)}e^{-i\Gamma_{Z(X)}}$ and $H_{Z(X)} \equiv \frac{\pi}{2\tau_{\rm P}}Z(X)$. The goal here is to work out expressions for K_i , to lowest order in small quantities $\phi_{\rm SB}\delta \equiv \|\tau_{\rm P}H_{\rm SB}\|$ and $\eta \geqslant \|\Gamma_{X(Z)}\|$.

We first need a key formula, which we show here as a lemma,

$$e^{A+\gamma B} = e^{A} \left[1 + \gamma \left(\frac{1 - e^{-ad_A}}{ad_A} \right) B + O(\gamma^2) \right], \quad (C2)$$

where A and B are arbitrary operators, and γ is a small scalar parameter. Here ad_A is the map $ad_A(\cdot) \equiv [A, \cdot]$, and $\frac{1-e^{-ad_A}}{ad_A}$ should be understood in terms of a Taylor series,

$$\frac{1 - e^{-ad_A}}{ad_A} = \sum_{n=0}^{\infty} \frac{(-1)^n}{(n+1)!} (ad_A)^n.$$
 (C3)

Proof. We begin with a standard formula available from Lie-algebra textbooks (see, for example, [47]), applicable to any operator X,

$$\frac{d}{dt}e^X = e^X \frac{1 - e^{-ad_A}}{ad_A} \frac{d}{dt}X.$$
 (C4)

We want to write $e^{A+\gamma B}$ as a power series in γ ,

$$e^{A+\gamma B} = e^A [1 + \gamma C + O(\gamma^2)],$$
 (C5)

for some C. C can be calculated as

$$C = \left[e^{-A} \frac{d}{dt} (e^{A + \gamma B}) \right]_{\gamma = 0} = \left(\frac{1 - e^{-ad_A}}{ad_A} \right) B, \quad (C6)$$

where we have made use of the identity (C4). This gives, immediately, formula Eq. (C2), as desired.

Armed with Eq. (C2), we can work out, say, K_3 . We first observe that $e^{-iK_3} \equiv \widetilde{Z}e^{i\tau_P H}Z = \widetilde{Z}Ze^{i\tau_P ZHZ} = \widetilde{Z}Ze^{i\delta\Omega_3}$, for $\Omega_3 \equiv Z\Omega Z = \tau ZHZ$. We compute $\widetilde{Z}Z$:

$$\widetilde{Z}Z = Ze^{-i(\tau_{\rm P}H_{\rm Z} + \delta\Omega_3)}e^{-iZ\Gamma_{\rm Z}Z}$$

$$= \left\{ 1 - i\tau_{\mathrm{P}} \sum_{n=0}^{\infty} \frac{(-\pi)^n}{(n+1)!} \left(\frac{ad_Z}{2i} \right)^n (ZHZ) + O(\gamma^2) \right\} e^{-iZ\Gamma_Z Z}, \tag{C7}$$

where we have used the key formula Eq. (C2) on the exponential $e^{-i(\tau_{\rm P}H_Z+\delta\Omega_3)}$, with $A\equiv -i\tau_{\rm P}H_Z$ and $\gamma B\equiv -i\delta\Omega_3$, so that $\gamma\equiv\delta\phi_{\rm SB}$ is a small parameter. We also note that $e^A\dot{=}Z$, where we recall that the notation $\dot{=}$ denotes equality up to an overall phase.

Now, recall that $H = \sum_{\alpha=0}^{3} \sigma_{\alpha} \otimes B_{\alpha}$. We let $h_Z \equiv -X \otimes B_X - Y \otimes B_Y$ so that $ZHZ = h_Z + I \otimes B_I + Z \otimes B_Z$. Observe that $(ad_Z)^0(ZHZ) = (ad_Z)^0(h_Z) + I \otimes B_I + Z \otimes B_Z$, and for integer n > 0, $(ad_Z)^n(ZHZ) = (ad_Z)^n(h_Z)$. It is easy to show that, for n = 0, 1, 2, ...,

$$\left(\frac{ad_Z}{2i}\right)^n (h_Z) = \begin{cases} (-1)^{n/2} h_Z, & n \text{ even,} \\ (-1)^{(n-1)/2} h_Z', & n \text{ odd,} \end{cases}$$
(C8)

for $h_Z' \equiv X \otimes B_Y - Y \otimes B_X$. From this, straightforward algebra gives

$$e^{-iK_3} = \widetilde{Z}Ze^{i\delta\Omega_3}$$

$$= \left\{ 1 - i\tau_{P}(I \otimes B_I + Z \otimes B_Z) + i\tau_{P}\frac{2}{\pi}h'_z + O(\gamma^2) \right\}$$

$$\times e^{-iZ\Gamma_Z Z}e^{i\delta\Omega_3}$$

$$= \exp\left[-i\tau_{P}\left(I \otimes B_I - \frac{2}{\pi}X \otimes B_Y + \frac{2}{\pi}Y \otimes B_X + Z \otimes B_Z \right) - iZ\Gamma_Z Z + i\delta\Omega_3 + O(\gamma^2, \gamma\eta) \right]. \quad (C9)$$

From this, we obtain K_3 as given in Eq. (28) after further simplification. K_0 , K_1 , and K_2 are derived in a similar manner.

APPENDIX D: IDEAL CDD DYNAMICS

Here we analyze the dynamics of the system under ideal CDD pulses, giving the derivation to Eq. (39) and related statements used in the main text.

The theory of Lie groups guarantees the existence of a homeomorphism between the Lie-group elements around I and its Lie-algebra elements in the vicinity of 0. This suggests

a one-to-one correspondence between the unitary dynamics U_n for CDDn and the Hermitian generator $\Omega_{\text{CDD}n} \equiv i \log(U_n)$ when noise is weak. It is more convenient to focus on the generators and regard DD as transformation among them as n changes. At the base level, we have the bare Hamiltonian $\Omega \equiv \tau H$. For n=1, CDD gives $\Omega_{\text{CDD}1} \equiv \Omega_{\text{PDD}}$, which can be expressed as a series $\sum_{m=1}^{\infty} \Omega_{\text{CDD}1}^{(m)}$ through the Magnus expansion, as we have done in the main text. To characterize this process, we introduce the decoupling maps \mathcal{D} and $\mathcal{D}^{(m)}$, defined as

$$\Omega_{\mathsf{CDD1}} = \mathcal{D}(\Omega), \quad \Omega_{\mathsf{CDD1}}^{(m)} = \mathcal{D}^{(m)}(\Omega), \qquad (D1)$$

with $\mathcal{D} \equiv \sum_{m=1}^{\infty} \mathcal{D}^{(m)}$ following the Magnus series. In particular, $\mathcal{D}^{(1)}$ and $\mathcal{D}^{(2)}$ are explicitly given in Eqs. (6) and (7). Higher-level concatenations are defined iteratively,

$$\Omega_{\text{CDD}k+1} = \mathcal{D}(\Omega_{\text{CDD}k}) = \sum_{m=1}^{\infty} \mathcal{D}^{(m)}(\Omega_{\text{CDD}k}).$$
(D2)

Backtracking from level n to level 0, we have $\Omega_{\text{CDD}n} = (\mathcal{D})^n(\Omega)$. In the end, $\Omega_{\text{CDD}n}$ can be expanded as a Magnus series in the bare Hamiltonian,

$$\Omega_{\text{CDD}n} = \left(\sum_{m=1}^{\infty} \mathcal{D}^{(m)}\right)^{n} (\Omega) = \sum_{m=1}^{\infty} \Omega_{\text{CDD}n}^{(m)}, \quad (D3)$$

where $\Omega_{\text{CDD}n}^{(m)} \sim \|\Omega\|^m$ is the *m*th-order term in Ω .

There is no analytical solution to the full $\Omega_{\text{CDD}n}$. But to estimate CDD dynamics, it suffices to find an analytically solvable estimator $\widehat{\Omega}_{\text{CDD}n}$ that reflects the leading-order behavior of the full series. We need the leading-order behavior of the pure-bath part and system-bath coupling separately, as the two pieces can be of different orders in small quantities, and we make use of them separately in our analyses. Formally, we define an estimator-error pair,

$$\Omega_{\text{CDD}n} = \widehat{\Omega}_{\text{CDD}n} + \delta \Omega_{\text{CDD}n}. \tag{D4}$$

For an accurate estimator, the error term $\delta\Omega_{\text{CDD}n}$ must be of higher-order smallness than the estimator, both in the purebath part and in the coupling. To quantify this statement, we introduce a two-component norm for any operator O with a split into the bath-only and system-bath coupling terms: $||O||| \equiv (||O_{\text{B}}||, ||O_{\text{SB}}||)$. Here we are interested in the two-component norm of both $\delta\Omega_{\text{CDD}n}$ and $\widehat{\Omega}_{\text{CDD}n}$. Accuracy of our estimation scheme demands

$$\||\delta\Omega_{\mathsf{CDD}n}|\| \equiv (\delta\phi_{\mathsf{B},n}, \delta\phi_{\mathsf{SB},n}) \ll \||\widehat{\Omega}_{\mathsf{CDD}n}|\| \equiv (\widehat{\phi}_{\mathsf{B},n}, \widehat{\phi}_{\mathsf{SB},n}),$$
(D5)

for every n. Here $\widehat{\phi}_{B(SB),n}$ and $\delta\phi_{B(SB),n}$ denote the norms of the pure-bath (system-bath coupling) parts of $\widehat{\Omega}_{CDDn}$ and $\delta\Omega_{CDDn}$, respectively. The comparison is implied for both components separately, and should be understood as a comparison of the leading powers of the polynomials, i.e., a smaller polynomial has a larger leading power.

Reference [5] constructed an estimator by keeping the first two Magnus terms for each iteration step. In our notation, we write this as

$$\widehat{\Omega}_{CDDk+1} \equiv (\mathcal{D}^{(1)} + \mathcal{D}^{(2)})(\widehat{\Omega}_{CDDk}). \tag{D6}$$

We explicitly decompose $\widehat{\Omega}_{CDDk}$ as $\sum_i \sigma_i \otimes \widehat{B}_{k,i}$. Eq. (7) suggests that $\widehat{B}_{k,Z}$ vanishes beyond the first level of concatenation. This gives a simple update rule that connects higher concatenation levels:

$$\tau B_{n+1,I} = 4\tau B_{n,I}
\tau B_{n+1,X} = -4i[\tau B_{n,I}, \tau B_{n,X}],
\tau B_{n+1,Y} = -2i[\tau B_{n,I}, \tau B_{n,X}],
\tau B_{n+1,Z} = 0.$$
(D7)

After a little bit of algebra, the estimator can be shown to be

$$\widehat{\Omega}_{CDDn} = I \otimes 4^{n} \tau B_{I} + X \otimes (-i)^{n} 2^{n(n+1)} \tau^{n+1} a d_{B_{I}}^{n} (B_{X})$$

$$+ Y \otimes (-i)^{n} 2^{n^{2}} \tau^{n+1} a d_{B_{I}}^{n-1} ([B_{I}, B_{Y}] - i \{B_{X}, B_{Z}\}).$$
(D8)

To track orders, we focus on the leading powers in τ , ignoring any coefficients. According to Eq. (D8), we have

$$(\widehat{\phi}_{B,n}, \widehat{\phi}_{SB,n}) \simeq (\tau, \tau^{n+1}).$$
 (D9)

To show that $\widehat{\Omega}_{CDDn}$ is indeed faithful, $\delta \phi_{B,n}$ and $\delta \phi_{SB,n}$ need to be of higher-order smallness compared to $\widehat{\phi}_{B,n}$ and $\widehat{\phi}_{SB,n}$. Indeed, we claim that

$$(\delta\phi_{\mathrm{B},n}, \delta\phi_{\mathrm{SB},n}) \lesssim (\tau^3, \tau^{n+2}).$$
 (D10)

We prove this assertion by mathematical induction.

Proof. For n = 1, the estimation error comes solely from the series truncation error, which is led by the third-order term,

$$(\delta\phi_{\mathrm{B},1}, \delta\phi_{\mathrm{SB},1}) = \left\| \left| \sum_{k=3}^{\infty} \Omega_{\mathsf{PDD}}^{(k)} \right\| \simeq (\tau^3, \tau^3).$$
 (D11)

At higher truncation levels, the estimation errors are also supplemented with contributions from the lower-level terms. Formally,

$$\begin{split} \delta\Omega_{\text{CDD}n+1} &= \mathcal{D}(\Omega_{\text{CDD}n}) - (\mathcal{D}^{(1)} + \mathcal{D}^{(2)})(\widehat{\Omega}_{\text{CDD}n}) \\ &= \mathcal{D}^{(1)}(\delta\Omega_{\text{CDD}n}) + \sum_{m=2}^{\infty} \mathcal{D}^{(m)}(\Omega_{\text{CDD}n}) \\ &- \mathcal{D}^{(2)}(\widehat{\Omega}_{\text{CDD}n}), \end{split} \tag{D12}$$

where we have used the fact that $\mathcal{D}^{(1)}$ is a linear map. To properly bound the size of the error term, we take the two-component norm on both sides of Eq. (D12) and apply the

triangle inequality:

$$(\delta \phi_{\mathrm{B},n+1}, \delta \phi_{\mathrm{SB},n+1}) \lesssim \||\mathcal{D}^{(1)}(\delta \Omega_{\mathrm{CDD}n})|\|$$

$$+ \||\mathcal{D}^{(2)}(\Omega_{\mathrm{CDD}n}) - \mathcal{D}^{(2)}(\widehat{\Omega}_{\mathrm{CDD}n})|\|$$

$$+ \left\| \sum_{m=3}^{\infty} \mathcal{D}^{(m)}(\Omega_{\mathrm{CDD}n}) \right\|. \tag{D13}$$

We need to show that the right-hand side is bounded by (τ^3, τ^{n+3}) .

Reference [5] demonstrated that the higher-order Magnus series $\sum_{m=3}^{\infty} \mathcal{D}^{(m)}(\widehat{\Omega}_{\text{CDD}n})$ is small. However, Eq. (D12) suggests that this argument is not sufficient, as the estimation error $\delta\Omega_{n+1}$ comes not only from the truncation error at the same level, but also propagates up from the lower-level error $\delta\Omega_{\text{CDD}n}$. Let us examine the size of these terms. The first term in (D13) is straightforward,

$$\||\mathcal{D}^{(1)}(\delta\Omega_{\text{CDD}n})\|\| = (4\delta\phi_{\text{B},n}, 0) \simeq (\tau^3, 0).$$
 (D14)

To bound the second term in (D13), we need to calculate the difference between the full $\Omega_{\text{CDD}n}$ and $\widehat{\Omega}_{\text{CDD}n}$ after applying $\mathcal{D}^{(2)}$. Using the explicit expression for the second-order Magnus term, we obtain

$$\begin{aligned} \||\mathcal{D}^{(2)}(\Omega_{\mathsf{CDD}n}) - \mathcal{D}^{(2)}(\widehat{\Omega}_{\mathsf{CDD}n})\|\| \\ &\simeq (0, \delta\phi_{\mathsf{B}\,n}\phi_{\mathsf{SB}\,n} + \phi_{\mathsf{B}\,n}\delta\phi_{\mathsf{SB}\,n}) \simeq (0, \tau^{n+3}), \quad (D15) \end{aligned}$$

where we have applied the induction hypothesis $\phi_{B,n} \simeq \widehat{\phi}_{B,n} \simeq \tau$ and $\phi_{SB,n} \simeq \widehat{\phi}_{SB,n} \simeq \tau^{n+1}$. The final term in (D13) involves an infinite sum of Magnus terms higher than the third order. Since the series is led by the third-order term, which can be explicitly bounded, we have

$$\left\| \left\| \sum_{m=3}^{\infty} \mathcal{D}^{(m)}(\Omega_{\text{CDD}n}) \right\| \right\| \simeq \left\| \left| \mathcal{D}^{(3)}(\Omega_{\text{CDD}n}) \right| \right\|$$

$$\lesssim \left(\phi_{\text{SB},n}^2(\phi_{\text{B},n} + \phi_{\text{SB},n}), \phi_{\text{SB},n}(\phi_{\text{B},n} + \phi_{\text{SB},n})^2 \right)$$

$$\simeq (\tau^{2n+2}, \tau^{n+3}). \tag{D16}$$

The estimation for $\mathcal{D}^{(3)}(\Omega_{\text{CDD}n})$ can be done by explicitly calculating the third-order Magnus term for PDD, whose expression is not relevant here. Since all three terms are bounded by (τ^3, τ^{n+3}) , so is their sum. This proves our assertion Eq. (D10).

We have now shown that the estimator constructed through Eq. (D6) is indeed accurate. This leads to the conclusion that ideal CDDn indeed achieves nth-order decoupling, with leading-order behavior given by $\widehat{\Omega}_n$ in Eq. (D8), which is also Eq. (39).

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